

### Full Speed USB Flash MCU Family

### **Analog Peripherals**

### 10-Bit ADC (C8051F340/1/2/3/4/5/6/7/A/B only)

- Up to 200 ksps
- Built-in analog multiplexer with single-ended and differential mode
- VREF from external pin, internal reference, or VDD
- Built-in temperature sensor External conversion start input option
- Two comparators
- Internal voltage reference
- (C8051F340/1/2/3/4/5/6/7/A/B only) Brown-out detector and POR Circuitry

### **USB Function Controller**

- USB specification 2.0 compliant
- Full speed (12 Mbps) or low speed (1.5 Mbps) operation
- Integrated clock recovery: no external crystal required for
- full speed or low speed
- Supports eight flexible endpoints
- 1 kB USB buffer memory
- Integrated transceiver; no external resistors required

### **On-Chip Debug**

- On-chip debug circuitry facilitates full speed, non-intrusive in-system debug (No emulator required)
- Provides breakpoints, single stepping, inspect/modify memory and registers
- Superior performance to emulation systems using ICE-chips, target pods, and sockets

### Voltage Supply Input: 2.7 to 5.25 V

Voltages from 3.6 to 5.25 V supported using On-Chip Voltage Regulator

### High Speed 8051 µC Core

- Pipelined instruction architecture; executes 70% of Instructions in 1 or 2 system clocks
- 48 MIPS and 25 MIPS versions available.
- Expanded interrupt handler

#### Memory

- 4352 or 2304 Bytes RAM
- 64 or 32 kB Flash; In-system programmable in 512-byte sectors

#### **Digital Peripherals**

- 40/25 Port I/O; All 5 V tolerant with high sink current
- Hardware enhanced SPI<sup>™</sup>, SMBus<sup>™</sup>, and one or two enhanced UART serial ports
- Four general purpose 16-bit counter/timers
- 16-bit programmable counter array (PCA) with five capture/compare modules
- External Memory Interface (EMIF)

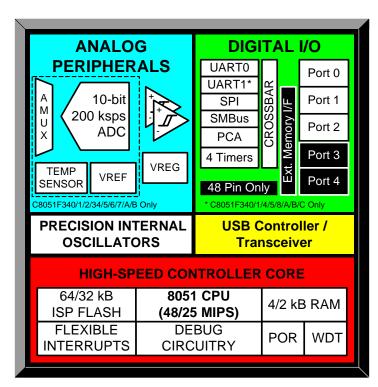
#### **Clock Sources**

- Internal Oscillator: ±0.25% accuracy with clock recovery enabled. Supports all USB and UART modes
- External Oscillator: Crystal, RC, C, or clock (1 or 2 Pin modes)
- Low Frequency (80 kHz) Internal Oscillator
- Can switch between clock sources on-the-fly

#### Packages

- 48-pin TQFP (C8051F340/1/4/5/8/C)
- 32-pin LQFP (C8051F342/3/6/7/9/A/B/D)
- 5x5 mm 32-pin QFN (C8051F342/3/6/7/9/A/B)

### Temperature Range: -40 to +85 °C





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### 1. System Overview

C8051F340/1/2/3/4/5/6/7/8/9/A/B/C/D devices are fully integrated mixed-signal System-on-a-Chip MCUs. Highlighted features are listed below. Refer to Table 1.1 for specific product feature selection.

- High-speed pipelined 8051-compatible microcontroller core (up to 48 MIPS)
- In-system, full-speed, non-intrusive debug interface (on-chip)
- Universal Serial Bus (USB) Function Controller with eight flexible endpoint pipes, integrated transceiver, and 1 kB FIFO RAM
- Supply Voltage Regulator
- True 10-bit 200 ksps differential / single-ended ADC with analog multiplexer
- On-chip Voltage Reference and Temperature Sensor
- On-chip Voltage Comparators (2)
- Precision internal calibrated 12 MHz internal oscillator and 4x clock multiplier
- Internal low-frequency oscillator for additional power savings
- Up to 64 kB of on-chip Flash memory
- Up to 4352 Bytes of on-chip RAM (256 + 4 kB)
- External Memory Interface (EMIF) available on 48-pin versions.
- SMBus/I2C, up to 2 UARTs, and Enhanced SPI serial interfaces implemented in hardware
- Four general-purpose 16-bit timers
- Programmable Counter/Timer Array (PCA) with five capture/compare modules and Watchdog Timer function
- On-chip Power-On Reset, V<sub>DD</sub> Monitor, and Missing Clock Detector
- Up to 40 Port I/O (5 V tolerant)

With on-chip Power-On Reset, V<sub>DD</sub> monitor, Voltage Regulator, Watchdog Timer, and clock oscillator, C8051F340/1/2/3/4/5/6/7/8/9/A/B/C/D devices are truly stand-alone System-on-a-Chip solutions. The Flash memory can be reprogrammed in-circuit, providing non-volatile data storage, and also allowing field upgrades of the 8051 firmware. User software has complete control of all peripherals, and may individually shut down any or all peripherals for power savings.

The on-chip Silicon Labs 2-Wire (C2) Development Interface allows non-intrusive (uses no on-chip resources), full speed, in-circuit debugging using the production MCU installed in the final application. This debug logic supports inspection and modification of memory and registers, setting breakpoints, single stepping, run and halt commands. All analog and digital peripherals are fully functional while debugging using C2. The two C2 interface pins can be shared with user functions, allowing in-system debugging without occupying package pins.

Each device is specified for 2.7–5.25 V operation over the industrial temperature range (-40 to +85 °C). For voltages above 3.6 V, the on-chip Voltage Regulator must be used. A minimum of 3.0 V is required for USB communication. The Port I/O and RST pins are tolerant of input signals up to 5 V. C8051F340/1/2/3/ 4/5/6/7/8/9/A/B/C/D devices are available in 48-pin TQFP, 32-pin LQFP, or 32-pin QFN packages. See Table 1.1, "Product Selection Guide," on page 18 for feature and package choices.



Ordering Part Number	MIPS (Peak)	Flash Memory (Bytes)	RAM	Calibrated Internal Oscillator	Low Frequency Oscillator	USB with 1k Endpoint RAM	Supply Voltage Regulator	SMBus/I2C	Enhanced SPI	UARTs	Timers (16-bit)	Programmable Counter Array	Digital Port I/Os	External Memory Interface (EMIF)	10-bit 200 ksps ADC	Temperature Sensor	Voltage Reference	Analog Comparators	Package
C8051F340-GQ	48	64k	4352	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	2	4	$\checkmark$	40	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	2	TQFP48
C8051F341-GQ	48	32k	2304	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	2	4	$\checkmark$	40	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	2	TQFP48
C8051F342-GQ	48	64k	4352	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	1	4	$\checkmark$	25	—	$\checkmark$	$\checkmark$	$\checkmark$	2	LQFP32
C8051F342-GM	48	64k	4352	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	1	4	$\checkmark$	25	—	$\checkmark$	$\checkmark$	$\checkmark$	2	QFN32
C8051F343-GQ	48	32k	2304	$\checkmark$	~	$\checkmark$	~	$\checkmark$	$\checkmark$	1	4	$\checkmark$	25	—	$\checkmark$	$\checkmark$	~	2	LQFP32
C8051F343-GM	48	32k	2304	$\checkmark$	~	$\checkmark$	~	$\checkmark$	$\checkmark$	1	4	$\checkmark$	25	_	$\checkmark$	$\checkmark$	~	2	QFN32
C8051F344-GQ	25	64k	4352	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	2	4	$\checkmark$	40	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	2	TQFP48
C8051F345-GQ	25	32k	2304	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	2	4	$\checkmark$	40	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	2	TQFP48
C8051F346-GQ	25	64k	4352	$\checkmark$	_	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	1	4	$\checkmark$	25		$\checkmark$	$\checkmark$	$\checkmark$	2	LQFP32
C8051F346-GM	25	64k	4352	$\checkmark$		$\checkmark$	~	<	$\checkmark$	1	4	$\checkmark$	25		$\checkmark$	$\checkmark$	~	2	QFN32
C8051F347-GQ	25	32k	2304	$\checkmark$	_	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	1	4	$\checkmark$	25		$\checkmark$	$\checkmark$	$\checkmark$	2	LQFP32
C8051F347-GM	25	32k	2304	$\checkmark$	_	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	1	4	$\checkmark$	25		$\checkmark$	$\checkmark$	$\checkmark$	2	QFN32
C8051F348-GQ	25	32k	2304	$\checkmark$	~	$\checkmark$	$\checkmark$	<	$\checkmark$	2	4	$\checkmark$	40	$\checkmark$				2	TQFP48
C8051F349-GQ	25	32k	2304	~	~	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	1	4	$\checkmark$	25	_	_	_	_	2	LQFP32
C8051F349-GM	25	32k	2304	$\checkmark$	~	$\checkmark$	~	<	$\checkmark$	1	4	$\checkmark$	25					2	QFN32
C8051F34A-GQ	48	64k	4352	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	2	4	$\checkmark$	25		$\checkmark$	$\checkmark$	$\checkmark$	2	LQFP32
C8051F34A-GM	48	64k	4352	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	2	4	$\checkmark$	25		$\checkmark$	$\checkmark$	$\checkmark$	2	QFN32
C8051F34B-GQ	48	32k	2304	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	2	4	$\checkmark$	25		$\checkmark$	$\checkmark$	$\checkmark$	2	LQFP32
C8051F34B-GM	48	32k	2304	~	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	2	4	$\checkmark$	25	_	$\checkmark$	$\checkmark$	$\checkmark$	2	QFN32
C8051F34C-GQ	48	64k	4352	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	2	4	$\checkmark$	40	$\checkmark$	_	_	_	2	TQFP48
C8051F34D-GQ	48	64k	4352	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	$\checkmark$	1	4	$\checkmark$	25	—	_	—	—	2	LQFP32

Table 1.1. Product Selection Guide



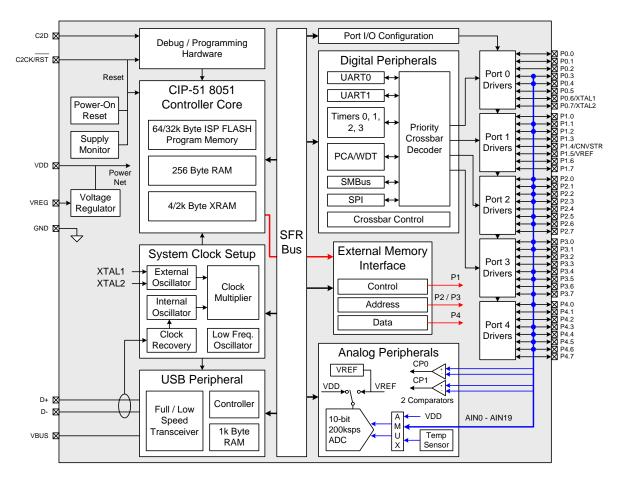


Figure 1.1. C8051F340/1/4/5 Block Diagram



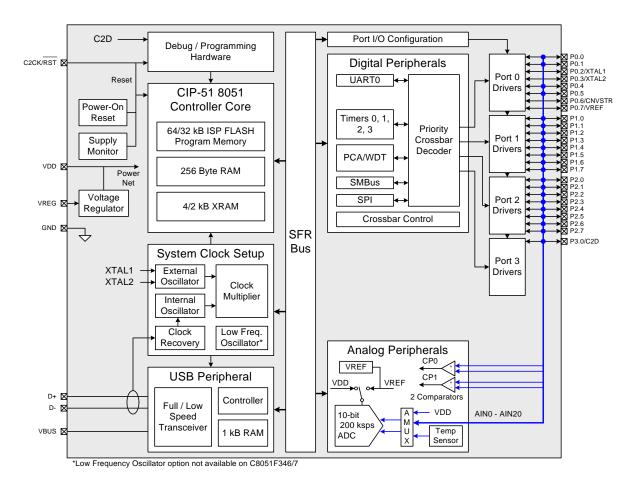


Figure 1.2. C8051F342/3/6/7 Block Diagram



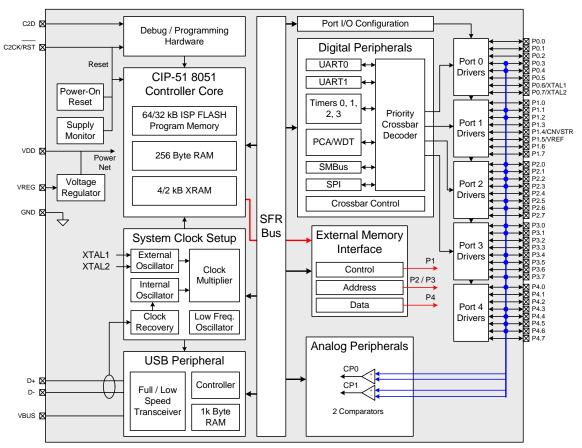


Figure 1.3. C8051F348/C Block Diagram



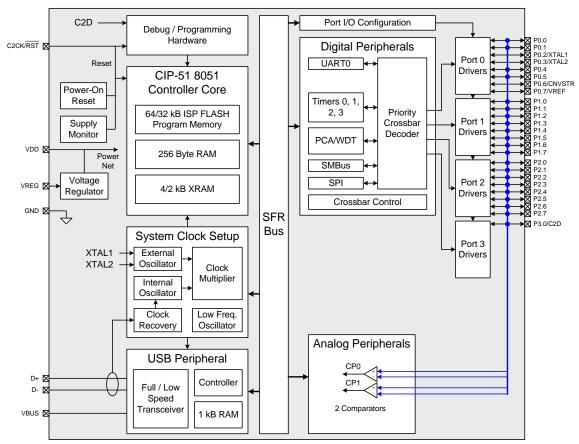


Figure 1.4. C8051F349/D Block Diagram



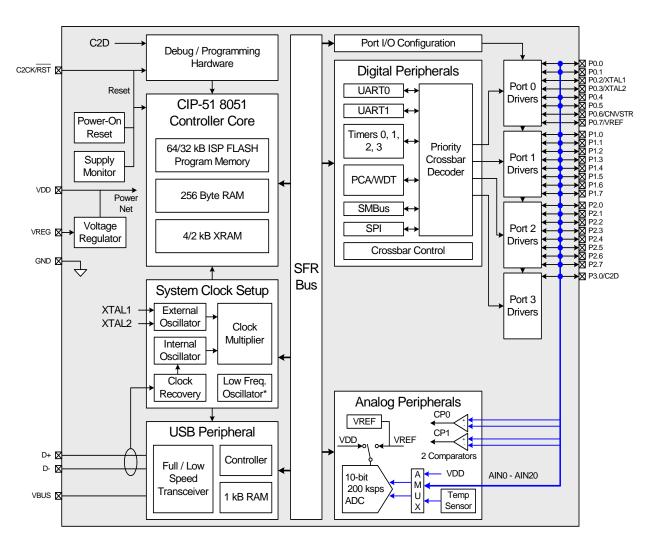


Figure 1.5. C8051F34A/B Block Diagram



## 2. Absolute Maximum Ratings

Parameter	Conditions	Min	Тур	Мах	Units
Ambient temperature under bias		-55		125	°C
Storage Temperature		-65		150	°C
Voltage on any Port I/O Pin or RST with respect to GND		-0.3		5.8	V
Voltage on $V_{DD}$ with respect to GND		-0.3		4.2	V
Maximum Total current through V <sub>DD</sub> and GND				500	mA
Maximum output current sunk by $\overline{RST}$ or any Port pin				100	mA

### Table 2.1. Absolute Maximum Ratings\*

\*Note: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the devices at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.



Rev. 1.3

## 3. Global DC Electrical Characteristics

### **Table 3.1. Global DC Electrical Characteristics**

### -40 to +85 °C, 25 MHz System Clock unless otherwise specified.

Parameter	Conditions	Min	Тур	Max	Units
Digital Supply Voltage <sup>1</sup>		VRST	3.3	3.6	V
Digital Supply RAM Data Retention Voltage			1.5		V
SYSCLK (System Clock) <sup>2</sup>	C8051F340/1/2/3/A/B/C/D C8051F344/5/6/7/8/9	0 0		48 25	MHz
Specified Operating Temperature Range		-40		+85	°C
Digital Supply Current - CPU	Active (Normal Mode, accessing I	Flash)			
I <sub>DD</sub> <sup>3</sup>	$V_{DD} = 3.3 \text{ V}, \text{ SYSCLK} = 48 \text{ MHz} \\ V_{DD} = 3.3 \text{ V}, \text{ SYSCLK} = 24 \text{ MHz} \\ V_{DD} = 3.3 \text{ V}, \text{ SYSCLK} = 1 \text{ MHz} \\ V_{DD} = 3.3 \text{ V}, \text{ SYSCLK} = 80 \text{ kHz} \\ \end{cases}$		25.9 13.9 0.69 55	28.5 15.7	mA mA mA μA
	V <sub>DD</sub> = 3.6 V, SYSCLK = 48 MHz V <sub>DD</sub> = 3.6 V, SYSCLK = 24 MHz		29.7 15.9	32.3 18	mA mA
I <sub>DD</sub> Supply Sensitivity <sup>3,4</sup>	SYSCLK = 1 MHz, relative to $V_{DD}$ = 3.3 V SYSCLK = 24 MHz, relative to $V_{DD}$ = 3.3 V		47 46		%/V %/V
I <sub>DD</sub> Frequency Sensitivity <sup>3,5</sup>	$V_{DD} = 3.3 \text{ V}, \text{ SYSCLK} \le 30 \text{ MHz},$ T = 25 °C $V_{DD} = 3.3 \text{ V}, \text{ SYSCLK} > 30 \text{ MHz},$ T = 25 °C		0.69 0.44		mA/MHz mA/MHz
	$V_{DD} = 3.6 \text{ V}, \text{ SYSCLK} \le 30 \text{ MHz},$ T = 25 °C $V_{DD} = 3.6 \text{ V}, \text{ SYSCLK} > 30 \text{ MHz},$ T = 25 °C		0.80 0.50		mA/MHz mA/MHz
Digital Supply Current - CPU	Inactive (Idle Mode, not accessing	g Flash)			
I <sub>DD</sub> <sup>3</sup>	$V_{DD} = 3.3 \text{ V}, \text{ SYSCLK} = 48 \text{ MHz} \\ V_{DD} = 3.3 \text{ V}, \text{ SYSCLK} = 24 \text{ MHz} \\ V_{DD} = 3.3 \text{ V}, \text{ SYSCLK} = 1 \text{ MHz} \\ V_{DD} = 3.3 \text{ V}, \text{ SYSCLK} = 80 \text{ kHz} \\ \end{cases}$	-	16.6 8.25 0.44 35	18.75 9.34	mA mA mA μA
	V <sub>DD</sub> = 3.6 V, SYSCLK = 48 MHz V <sub>DD</sub> = 3.6 V, SYSCLK = 24 MHz		18.6 9.26	20.9 10.5	mA mA
I <sub>DD</sub> Supply Sensitivity <sup>3,4</sup>	SYSCLK = 1 MHz, relative to $V_{DD}$ = 3.3 V SYSCLK = 24 MHz, relative to $V_{DD}$ = 3.3 V		41 39		%/V %/V



### Table 3.1. Global DC Electrical Characteristics (Continued)

### -40 to +85 °C, 25 MHz System Clock unless otherwise specified.

Parameter	Conditions	Min	Тур	Max	Units
I <sub>DD</sub> Frequency Sensitivity <sup>3,6</sup>	$V_{DD}$ = 3.3 V, SYSCLK $\leq$ 1 MHz, T = 25 °C		0.44 0.32		mA/MHz mA/MHz
	V <sub>DD</sub> = 3.3 V, SYSCLK > 1 MHz, T = 25 ⁰C		0.32		
	V <sub>DD</sub> = 3.6 V, SYSCLK <u>≤</u> 1 MHz, T = 25 ⁰C		0.49		mA/MHz
	V <sub>DD</sub> = 3.6 V, SYSCLK > 1 MHz, T = 25 ⁰C		0.36		mA/MHz
Digital Supply Current (Stop Mode, shutdown)	Oscillator not running, V <sub>DD</sub> monitor disabled		< 0.1		μA
Digital Supply Current for USB Module (USB Active Mode)	$V_{DD}$ = 3.3 V, USB Clock = 48 MHz		8.69		mA
	$V_{DD}$ = 3.6 V, USB Clock = 48 MHz		9.59		mA
Digital Supply Current for USB Module (USB Suspend Mode)	Oscillator not running V <sub>DD</sub> monitor disabled		< 0.1		μA

#### Notes:

- 1. USB Requires 3.0 V Minimum Supply Voltage.
- 2. SYSCLK must be at least 32 kHz to enable debugging.
- 3. Based on device characterization of data; Not production tested.
- 4. Active and Inactive I<sub>DD</sub> at voltages and frequencies other than those specified can be calculated using the I<sub>DD</sub> Supply Sensitivity. For example, if the V<sub>DD</sub> is 3.0 V instead of 3.3 V at 24 MHz: I<sub>DD</sub> = 13.9 mA typical at 3.3 V and SYSCLK = 24 MHz. From this, I<sub>DD</sub> = 13.9 mA + 0.46 x (3.0 V 3.3 V) = 13.76 mA at 3.0 V and SYSCLK = 24 MHz.
- 5.  $I_{DD}$  can be estimated for frequencies  $\leq$  30 MHz by multiplying the frequency of interest by the frequency sensitivity number for that range. When using these numbers to estimate  $I_{DD}$  for > 30 MHz, the estimate should be the current at 24 MHz (or 48 MHz) minus the difference in current indicated by the frequency sensitivity number. For example:  $V_{DD}$  = 3.3 V; SYSCLK = 35 MHz,  $I_{DD}$  = 13.9 mA (24 MHz 35 MHz) x 0.44 mA/MHz = 18.74 mA.
- 6. Idle  $I_{DD}$  can be estimated for frequencies  $\leq 1$  MHz by multiplying the frequency of interest by the frequency sensitivity number for that range. When using these numbers to estimate Idle  $I_{DD}$  for > 1 MHz, the estimate should be the current at 24 MHz (or 48 MHz) minus the difference in current indicated by the frequency sensitivity number. For example:  $V_{DD} = 3.3$  V; SYSCLK = 5 MHz, Idle  $I_{DD} = 8.25$  mA (24 MHz 5 MHz) x 0.32 mA/MHz = 2.17 mA.

Other electrical characteristics tables are found in the data sheet section corresponding to the associated peripherals. For more information on electrical characteristics for a specific peripheral, refer to the page indicated in Table 3.2.



### Table 3.2. Index to Electrical Characteristics Tables

Table Title	Page No.
ADC0 Electrical Characteristics	56
Voltage Reference Electrical Characteristics	58
Comparator Electrical Characteristics	68
Voltage Regulator Electrical Specifications	69
Reset Electrical Characteristics	106
Flash Electrical Characteristics	109
AC Parameters for External Memory Interface	130
Oscillator Electrical Characteristics	141
Port I/O DC Electrical Characteristics	158
USB Transceiver Electrical Characteristics	187



## 4. Pinout and Package Definitions

## Table 4.1. Pin Definitions for the C8051F340/1/2/3/4/5/6/7/8/9/A/B/C/D

Nome	Pin Numbers		Turne	Description				
Name	48-pin	32-pin	Туре	Description				
V <sub>DD</sub>	10	6	Power In	2.7–3.6 V Power Supply Voltage Input.				
			Power Out	3.3 V Voltage Regulator Output. See Section 8.				
GND	7	3		Ground.				
RST/	13	9	D I/O	Device Reset. Open-drain output of internal POR or $V_{DD}$ monitor. An external source can initiate a system reset by driving this pin low for at least 15 µs. See Section 11.				
C2CK			D I/O	Clock signal for the C2 Debug Interface.				
C2D	14	_	D I/O	Bi-directional data signal for the C2 Debug Interface.				
P3.0 / C2D	_	10	D I/O D I/O	Port 3.0. See Section 15 for a complete description of Port 3. Bi-directional data signal for the C2 Debug Interface.				
	11	7	Dowerla					
REGIN	11	7	Power In	5 V Regulator Input. This pin is the input to the on-chip volt- age regulator.				
VBUS	12	8	D In	VBUS Sense Input. This pin should be connected to the VBUS signal of a USB network. A 5 V signal on this pin indicates a USB network connection.				
D+	8	4	D I/O	USB D+.				
D-	9	5	D I/O	USB D–.				
P0.0	6	2	D I/O or A In	Port 0.0. See <b>Section 15</b> for a complete description of Port 0.				
P0.1	5	1	D I/O or A In	Port 0.1.				
P0.2	4	32	D I/O or A In	Port 0.2.				
P0.3	3	31	D I/O or A In	Port 0.3.				
P0.4	2	30	D I/O or A In	Port 0.4.				
P0.5	1	29	D I/O or A In	Port 0.5.				
P0.6	48	28	D I/O or A In	Port 0.6.				
P0.7	47	27	D I/O or A In	Port 0.7.				



### Table 4.1. Pin Definitions for the C8051F340/1/2/3/4/5/6/7/8/9/A/B/C/D (Continued)

Pin N		Pin Numbers		Description
Name	48-pin	32-pin	Туре	Description
P1.0	46	26	D I/O or A In	Port 1.0. See Section 15 for a complete description of Port 1.
P1.1	45	25	D I/O or A In	Port 1.1.
P1.2	44	24	D I/O or A In	Port 1.2.
P1.3	43	23	D I/O or A In	Port 1.3.
P1.4	42	22	D I/O or A In	Port 1.4.
P1.5	41	21	D I/O or A In	Port 1.5.
P1.6	40	20	D I/O or A In	Port 1.6.
P1.7	39	19	D I/O or A In	Port 1.7.
P2.0	38	18	D I/O or A In	Port 2.0. See <b>Section 15</b> for a complete description of Port 2.
P2.1	37	17	D I/O or A In	Port 2.1.
P2.2	36	16	D I/O or A In	Port 2.2.
P2.3	35	15	D I/O or A In	Port 2.3.
P2.4	34	14	D I/O or A In	Port 2.4.
P2.5	33	13	D I/O or A In	Port 2.5.
P2.6	32	12	D I/O or A In	Port 2.6.
P2.7	31	11	D I/O or A In	Port 2.7.
P3.0	30		D I/O or A In	Port 3.0. See <b>Section 15</b> for a complete description of Port 3.
P3.1	29	—	D I/O or A In	Port 3.1.
P3.2	28	—	D I/O or A In	Port 3.2.



### Table 4.1. Pin Definitions for the C8051F340/1/2/3/4/5/6/7/8/9/A/B/C/D (Continued)

Name	Pin Nu	mbers	Туре	Description
Name	48-pin	32-pin	Type	Description
P3.3	27	—	D I/O or A In	Port 3.3.
P3.4	26	—	D I/O or A In	Port 3.4.
P3.5	25		D I/O or A In	Port 3.5.
P3.6	24	—	D I/O or A In	Port 3.6.
P3.7	23	—	D I/O or A In	Port 3.7.
P4.0	22		D I/O or A In	Port 4.0. See <b>Section 15</b> for a complete description of Port 4.
P4.1	21	—	D I/O or A In	Port 4.1.
P4.2	20		D I/O or A In	Port 4.2.
P4.3	19		D I/O or A In	Port 4.3.
P4.4	18		D I/O or A In	Port 4.4.
P4.5	17	—	D I/O or A In	Port 4.5.
P4.6	16		D I/O or A In	Port 4.6.
P4.7	15	—	D I/O or A In	Port 4.7.



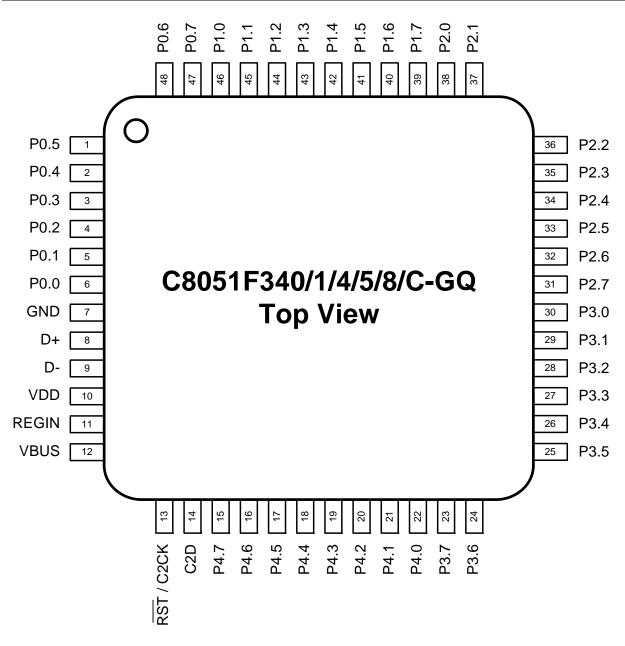
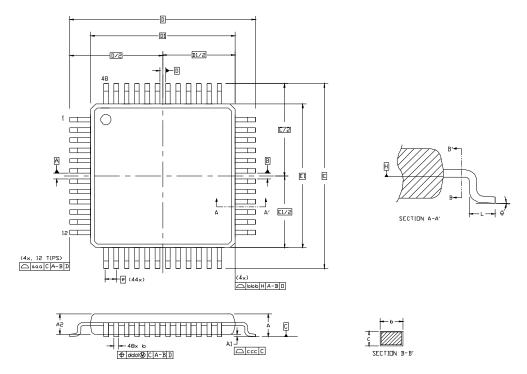


Figure 4.1. TQFP-48 Pinout Diagram (Top View)





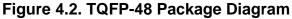


Table 4.2. TQFF-40 Fackage Dimensions								
Min	Nom	Max						
—	—	1.20						
0.05	—	0.15						
0.95	1.00	1.05						
0.17	0.22	0.27						
0.09	—	0.20						
	9.00 BSC							
7.00 BSC								
	0.50 BSC							
	9.00 BSC							
	7.00 BSC							
0.45	0.60	0.75						
	0.20							
0.20								
0.08								
0.08								
0° 3.5° 7°								
	Min — 0.05 0.95 0.17 0.09 — 0.45	Min         Nom               0.05            0.95         1.00           0.17         0.22           0.09            9.00 BSC         7.00 BSC           0.50 BSC         9.00 BSC           0.45         0.60           0.20         0.20           0.08         0.08						

### Table 4.2. TQFP-48 Package Dimensions

### Notes:

1. All dimensions shown are in millimeters (mm) unless otherwise noted.

2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.

3. This drawing conforms to JEDEC outline MS-026, variation ABC.

**4.** The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.



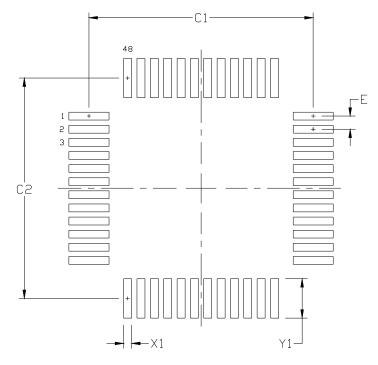


Figure 4.3. TQFP-48 Recommended PCB Land Pattern

	Dimension	Min	Max					
	C1	8.30	8.40					
	C2	8.30	8.40					
	E	0.50	BSC					
	X1	0.20	0.30					
	Y1	1.40	1.50					
Notes	:							
Gener	ral:							
1.	All dimensions shown are	in millimeters (mm) unless	otherwise noted.					
2.	This Land Pattern Design	is based on the IPC-7351	guidelines.					
	r Mask Design:		0					
3.	All metal pads are to be no	on-solder mask defined (N	SMD). Clearance between					
	•	netal pad is to be 60 µm mi						
	the pad.							
Stenc	il Design:							
	0	t and electro-polished sten	cil with trapezoidal walls					
	<ol> <li>A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.</li> </ol>							
5.	5. The stencil thickness should be 0.125 mm (5 mils).							
	<ol> <li>The ratio of stencil aperture to land pad size should be 1:1 for all pads.</li> </ol>							
Card Assembly:								
	-	r paste is recommended						
	A No-Clean, Type-3 solder paste is recommended.							

### Table 4.3. TQFP-48 PCB Land Pattern Dimensions

8. The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.



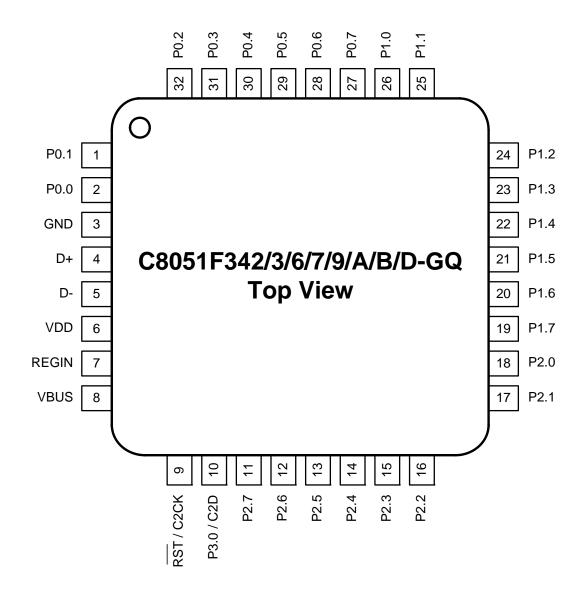


Figure 4.4. LQFP-32 Pinout Diagram (Top View)



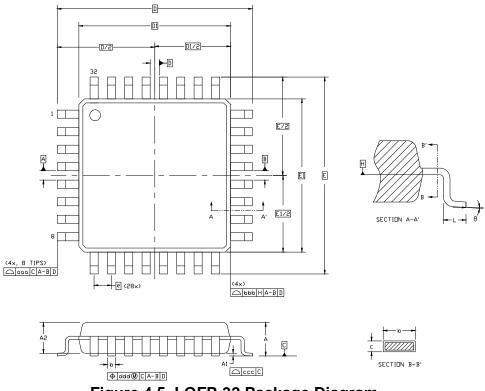


Figure 4.5. LQFP-32 Package Diagram

Table 4.4. LQFP-32 Package Dimensions				
Dimension	Min	Nom	Max	
A	—	—	1.60	
A1	0.05	—	0.15	
A2	1.35	1.40	1.45	
b	0.30	0.37	0.45	
С	0.09	—	0.20	
D	9.00 BSC			
D1	7.00 BSC			
е	0.80 BSC			
E	9.00 BSC			
E1	7.00 BSC			
L	0.45	0.60	0.75	
aaa	0.20			
bbb	0.20			
CCC	0.10			
ddd	0.20			
θ	0°	3.5°	7°	

### Table 4.4. LQFP-32 Package Dimensions

#### Notes:

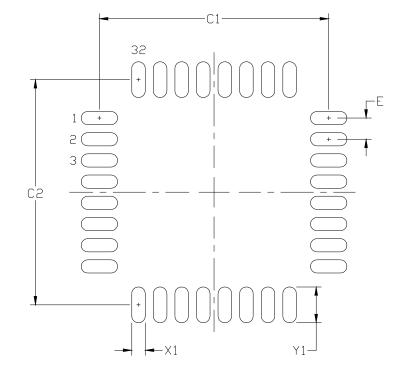
1. All dimensions shown are in millimeters (mm) unless otherwise noted.

2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.

3. This drawing conforms to JEDEC outline MS-026, variation BBA.

4. The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.





### Figure 4.6. LQFP-32 Recommended PCB Land Pattern

### Table 4.5. LQFP-32 PCB Land Pattern Dimensions

	Dimension	Min	Max		
	C1	8.40	8.50		
	C2	8.40	8.50		
	E	0.80 BSC			
	X1	0.40	0.50		
	Y1	1.25	1.35		
Notes	:		•		
Gener	al:				
1. All dimensions shown are in millimeters (mm) unless otherwise noted.					
2. This Land Pattern Design is based on the IPC-7351 guidelines.					
Solde	r Mask Design:				
3.	3. All metal pads are to be non-solder mask defined (NSMD). Clearance between				
	the solder mask and the m	ietal pad is to be 60 μm mii	nimum, all the way around		
	the pad.				
Stenc	il Design:				
4.	A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls				
	should be used to assure good solder paste release.				
5.	The stencil thickness should be 0.125 mm (5 mils).				
6.	. The ratio of stencil aperture to land pad size should be 1:1 for all pads.				
	Assembly:	-	•		
7.	A No-Clean, Type-3 solder paste is recommended.				
	The recommended card reflow profile is per the JEDEC/IPC J-STD-020				
-	specification for Small Boo	• •			
	•	· ·			



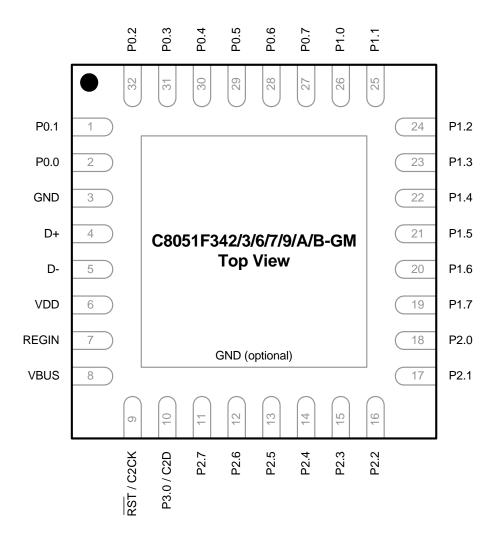
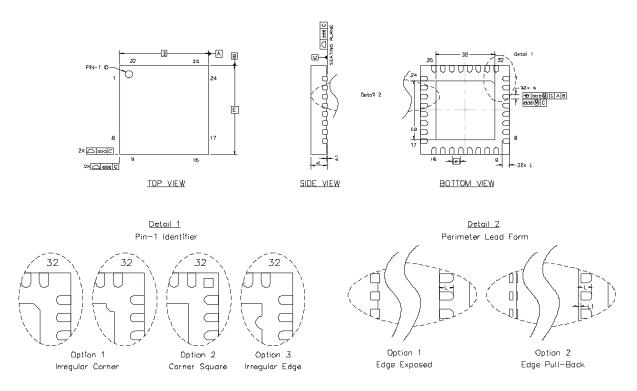


Figure 4.7. QFN-32 Pinout Diagram (Top View)





#### Figure 4.8. QFN-32 Package Drawing

		•			
Dimension	Min	Nom	Max		
А	0.80	0.9	1.00		
A1	0.00	0.02	0.05		
b	0.18	0.25	0.30		
D	5.00 BSC				
D2	3.20	3.30	3.40		
е		0.50 BSC			
E		5.00 BSC			
E2	3.20 3.30 3.40				
L	0.30 0.40 0.50				
Notes:		•			

### Table 4.6. QFN-32 Package Dimensions

- 1. All dimensions shown are in millimeters (mm) unless otherwise noted.
- 2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.
- 3. This drawing conforms to the JEDEC Solid State Outline MO-220, variation VHHD except for custom features D2, E2, and L which are toleranced per supplier designation.
- 4. Recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.



Dimension	Min	Nom	Мах				
L1	0.00	_	0.15				
aaa	—	_	0.15				
bbb	—	_	0.10				
ddd	—	_	0.05				
eee	—	—	0.08				
Notes: 1. All dimensions shown are in millimeters (mm) unless otherwise noted.							

#### Table 4.6. QFN-32 Package Dimensions (Continued)

All dimensions shown are in minimeters (infit) dimess otherwise
 Dimensioning and Tolerancing per ANSI Y14.5M-1994.

- **3.** This drawing conforms to the JEDEC Solid State Outline MO-220, variation VHHD except for custom features D2, E2, and L which are toleranced per supplier designation.
- **4.** Recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.



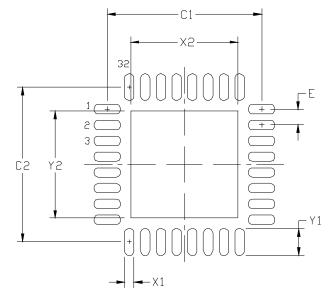


Figure 4.9. QFN-32 Recommended PCB Land Pattern

Dimension	Min	Max			
C1	4.80	4.90			
C2	4.80	4.90			
E	0.50 BSC				
X1	0.20	0.30			

Dimension	Min	Max
X2	3.20	3.40
Y1	0.75	0.85
Y2	3.20	3.40

#### Notes:

#### General:

- 1. All dimensions shown are in millimeters (mm) unless otherwise noted.
- 2. This Land Pattern Design is based on the IPC-7351 guidelines.

#### Solder Mask Design:

 All metal pads are to be non-solder mask defined (NSMD). Clearance between the solder mask and the metal pad is to be 60µm minimum, all the way around the pad.

#### Stencil Design:

- **4.** A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.
- 5. The stencil thickness should be 0.125 mm (5 mils).
- 6. The ratio of stencil aperture to land pad size should be 1:1 for all perimeter pins.
- **7.** A 3x3 array of 1.0 mm openings on a 1.2mm pitch should be used for the center pad to assure the proper paste volume.

#### Card Assembly:

- 8. A No-Clean, Type-3 solder paste is recommended.
- **9.** The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.



## 5. 10-Bit ADC (ADC0, C8051F340/1/2/3/4/5/6/7/A/B Only)

The ADC0 subsystem for the C8051F34x devices consists of two analog multiplexers (referred to collectively as AMUX0), and a 200 ksps, 10-bit successive-approximation-register ADC with integrated track-and-hold and programmable window detector. The AMUX0, data conversion modes, and window detector are all configured under software control via the Special Function Registers shown in Figure 5.1. ADC0 operates in both Single-ended and Differential modes, and may be configured to measure voltages at port pins, the Temperature Sensor output, or  $V_{DD}$  with respect to a port pin, VREF, or GND. The connection options for AMUX0 are detailed in SFR Definition 5.1 and SFR Definition 5.2. The ADC0 subsystem is enabled only when the AD0EN bit in the ADC0 Control register (ADC0CN) is set to logic 1. The ADC0 subsystem is in low power shutdown when this bit is logic 0.

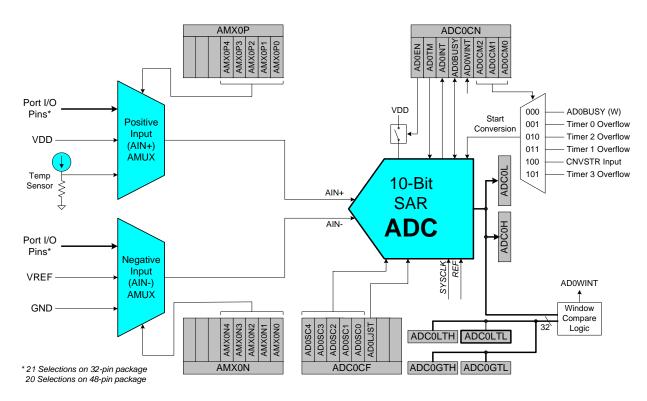


Figure 5.1. ADC0 Functional Block Diagram



### 5.1. Analog Multiplexer

AMUX0 selects the positive and negative inputs to the ADC. The positive input (AIN+) can be connected to individual Port pins, the on-chip temperature sensor, or the positive power supply ( $V_{DD}$ ). The negative input (AIN-) can be connected to individual Port pins, VREF, or GND. When GND is selected as the negative input, ADC0 operates in Single-ended Mode; at all other times, ADC0 operates in Differential Mode. The ADC0 input channels are selected in the AMX0P and AMX0N registers as described in SFR Definition 5.1 and SFR Definition 5.2.

The conversion code format differs between Single-ended and Differential modes. The registers ADC0H and ADC0L contain the high and low bytes of the output conversion code from the ADC at the completion of each conversion. Data can be right-justified or left-justified, depending on the setting of the AD0LJST bit (ADC0CN.0). When in Single-ended Mode, conversion codes are represented as 10-bit unsigned integers. Inputs are measured from '0' to VREF x 1023/1024. Example codes are shown below for both right-justified and left-justified data. Unused bits in the ADC0H and ADC0L registers are set to '0'.

Input Voltage (Single-Ended)	Right-Justified ADC0H:ADC0L (AD0LJST = 0)	Left-Justified ADC0H:ADC0L (AD0LJST = 1)
VREF x 1023/1024	0x03FF	0xFFC0
VREF x 512/1024	0x0200	0x8000
VREF x 256/1024	0x0100	0x4000
0	0x0000	0x0000

When in Differential Mode, conversion codes are represented as 10-bit signed 2's complement numbers. Inputs are measured from –VREF to VREF x 511/512. Example codes are shown below for both right-justified and left-justified data. For right-justified data, the unused MSBs of ADC0H are a sign-extension of the data word. For left-justified data, the unused LSBs in the ADC0L register are set to '0'.

Input Voltage (Differential)	Right-Justified ADC0H:ADC0L (AD0LJST = 0)	Left-Justified ADC0H:ADC0L (AD0LJST = 1)			
VREF x 511/512	0x01FF	0x7FC0			
VREF x 256/512	0x0100	0x4000			
0	0x0000	0x0000			
–VREF x 256/512	0xFF00	0xC000			
-VREF	0xFE00	0x8000			

**Important Note About ADC0 Input Configuration:** Port pins selected as ADC0 inputs should be configured as analog inputs, and should be skipped by the Digital Crossbar. To configure a Port pin for analog input, set to '0' the corresponding bit in register PnMDIN (for n = 0,1,2,3). To force the Crossbar to skip a Port pin, set to '1' the corresponding bit in register PnSKIP (for n = 0,1,2). See Section "15. Port Input/ Output" on page 142 for more Port I/O configuration details.



#### 5.2. Temperature Sensor

The temperature sensor transfer function is shown in Figure 5.2. The output voltage ( $V_{TEMP}$ ) is the positive ADC input when the temperature sensor is selected by bits AMX0P4-0 in register AMX0P. Values for the Offset and Slope parameters can be found in Table 5.1.

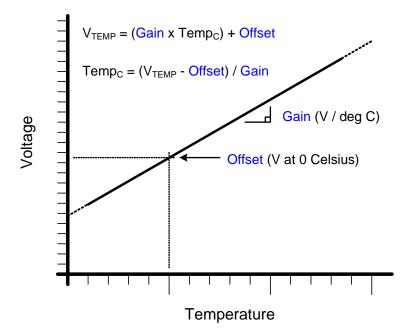


Figure 5.2. Temperature Sensor Transfer Function

The uncalibrated temperature sensor output is extremely linear and suitable for relative temperature measurements (see Table 5.1 for linearity specifications). For absolute temperature measurements, offset and/ or gain calibration is recommended. Typically a 1-point (offset) calibration includes the following steps:

- Step 1. Control/measure the ambient temperature (this temperature must be known).
- Step 2. Power the device, and delay for a few seconds to allow for self-heating.
- Step 3. Perform an ADC conversion with the temperature sensor selected as the positive input and GND selected as the negative input.
- Step 4. Calculate the offset characteristics, and store this value in non-volatile memory for use with subsequent temperature sensor measurements.

Figure 5.3 shows the typical temperature sensor error assuming a 1-point calibration at 25 °C. Note that parameters which affect ADC measurement, in particular the voltage reference value, will also affect temperature measurement.



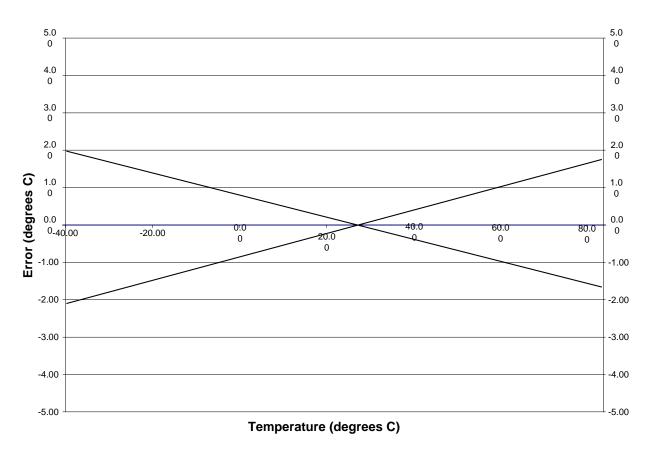


Figure 5.3. Temperature Sensor Error with 1-Point Calibration (VREF = 2.40 V)



#### 5.3. Modes of Operation

ADC0 has a maximum conversion speed of 200 ksps. The ADC0 conversion clock is a divided version of the system clock, determined by the AD0SC bits in the ADC0CF register (system clock divided by (AD0SC + 1) for  $0 \le AD0SC \le 31$ ).

#### 5.3.1. Starting a Conversion

A conversion can be initiated in one of five ways, depending on the programmed states of the ADC0 Start of Conversion Mode bits (AD0CM2–0) in register ADC0CN. Conversions may be initiated by one of the following:

- 1. Writing a '1' to the AD0BUSY bit of register ADC0CN
- 2. A Timer 0 overflow (i.e., timed continuous conversions)
- 3. A Timer 2 overflow
- 4. A Timer 1 overflow
- 5. A rising edge on the CNVSTR input signal
- 6. A Timer 3 overflow

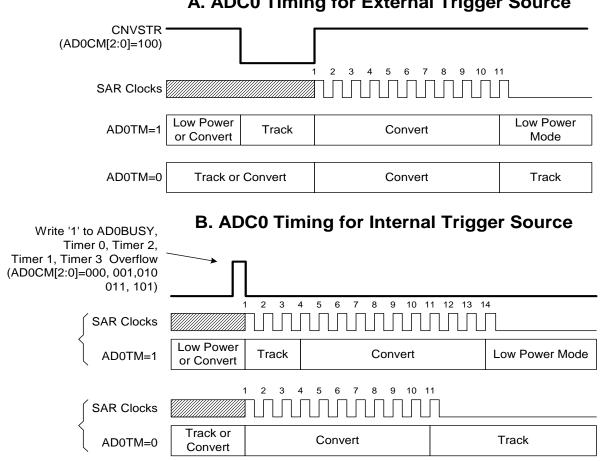
Writing a '1' to AD0BUSY provides software control of ADC0 whereby conversions are performed "on-demand". During conversion, the AD0BUSY bit is set to logic 1 and reset to logic 0 when the conversion is complete. The falling edge of AD0BUSY triggers an interrupt (when enabled) and sets the ADC0 interrupt flag (AD0INT). Note: When polling for ADC conversion completions, the ADC0 interrupt flag (AD0INT) should be used. Converted data is available in the ADC0 data registers, ADC0H:ADC0L, when bit AD0INT is logic 1. Note that when Timer 2 or Timer 3 overflows are used as the conversion source, Low Byte overflows are used if Timer 2/3 is in 8-bit mode; High byte overflows are used if Timer 2/3 is in 16-bit mode. See Section "21. Timers" on page 235 for timer configuration.

**Important Note About Using CNVSTR:** The CNVSTR input pin also functions as a Port pin. When the CNVSTR input is used as the ADC0 conversion source, the associated Port pin should be skipped by the Digital Crossbar. To configure the Crossbar to skip a pin, set the corresponding bit in the PnSKIP register to '1'. See Section "15. Port Input/Output" on page 142 for details on Port I/O configuration.



#### 5.3.2. Tracking Modes

The AD0TM bit in register ADC0CN controls the ADC0 track-and-hold mode. In its default state, the ADC0 input is continuously tracked, except when a conversion is in progress. When the AD0TM bit is logic 1, ADC0 operates in low-power track-and-hold mode. In this mode, each conversion is preceded by a tracking period of 3 SAR clocks (after the start-of-conversion signal). When the CNVSTR signal is used to initiate conversions in low-power tracking mode, ADC0 tracks only when CNVSTR is low; conversion begins on the rising edge of CNVSTR (see Figure 5.4). Tracking can also be disabled (shutdown) when the device is in low power standby or sleep modes. Low-power track-and-hold mode is also useful when AMUX settings are frequently changed, due to the settling time requirements described in Section "5.3.3. Settling Time Requirements" on page 47.



A. ADC0 Timing for External Trigger Source

Figure 5.4. 10-Bit ADC Track and Conversion Example Timing



#### 5.3.3. Settling Time Requirements

When the ADC0 input configuration is changed (i.e., a different AMUX0 selection is made), a minimum tracking time is required before an accurate conversion can be performed. This tracking time is determined by the AMUX0 resistance, the ADC0 sampling capacitance, any external source resistance, and the accuracy required for the conversion. Note that in low-power tracking mode, three SAR clocks are used for tracking at the start of every conversion. For most applications, these three SAR clocks will meet the minimum tracking time requirements.

Figure 5.5 shows the equivalent ADC0 input circuits for both Differential and Single-ended modes. Notice that the equivalent time constant for both input circuits is the same. The required ADC0 settling time for a given settling accuracy (SA) may be approximated by Equation 5.1. When measuring the Temperature Sensor output or  $V_{DD}$  with respect to GND,  $R_{TOTAL}$  reduces to  $R_{MUX}$ . See Table 5.1 for ADC0 minimum settling time requirements.

$$t = \ln\left(\frac{2^n}{SA}\right) \times R_{TOTAL} C_{SAMPLE}$$

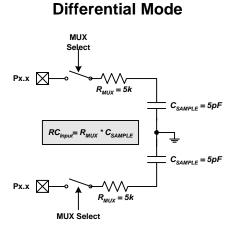
### Equation 5.1. ADC0 Settling Time Requirements

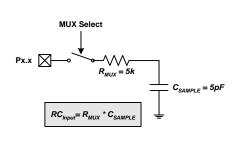
Where:

SA is the settling accuracy, given as a fraction of an LSB (for example, 0.25 to settle within 1/4 LSB) *t* is the required settling time in seconds

 $R_{TOTAL}$  is the sum of the AMUX0 resistance and any external source resistance.

*n* is the ADC resolution in bits (10).





Single-Ended Mode

### Figure 5.5. ADC0 Equivalent Input Circuits



#### SFR Definition 5.1. AMX0P: AMUX0 Positive Channel Select

R _	R	R	R/W AMX0P4	AMX0P3	AMX0P2	AMX0P1	R/W AMX0P0	0000000	
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address	
Biti	Bito	Bito	BRI	Bito	BILL	Bitt	Bito	0xBB	
								••••	
Bits7–5:	UNUSED. Re	ead = 000	b; Write = do	on't care.					
Bits4–0:	AMX0P4-0:	AMUX0 P	ositive Input	Selection					
		<b>B</b> / A		<b>B</b> <i>W</i> <b>I</b>					
	AMX0	P4-0		Positive Ir			sitive Input		
	000	00	(32-	pin Packag P1.0	e)	• •	Package) 2.0		
	000			P1.0 P1.1			2.0		
	000			P1.1 P1.2			2.1		
	000			P1.2 P1.3			2.2		
	000			P1.4 P1.5 P1.6 P1.7 P2.0 P2.1 P2.2 P2.3 P2.4 P2.5 P2.6			P2.5 P2.6 P3.0 P3.1 P3.4 P3.5 P3.7 P4.0 P4.3 P4.4 P4.5		
	001								
	001	-							
	001								
	010	00							
	010	01							
	010	10							
	010	11							
	011								
	011								
	011								
	011			P2.7			4.6		
	100			P3.0			RVED		
	100			P0.0			0.3		
	100			P0.1			0.4		
	100		_	P0.4			1.1		
	101			P0.5			1.2		
	- 10101 - 111			ESERVED			RVED		
			IE	mp Sensor			Sensor		
	111	11	V <sub>DD</sub>			V <sub>DD</sub>			



#### SFR Definition 5.2. AMX0N: AMUX0 Negative Channel Select

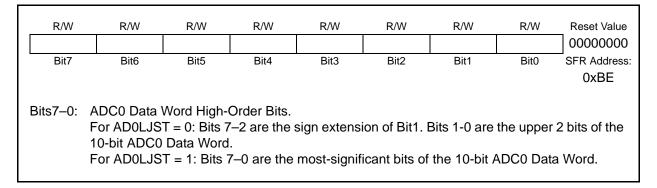
R	R	R	R/W	R/W	R/W	R/W	R/W	Reset Value	
-	-	-	AMX0N4	AMX0N3	AMX0N2	AMX0N1	AMX0N0	00000000	
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address	
								0xBA	
			1 147.36 1.						
	UNUSED. Re								
DIIS4-0.	AMX0N4-0: Note that whe		•			DC0 operat	os in Sinalo	-ondod	
	mode. For all								
			gaille input	00100110110,					
	AMX0	N4-0	ADC0	Negative I	nput	ADC0 Neg	gative Inpu	t	
				pin Packag			Package)		
	000	00		P1.0		P	2.0		
	000	01		P1.1		P	2.1		
	000	10		P1.2		P	2.2		
	000	11		P1.3			P2.3		
	001	00		P1.4			P2.5		
	001			P1.5			P2.6		
	001			P1.6			P3.0		
	001			P1.7			P3.1		
	010			P2.0			P3.4		
	010			P2.1			P3.5		
	010	-		P2.2			3.7		
	010			P2.3			P4.0		
	011			P2.4			P4.3		
	011			P2.5			4.4		
	011			P2.6			P4.5		
	011			P2.7			P4.6		
	100			P3.0			RESERVED		
	100			P0.0			P0.3		
	100			P0.1			P0.4		
	100			P0.4			P1.1		
	101 10101 -			P0.5 RESERVED			P1.2 RESERVED		
			R R				REF		
	111 111				Mode) (			do)	
	1 111	11		GND (Single-Ended Mode)			GND (Single-Ended Mode)		



#### SFR Definition 5.3. ADC0CF: ADC0 Configuration

R/W	R/W AD0SC3	R/W AD0SC2	R/W AD0SC1	R/W AD0SC0	R/W AD0LJST	R/W -	R/W -	Reset Value 11111000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xBC
Bits7–3: AD0SC4–0: ADC0 SAR Conversion Clock Period Bits. SAR Conversion clock is derived from system clock by the following equation, where AD0SC refers to the 5-bit value held in bits AD0SC4-0. SAR Conversion clock requirements are given in Table 5.1. $AD0SC = \frac{SYSCLK}{CLK_{SAR}} - 1$								
Bit2: Bits1–0:	AD0LJST: A 0: Data in AI 1: Data in AI UNUSED. R	DC0H:ADC	OL registers	s are right-ju s are left-jus				

#### SFR Definition 5.4. ADC0H: ADC0 Data Word MSB



### SFR Definition 5.5. ADC0L: ADC0 Data Word LSB

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xBD
	ADC0 Data V For AD0LJST For AD0LJST read '0'.	Γ = 0: Bits 7	7-0 are the					) will always



#### SFR Definition 5.6. ADC0CN: ADC0 Control

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
ADOEN	-	ADOINT			AD0CM2	AD0CM1	AD0CM0	-
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address
Biti	Bito	Dito	BRT	Dito	BILL		t addressable	
						(		,
Bit7:	AD0EN: AD	C0 Enable	Bit.					
	0: ADC0 Dis	abled. ADC	CO is in low-	power shute	down.			
	1: ADC0 En			and ready fo	or data conv	versions.		
Bit6:	AD0TM: AD							
	0: Normal Tr		When ADC	0 is enable	d, tracking i	s continuou	s unless a	conversion
	is in progres		de. Treeldin			O hito (ana	h a law)	
Bit5:	1: Low-powe AD0INT: AD			-	•	-0 bits (see	below).	
Dito.	0: ADC0 has					ast time AD	0INT was o	leared
	1: ADC0 has	•						
Bit4:	AD0BUSY: /							
	Read:	,						
	0: ADC0 cor	nversion is	complete o	r a conversio	on is not cu	rrently in pro	ogress. AD	0INT is set
	to logic 1 on	-	-					
	1: ADC0 cor	nversion is	in progress					
	Write:							
	0: No Effect. 1: Initiates A		orgion if AD	0 $M$ $2$ $0$ $ 0$	006			
Bit3:	ADOWINT: A							
Dito.	0: ADC0 Wi		•	•	-	ed since this	s flag was b	ast cleared
	1: ADC0 Wi						inag nao n	
Bits2–0:	AD0CM2-0:							
	When AD0T	M = 0:						
	000: ADC0 co					Y.		
	001: ADC0 co							
	010: ADC0 cc 011: ADC0 cc							
	100: ADC0 cc					TR.		
	101: ADC0 co							
	11x: Reserved							
	When AD0T							
	000: Tracking							
	001: Tracking 010: Tracking							
	011: Tracking							
	100: ADC0 tra							
	101: Tracking		overflow of T	imer 3 and la	ists 3 SAR cl	ocks, followe	ed by conver	sion.
	11x: Reserved	d.						

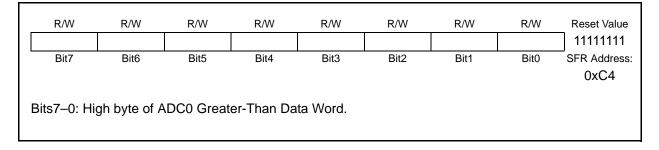


### 5.4. Programmable Window Detector

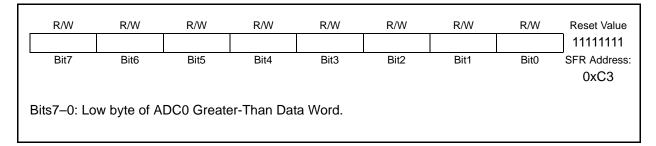
The ADC Programmable Window Detector continuously compares the ADC0 conversion results to user-programmed limits, and notifies the system when a desired condition is detected. This is especially effective in an interrupt-driven system, saving code space and CPU bandwidth while delivering faster system response times. The window detector interrupt flag (AD0WINT in register ADC0CN) can also be used in polled mode. The ADC0 Greater-Than (ADC0GTH, ADC0GTL) and Less-Than (ADC0LTH, ADC0LTL) registers hold the comparison values. The window detector flag can be programmed to indicate when measured data is inside or outside of the user-programmed limits, depending on the contents of the ADC0 Less-Than and ADC0 Greater-Than registers.

The Window Detector registers must be written with the same format (left/right justified, signed/unsigned) as that of the current ADC configuration (left/right justified, single-ended/differential).

### SFR Definition 5.7. ADC0GTH: ADC0 Greater-Than Data High Byte

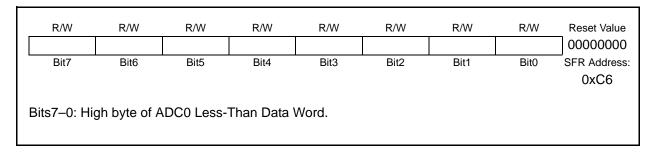


## SFR Definition 5.8. ADC0GTL: ADC0 Greater-Than Data Low Byte

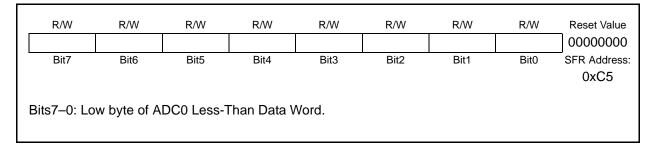




### SFR Definition 5.9. ADC0LTH: ADC0 Less-Than Data High Byte



### SFR Definition 5.10. ADC0LTL: ADC0 Less-Than Data Low Byte





#### 5.4.1. Window Detector In Single-Ended Mode

Figure 5.6 shows two example window comparisons for right-justified, single-ended data, with ADC0LTH:ADC0LTL = 0x0080 (128d) and ADC0GTH:ADC0GTL = 0x0040 (64d). In single-ended mode, the input voltage can range from '0' to VREF x (1023/1024) with respect to GND, and is represented by a 10-bit unsigned integer value. In the left example, an AD0WINT interrupt will be generated if the ADC0 conversion word (ADC0H:ADC0L) is within the range defined by ADC0GTH:ADC0GTL and ADC0LTH:ADC0LTL (if 0x0040 < ADC0H:ADC0L < 0x0080). In the right example, and AD0WINT interrupt will be generated if the ADC0 conversion word is outside of the range defined by the ADC0GT and ADC0LT registers (if ADC0H:ADC0L < 0x0040 or ADC0H:ADC0L > 0x0080). Figure 5.7 shows an example using left-justified data with equivalent ADC0GT and ADC0LT register settings.

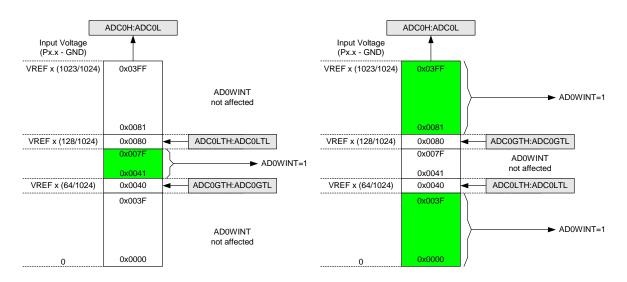


Figure 5.6. ADC Window Compare Example: Right-Justified Single-Ended Data

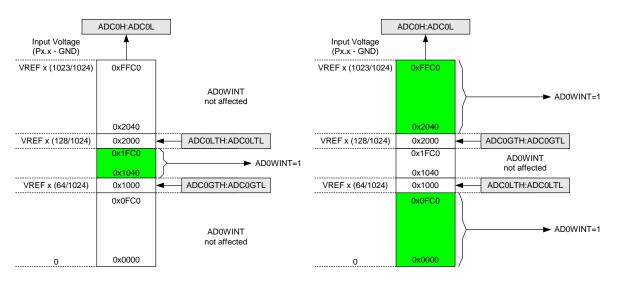
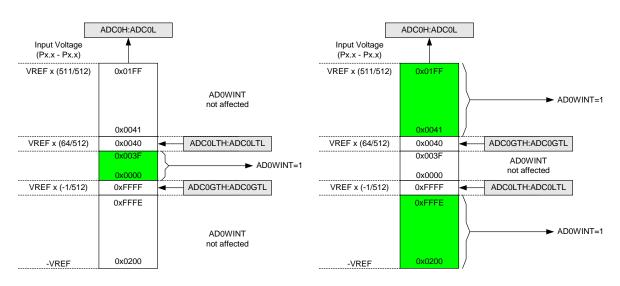


Figure 5.7. ADC Window Compare Example: Left-Justified Single-Ended Data

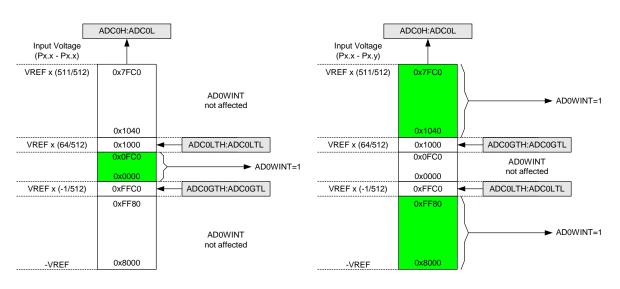


#### 5.4.2. Window Detector In Differential Mode

Figure 5.8 shows two example window comparisons for right-justified, differential data, with ADC0LTH:ADC0LTL = 0x0040 (+64d) and ADC0GTH:ADC0GTH = 0xFFFF (-1d). In differential mode, the measurable voltage between the input pins is between -VREF and VREF\*(511/512). Output codes are represented as 10-bit 2's complement signed integers. In the left example, an AD0WINT interrupt will be generated if the ADC0 conversion word (ADC0H:ADC0L) is within the range defined by ADC0GTH:ADC0GTL and ADC0LTH:ADC0LTL (if 0xFFFF (-1d) < ADC0H:ADC0L < 0x0040 (64d)). In the right example, an AD0WINT interrupt will be generated if the ADC0 conversion word is outside of the range defined by the ADC0GT and ADC0LT registers (if ADC0H:ADC0L < 0xFFFF (-1d) or ADC0H:ADC0L > 0x0040 (+64d)). Figure 5.9 shows an example using left-justified data with equivalent ADC0GT and ADC0LT register settings.







### Figure 5.9. ADC Window Compare Example: Left-Justified Differential Data



### Table 5.1. ADC0 Electrical Characteristics

#### $V_{DD}$ = 3.0 V, VREF = 2.40 V, -40 to +85 °C unless otherwise specified

Parameter	Conditions	Min	Тур	Мах	Units
	DC Accuracy	1			
Resolution			10		bits
Integral Nonlinearity			±0.5	±1	LSB
Differential Nonlinearity	Guaranteed Monotonic		±0.5	±1	LSB
Offset Error		-15	0	+15	LSB
Full Scale Error		-15	-1	+15	LSB
Offset Temperature Coefficient			10		ppm/°C
Dynamic Performance (10 kHz	z sine-wave Single-ended inpu	ut, 1 dB be	elow Full	Scale, 2	00 ksps)
Signal-to-Noise Plus Distortion		51	52.5		dB
Total Harmonic Distortion	Up to the 5 <sup>th</sup> harmonic		-67		dB
Spurious-Free Dynamic Range			78		dB
	Conversion Rate				
SAR Conversion Clock				3	MHz
Conversion Time in SAR Clocks		10			clocks
Track/Hold Acquisition Time		300			ns
Throughput Rate				200	ksps
	Analog Inputs				
ADC Input Voltage Range	Single Ended (AIN+ – GND) Differential (AIN+ – AIN–)	0 –VREF		VREF VREF	V V
Absolute Pin Voltage with respect to GND	Single Ended or Differential	0		$V_{DD}$	V
Input Capacitance			5		pF
	Temperature Sensor				
Linearity <sup>1</sup>			±0.1		°C
Gain			2.86		mV/°C
Gain Error <sup>2</sup>			±33.5		µV/⁰C
Offset <sup>1</sup>	(Temp = 0 °C)		776		mV
Offset Error <sup>2</sup>			±8.51		mV
	Power Specifications				
Power Supply Current (V <sub>DD</sub> sup- plied to ADC0)	Operating Mode, 200 ksps		400	900	μA
Power Supply Rejection			±0.3		mV/V

Notes:

1. Includes ADC offset, gain, and linearity variations.

2. Represents one standard deviation from the mean.



## 6. Voltage Reference (C8051F340/1/2/3/4/5/6/7/A/B Only)

The Voltage reference MUX on C8051F34x devices is configurable to use an externally connected voltage reference, the on-chip reference voltage generator, or the power supply voltage  $V_{DD}$  (see Figure 6.1). The REFSL bit in the Reference Control register (REF0CN) selects the reference source. For the internal reference or an external source, REFSL should be set to '0'; For  $V_{DD}$  as the reference source, REFSL should be set to '1'.

The BIASE bit enables the internal ADC bias generator, which is used by the ADC and Internal Oscillator. This enable is forced to logic 1 when either of the aforementioned peripherals is enabled. The ADC bias generator may be enabled manually by writing a '1' to the BIASE bit in register REFOCN; see SFR Definition 6.1 for REFOCN register details. The Reference bias generator (see Figure 6.1) is used by the Internal Voltage Reference, Temperature Sensor, and Clock Multiplier. The Reference bias is automatically enabled when any of the aforementioned peripherals are enabled. The electrical specifications for the voltage reference and bias circuits are given in Table 6.1.

**Important Note About the VREF Pin:** The VREF pin, when not using the on-chip voltage reference or an external precision reference, can be configured as a GPIO Port pin. When using an external voltage reference or the on-chip reference, the VREF pin should be configured as analog pin and skipped by the Digital Crossbar. To configure the VREF pin for analog mode, set the corresponding bit in the PnMDIN register to '0'. To configure the Crossbar to skip the VREF pin, set the corresponding bit in register PnSKIP to '1'. Refer to Section "15. Port Input/Output" on page 142 for complete Port I/O configuration details.

The temperature sensor connects to the ADC0 positive input multiplexer (see Section "5.1. Analog Multiplexer" on page 42 for details). The TEMPE bit in register REF0CN enables/disables the temperature sensor. While disabled, the temperature sensor defaults to a high impedance state and any ADC0 measurements performed on the sensor result in meaningless data.

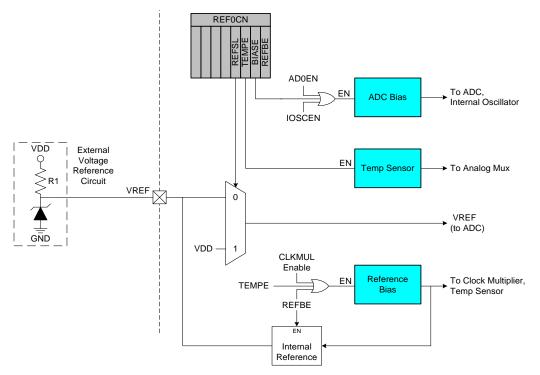


Figure 6.1. Voltage Reference Functional Block Diagram



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value		
-	-	-	-	REFSL	TEMPE	BIASE	REFBE	00000000		
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:		
								0xD1		
Bit3:	Bits7–3: UNUSED. Read = 00000b; Write = don't care. Bit3: REFSL: Voltage Reference Select. This bit selects the source for the internal voltage reference. 0: VREF pin used as voltage reference. 1: V <sub>DD</sub> used as voltage reference.									
Bit2:	TEMPE: Ten 0: Internal Te 1: Internal Te	emperature	Sensor off.							
Bit1:										
Bit0:	REFBE: Inte 0: Internal R 1: Internal R	eference B	uffer disable	ed.	voltage refe	rence drive	n on the VI	REF pin.		

## SFR Definition 6.1. REF0CN: Reference Control

## Table 6.1. Voltage Reference Electrical Characteristics

### $V_{DD}$ = 3.0 V; -40 to +85 °C Unless Otherwise Specified

Parameter	Conditions	Min	Тур	Max	Units
	Internal Reference (REFBE = 1	)			
Output Voltage	25 °C ambient	2.38	2.44	2.50	V
VREF Short-Circuit Current				10	mA
VREF Temperature Coeffi- cient			15		ppm/°C
Load Regulation	Load = 0 to 200 µA to GND		1.5		ppm/µA
VREF Turn-on Time 1	4.7 μF tantalum, 0.1 μF ceramic bypass		2		ms
VREF Turn-on Time 2	0.1 µF ceramic bypass		20		μs
VREF Turn-on Time 3	no bypass cap		10		μs
Power Supply Rejection			140		ppm/V
	External Reference (REFBE = 0	)			
Input Voltage Range		0		V <sub>DD</sub>	V
Input Current	Sample Rate = 200 ksps; VREF = 3.0 V		12		μA
	Bias Generators	•	•	•	
ADC Bias Generator	BIASE = '1'		100		μA
Reference Bias Generator			40		μA



## 7. Comparators

C8051F34x devices include two on-chip programmable voltage Comparators. A block diagram of the comparators is shown in Figure 7.1, where "n" is the comparator number (0 or 1). The two Comparators operate identically with the following exceptions: (1) Their input selections differ, and (2) Comparator0 can be used as a reset source. For input selection details, refer to SFR Definition 7.2 and SFR Definition 7.5.

Each Comparator offers programmable response time and hysteresis, an analog input multiplexer, and two outputs that are optionally available at the Port pins: a synchronous "latched" output (CP0, CP1), or an asynchronous "raw" output (CP0A, CP1A). The asynchronous signal is available even when the system clock is not active. This allows the Comparators to operate and generate an output with the device in STOP mode. When assigned to a Port pin, the Comparator outputs may be configured as open drain or push-pull (see Section "15.2. Port I/O Initialization" on page 147). Comparator0 may also be used as a reset source (see Section "11.5. Comparator0 Reset" on page 103).

The Comparator0 inputs are selected in the CPT0MX register (SFR Definition 7.2). The CMX0P1-CMX0P0 bits select the Comparator0 positive input; the CMX0N1-CMX0N0 bits select the Comparator0 negative input. The Comparator1 inputs are selected in the CPT1MX register (SFR Definition 7.5). The CMX-1P1-CMX1P0 bits select the Comparator1 positive input; the CMX1N1-CMX1N0 bits select the Comparator1 negative input.

**Important Note About Comparator Inputs:** The Port pins selected as Comparator inputs should be configured as analog inputs in their associated Port configuration register, and configured to be skipped by the Crossbar (for details on Port configuration, see Section "15.3. General Purpose Port I/O" on page 150).



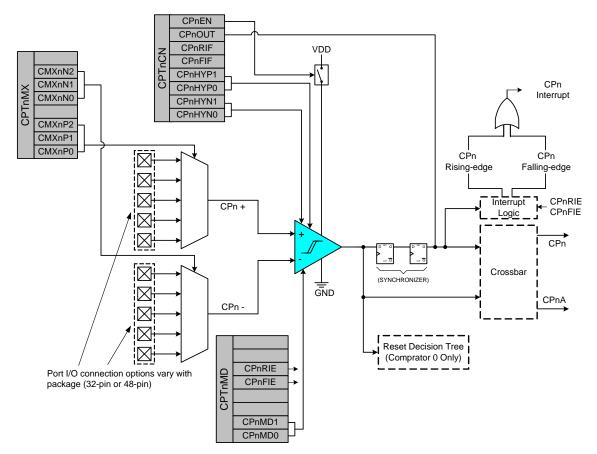


Figure 7.1. Comparator Functional Block Diagram

Comparator outputs can be polled in software, used as an interrupt source, and/or routed to a Port pin. When routed to a Port pin, Comparator outputs are available asynchronous or synchronous to the system clock; the asynchronous output is available even in STOP mode (with no system clock active). When disabled, the Comparator output (if assigned to a Port I/O pin via the Crossbar) defaults to the logic low state, and supply current falls to less than 100 nA. See **Section** "15.1. Priority Crossbar Decoder" on page 144 for details on configuring Comparator outputs via the digital Crossbar. Comparator inputs can be externally driven from -0.25 V to (V<sub>DD</sub>) + 0.25 V without damage or upset. The complete Comparator electrical specifications are given in Table 7.1.

Comparator response time may be configured in software via the CPTnMD registers (see SFR Definition 7.3 and SFR Definition 7.6). Selecting a longer response time reduces the Comparator supply current. See Table 7.1 for complete timing and supply current specifications.



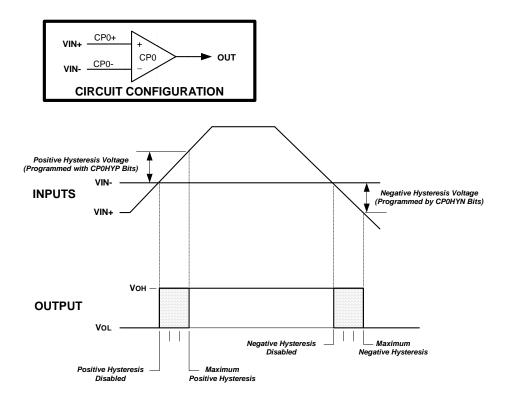


Figure 7.2. Comparator Hysteresis Plot

Comparator hysteresis is programmed using Bits3-0 in the Comparator Control Register CPTnCN (shown in SFR Definition 7.1 and SFR Definition 7.4). The amount of negative hysteresis voltage is determined by the settings of the CPnHYN bits. As shown in Figure 7.2, various levels of negative hysteresis can be programmed, or negative hysteresis can be disabled. In a similar way, the amount of positive hysteresis is determined by the setting the CPnHYP bits.

Comparator interrupts can be generated on both rising-edge and falling-edge output transitions. (For Interrupt enable and priority control, see **Section "9.3. Interrupt Handler" on page 88**.) The CPnFIF flag is set to '1' upon a Comparator falling-edge, and the CPnRIF flag is set to '1' upon the Comparator rising-edge. Once set, these bits remain set until cleared by software. The output state of the Comparator can be obtained at any time by reading the CPnOUT bit. The Comparator is enabled by setting the CPnEN bit to '1', and is disabled by clearing this bit to '0'.



R/W	R	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
CP0E		CPORIF	CP0FIF	CP0HYP1			CPOHYNO	
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
								0x9B
								0/10 2
Bit7:	CP0EN: Cor	mparator0 E	nable Bit.					
	0: Comparat	or0 Disable	d.					
	1: Comparat	or0 Enable	d.					
Bit6:	CP0OUT: Co	omparator0	Output Sta	ate Flag.				
	0: Voltage or							
	1: Voltage or							
Bit5:	CP0RIF: Co							
	0: No Comp				since this fla	ag was last	cleared.	
514	1: Comparat							
Bit4:	CP0FIF: Co	•		•				
	0: No Comp					ag was last	cleared.	
Bits3-2:	1: Comparat	-	-	•				
DII53-2.	CP0HYP1–0 00: Positive	•		e nysteresi		.5.		
	00. Positive 01: Positive							
	10: Positive	•						
	11: Positive							
Bits1-0:	CP0HYN1-(	•		ive Hysteres	sis Control B	its.		
2.101 01	00: Negative	•	-	•				
	01: Negative							
	10: Negative							
	11: Negative	•						
	-							

### SFR Definition 7.1. CPT0CN: Comparator0 Control



#### SFR Definition 7.2. CPT0MX: Comparator0 MUX Selection

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
-	CMX0N2	2 CMX0N	1 CMX0N	0 -	CMX0P2	CMX0P1	CMX0P0	0000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address
								0x9F
Bit7:	UNUSED.	Read = 0b	, Write = do	n't care.				
Bits6–4:				0 Negative Inp				
	These bits	select which	ch Port pin i	is used as the	Comparate	or0 negative	e input.	
	CMX0N1	CMX0N1	CMX0N0	Negative I	nput	Negativ	e Input	]
				(32-pin Pac	kage)	(48-pin P	ackage)	
	0	0	0	P1.1		P2	.1	
	0	0	1	P1.5		P2	.6	
	<u>^</u>		<u> </u>	<b>DO</b> 1		D0	-	
	0	1	0	P2.1		P3	.5	
	0	1 1	0	P2.1 P2.5		P3		-
Di#2.	0	1 0	1 0	P2.5 P0.1			.4	
Bit3: Bits2–0:	0 1 UNUSED. CMX0P2	1 0 Read = 0b CMX0P0: (	1 0 , Write = do Comparator ch Port pin i	P2.5 P0.1 on't care. 0 Positive Inpu is used as the <b>Positive I</b>	Comparato	P4 P0 ect. or0 positive <b>Positive</b>	4 4 input.	
	0 1 UNUSED. CMX0P2 These bits CMX0P1	1 0 Read = 0b CMX0P0: 0 select whic <b>CMX0P1</b>	1 0 , Write = do Comparatori ch Port pin i <b>CMX0P0</b>	P2.5 P0.1 on't care. 0 Positive Inpu is used as the <b>Positive I</b> (32-pin Pac	Comparato	P4 P0 ect. pr0 positive Positive (48-pin P	4 input. Input ackage)	
	0 1 UNUSED. CMX0P2 These bits CMX0P1 0	1 0 Read = 0b CMX0P0: 0 select white <b>CMX0P1</b> 0	1 0 Write = do Comparator ch Port pin i <b>CMX0P0</b> 0	P2.5 P0.1 on't care. 0 Positive Inpu is used as the Positive I (32-pin Pac P1.0	Comparato	P4 P0 ect. pr0 positive Positive (48-pin P P2	4 4 input. e Input ackage)	
	0 1 UNUSED. CMX0P2-0 These bits CMX0P1 0 0	1 0 Read = 0b CMX0P0: 0 select white <b>CMX0P1</b> 0 0	1 0 Write = do Comparator ch Port pin i <b>CMX0P0</b> 0 1	P2.5 P0.1 on't care. 0 Positive Inpu is used as the <b>Positive I</b> (32-pin Pac P1.0 P1.4	Comparato	P4 P0 ect. or0 positive <b>Positive</b> (48-pin P P2 P2	4 4 input. e Input ackage) .0 .5	
	0 1 UNUSED. CMX0P2-0 These bits CMX0P1 0 0 0	1 0 Read = 0b CMX0P0: 0 select white <b>CMX0P1</b> 0 0 1	1 0 write = do Comparator ch Port pin i <b>CMX0P0</b> 0 1 0	P2.5 P0.1 on't care. 0 Positive Inpu is used as the <b>Positive I</b> (32-pin Pac P1.0 P1.4 P2.0	Comparato	P4 P0 ect. or0 positive <b>Positive</b> (48-pin P P2 P2 P3	4 4 input. e Input ackage) .0 .5 .4	
	0 1 UNUSED. CMX0P2-0 These bits CMX0P1 0 0	1 0 Read = 0b CMX0P0: 0 select white <b>CMX0P1</b> 0 0	1 0 Write = do Comparator ch Port pin i <b>CMX0P0</b> 0 1	P2.5 P0.1 on't care. 0 Positive Inpu is used as the <b>Positive I</b> (32-pin Pac P1.0 P1.4	Comparato	P4 P0 ect. or0 positive <b>Positive</b> (48-pin P P2 P2	4 4 input. e Input ackage) .0 .5 .4 .3	



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value		
-	-	<b>CP0RIE</b>	CP0FIE	-	-	CP0MD1	CP0MD0	00000010		
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:		
								0x9D		
Bit5: Bit4: Bits3–2:	Bits7–6: UNUSED. Read = 00b. Write = don't care. Bit5: CP0RIE: Comparator0 Rising-Edge Interrupt Enable. 0: Comparator0 rising-edge interrupt disabled. 1: Comparator0 rising-edge interrupt enabled.									
	These bits s				141010.					
	Mode	CP0MD1	CP0MD0	CP0 Res	ponse Tim	e*				
	0	0	0	Fastest	Response					
	1	0	1							
	2	1	0							
	3	1	1	Lowe	st Power					
* See Table 7.1 for response time parameters.										

#### SFR Definition 7.3. CPT0MD: Comparator0 Mode Selection



R/W	R	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value		
CP1EN	CP1OUT	CP1RIF	CP1FIF	CP1HYP1	CP1HYP0	CP1HYN1	CP1HYN0	00000000		
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:		
								0x9A		
Bit7:	CP1EN: Cor	nparator1 E	nable Bit.							
	0: Comparator1 Disabled.									
	1: Comparat	or1 Enable	d.							
Bit6:	CP1OUT: Co			ite Flag.						
	0: Voltage or									
	1: Voltage or									
Bit5:	CP1RIF: Co			•						
	0: No Compa				since this fla	ag was last	cleared.			
	1: Comparat	-	-							
Bit4:	CP1FIF: Cor			•						
	0: No Compa				since this fi	ag was las	t cleared.			
	1: Comparat	0	0		Control Di					
Bits3–2:				e Hysteresis	s Control Bil	.S.				
	00: Positive 01: Positive	•								
	10: Positive									
	11: Positive I									
Bits1–0:	CP1HYN1-C			ve Hysteres	is Control B	ite				
Ditor 0.	00: Negative	•	-	ive riysteres						
	01: Negative									
	10: Negative									
	11: Negative	•								
		,								

### SFR Definition 7.4. CPT1CN: Comparator1 Control



### SFR Definition 7.5. CPT1MX: Comparator1 MUX Selection

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
-	CMX1N2	2 CMX1N	1 CMX1N	10 - CI	/IX1P2	CMX1P1	CMX1P0	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address
Bit7: Bits6–4:		CMX1N0: (	Comparato	1 Negative Input				0x9E
	CMX1N2	CMX1N1	•	is used as the Co		Negative	•	
	O MARINE			(32-pin Packag		48-pin Pack		
	0	0	0	P1.3		P2.3		
	0	0	1	P1.7		P3.1		
	0	1	0	D0 0		P4.0		
	0	I	0	P2.3		F4.0		
	0	1	1	P2.3 P2.7		P4.0 P4.6		
	-	-		-		-		
Bit3: Bits2–0:	0 1 UNUSED. CMX1P1-0	1 0 Read = 0b CMX1P0: 0	1 0 , Write = do Comparator ch Port pin	P2.7 P0.5 on't care. 1 Positive Input M is used as the Co Positive Inpu	mparato	P4.6 P1.2 ect. or1 positive <b>Positive In</b>	put	
	0 1 UNUSED. CMX1P1-4 These bits	1 0 Read = 0b CMX1P0: 0 select white CMX1P1	1 0 , Write = dc Comparator ch Port pin CMX1P0	P2.7 P0.5 on't care. 1 Positive Input M is used as the Co Positive Inpu (32-pin Packag	mparato	P4.6 P1.2 ect. pr1 positive Positive In 48-pin Pacl	put	
	0 1 UNUSED. CMX1P1-0 These bits CMX1P2 0	1 0 Read = 0b CMX1P0: 0 select white CMX1P1	1 0 , Write = do Comparator ch Port pin	P2.7 P0.5 Pon't care. 1 Positive Input M is used as the Co Positive Inpu (32-pin Packag P1.2	mparato	P4.6 P1.2 ect. pr1 positive Positive In 48-pin Pacl P2.2	put	
	0 1 UNUSED. CMX1P1-0 These bits CMX1P2 0 0	1 0 Read = 0b CMX1P0: 0 select white <b>CMX1P1</b> 0 0	1 0 Comparator ch Port pin CMX1P0 0 1	P2.7 P0.5 on't care. 1 Positive Input M is used as the Co Positive Inpu (32-pin Packag	mparato	P4.6 P1.2 ect. pr1 positive Positive In 48-pin Pacl	put	
	0 1 UNUSED. CMX1P1-0 These bits CMX1P2 0	1 0 Read = 0b CMX1P0: 0 select white CMX1P1	1 0 Write = do Comparator ch Port pin <b>CMX1P0</b> 0	P2.7 P0.5 Pon't care. 1 Positive Input M is used as the Co Positive Inpu (32-pin Packag P1.2 P1.6	mparato	P4.6 P1.2 ect. or1 positive Positive In 48-pin Pacl P2.2 P3.0	put	



SFR Definition 7.6. CPT1MD: Comparator1 Mc	ode Selection
--	---------------

Bit7       Bit6       Bit5       Bit4       Bit3       Bit2       Bit1       Bit0       SFR Addres         0x9C         Bits7-6:       UNUSED. Read = 00b, Write = don't care.         Bit5:       CP1RIE: Comparator1 Rising-Edge Interrupt Enable.         0:       Comparator1 rising-edge interrupt disabled.         1:       Comparator1 rising-edge interrupt enabled.         Bit4:       CP1FIE: Comparator1 Falling-Edge Interrupt Enable.         0:       Comparator1 Falling-Edge Interrupt Enable.         0:       Comparator1 falling-edge interrupt disabled.         1:       Comparator1 falling-edge interrupt disabled.	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value		
Bits7-6:UNUSED. Read = 00b, Write = don't care.Bit5:CP1RIE: Comparator1 Rising-Edge Interrupt Enable. 0: Comparator1 rising-edge interrupt disabled. 1: Comparator1 rising-edge interrupt enabled.Bit4:CP1FIE: Comparator1 Falling-Edge Interrupt Enable. 0: Comparator1 falling-edge interrupt disabled. 1: Comparator1 falling-edge interrupt enabled.Bit51-0:CP1MD1-CP1MD0: Comparator1 Mode Select. These bits select the response time for Comparator1.ModeCP1MD1CP1MD0CP1 Response Time* 100102101211Lowest Power	-	-	CP1RIE	CP1FIE	-	-	CP1MD1	CP1MD0	00000010		
Bits7-6:UNUSED. Read = 00b, Write = don't care.Bit5:CP1RIE: Comparator1 Rising-Edge Interrupt Enable. 0: Comparator1 rising-edge interrupt disabled. 1: Comparator1 rising-edge interrupt enabled.Bit4:CP1FIE: Comparator1 Falling-Edge Interrupt Enable. 0: Comparator1 falling-edge interrupt disabled. 1: Comparator1 falling-edge interrupt enabled.Bits1-0:CP1MD1-CP1MD0: Comparator1 Mode Select. These bits select the response time for Comparator1.	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address		
Bit5:       CP1RIE: Comparator1 Rising-Edge Interrupt Enable.         0: Comparator1 rising-edge interrupt disabled.         1: Comparator1 rising-edge interrupt enabled.         Bit4:       CP1FIE: Comparator1 Falling-Edge Interrupt Enable.         0: Comparator1 falling-edge interrupt disabled.         1: Comparator1 falling-edge interrupt disabled.         1: Comparator1 falling-edge interrupt enabled.         Bits1-0:       CP1MD1-CP1MD0: Comparator1 Mode Select.         These bits select the response time for Comparator1.         Mode       CP1MD1         CP1MD0       CP1 Response Time*         0       0         1       0         2       1         3       1         1       Lowest Power									0x9C		
Bit5:       CP1RIE: Comparator1 Rising-Edge Interrupt Enable.         0: Comparator1 rising-edge interrupt disabled.         1: Comparator1 rising-edge interrupt enabled.         Bit4:       CP1FIE: Comparator1 Falling-Edge Interrupt Enable.         0: Comparator1 falling-edge interrupt disabled.         1: Comparator1 falling-edge interrupt disabled.         1: Comparator1 falling-edge interrupt enabled.         Bits1-0:       CP1MD1-CP1MD0: Comparator1 Mode Select.         These bits select the response time for Comparator1.         Mode       CP1MD1         CP1MD0       CP1 Response Time*         0       0         1       0         2       1         3       1         1       Lowest Power	Bits7–6 <sup>.</sup>	UNUSED F	}ead = 00b	Write = don	't care						
0: Comparator1 rising-edge interrupt disabled. 1: Comparator1 rising-edge interrupt enabled. Bit4: CP1FIE: Comparator1 Falling-Edge Interrupt Enable. 0: Comparator1 falling-edge interrupt disabled. 1: Comparator1 falling-edge interrupt enabled. Bits1–0: CP1MD1–CP1MD0: Comparator1 Mode Select. These bits select the response time for Comparator1. $\frac{Mode \ CP1MD1 \ CP1MD0 \ CP1 Response Time*}{0 \ 0 \ 0 \ Fastest Response}$ $\frac{1 \ 0 \ 1 \ 0 \ 1}{3 \ 1 \ 1 \ Lowest Power}$											
1: Comparator1 rising-edge interrupt enabled.Bit4:CP1FIE: Comparator1 Falling-Edge Interrupt Enable. 0: Comparator1 falling-edge interrupt disabled. 1: Comparator1 falling-edge interrupt enabled.Bits1-0:CP1MD1-CP1MD0: Comparator1 Mode Select. These bits select the response time for Comparator1.ModeCP1MD1CP1MD1CP1MD0CP1 Response Time* 000102121311Lowest Power											
Bit4:       CP1FIE: Comparator1 Falling-Edge Interrupt Enable.         0:       Comparator1 falling-edge interrupt disabled.         1:       Comparator1 falling-edge interrupt enabled.         Bits1–0:       CP1MD1–CP1MD0: Comparator1 Mode Select.         These bits select the response time for Comparator1.         Mode       CP1MD1       CP1MD0       CP1 Response Time*         0       0       0       Fastest Response         1       0       1       2         2       1       0       1         3       1       1       Lowest Power											
0: Comparator1 falling-edge interrupt disabled. 1: Comparator1 falling-edge interrupt enabled. Bits1–0: CP1MD1–CP1MD0: Comparator1 Mode Select. These bits select the response time for Comparator1. Mode CP1MD1 CP1MD0 CP1 Response Time* 0 0 0 Fastest Response 1 0 1 2 1 0 3 1 1 Lowest Power											
1: Comparator1 falling-edge interrupt enabled. Bits1–0: CP1MD1–CP1MD0: Comparator1 Mode Select. These bits select the response time for Comparator1. Mode CP1MD1 CP1MD0 CP1 Response Time* 0 0 0 Fastest Response 1 0 1 2 1 0 3 1 1 Lowest Power											
Bits1-0:       CP1MD1-CP1MD0: Comparator1 Mode Select. These bits select the response time for Comparator1.         Mode       CP1MD1       CP1MD0       CP1 Response Time*         0       0       0       Fastest Response         1       0       1       1         2       1       0       1         3       1       1       Lowest Power		•	•	•							
ModeCP1MD1CP1MD0CP1 Response Time*000Fastest Response101210311Lowest Power1	Bits1–0:		•	•		t.					
0         0         0         Fastest Response           1         0         1         1           2         1         0         1           3         1         1         Lowest Power		These bits s	elect the re	sponse time	e for Compa	rator1.					
0         0         0         Fastest Response           1         0         1         1           2         1         0         1           3         1         1         Lowest Power											
1         0         1           2         1         0           3         1         1         Lowest Power			0041404	004400	004.0		-				
2         1         0           3         1         1         Lowest Power											
3 1 1 Lowest Power			0								
			0								
* See Table 7.1 for response time parameters.		0	0 0	0 1							
* See Table 7.1 for response time parameters.		0 1 2	0 0	0 1	Fastest	Response					
		0 1 2 3	0 0 1 1	0 1 0 1	Fastest Lowe	Response					
	* See Tab	0 1 2 3	0 0 1 1	0 1 0 1	Fastest Lowe	Response					



### **Table 7.1. Comparator Electrical Characteristics**

#### $V_{DD}$ = 3.0 V, -40 to +85 °C unless otherwise noted.

All specifications apply to both Comparator0 and Comparator1 unless otherwise noted.

Parameter	Conditions	Min	Тур	Мах	Units
Response Time:	CP0+-CP0-=100 mV		100		ns
Mode 0, Vcm* = 1.5 V	CP0+ - CP0- = -100 mV		250		ns
Response Time:	CP0+-CP0-=100 mV		175		ns
Mode 1, Vcm* = 1.5 V	CP0+ - CP0- = -100 mV		500		ns
Response Time:	CP0+-CP0-=100 mV		320		ns
Mode 2, Vcm* = 1.5 V	CP0+ - CP0- = -100 mV		1100		ns
Response Time:	CP0+ - CP0- = 100 mV		1050		ns
Mode 3, Vcm* = 1.5 V	CP0+ - CP0- = -100 mV		5200		ns
Common-Mode Rejection Ratio			1.5	4	mV/V
Positive Hysteresis 1	CP0HYP1-0 = 00		0	1	mV
Positive Hysteresis 2	CP0HYP1-0 = 01	2	5	10	mV
Positive Hysteresis 3	CP0HYP1-0 = 10	7	10	20	mV
Positive Hysteresis 4	CP0HYP1-0 = 11	15	20	30	mV
Negative Hysteresis 1	CP0HYN1-0 = 00		0	1	mV
Negative Hysteresis 2	CP0HYN1-0 = 01	2	5	10	mV
Negative Hysteresis 3	CP0HYN1-0 = 10	7	10	20	mV
Negative Hysteresis 4	CP0HYN1-0 = 11	15	20	30	mV
Inverting or Non-Inverting Input Voltage Range		-0.25		V <sub>DD</sub> + 0.25	V
Input Capacitance			3		pF
Input Bias Current			0.001		nA
Input Offset Voltage		-5		+5	mV
	Power Supp	ly			
Power Supply Rejection			0.1		mV/V
Power-up Time			10		μs
	Mode 0		7.6		μA
Supply Current at DC	Mode 1		3.2		μA
Supply Current at DC	Mode 2		1.3		μA
	Mode 3		0.4		μA

\*Note: Vcm is the common-mode voltage on CP0+ and CP0-.



## 8. Voltage Regulator (REG0)

C8051F34x devices include a voltage regulator (REG0). When enabled, the REG0 output appears on the  $V_{DD}$  pin and can be used to power external devices. REG0 can be enabled/disabled by software using bit REGEN in register REG0CN. See Table 8.1 for REG0 electrical characteristics.

Note that the VBUS signal must be connected to the VBUS pin when using the device in a USB network. The VBUS signal should only be connected to the REGIN pin when operating the device as a bus-powered function. REG0 configuration options are shown in Figure 8.1–Figure 8.4.

#### 8.1. Regulator Mode Selection

REG0 offers a low power mode intended for use when the device is in suspend mode. In this low power mode, the REG0 output remains as specified; however the REG0 dynamic performance (response time) is degraded. See Table 8.1 for normal and low power mode supply current specifications. The REG0 mode selection is controlled via the REGMOD bit in register REG0CN.

#### 8.2. VBUS Detection

When the USB Function Controller is used (see section **Section "16. Universal Serial Bus Controller (USB0)" on page 159**), the VBUS signal should be connected to the VBUS pin. The VBSTAT bit (register REGOCN) indicates the current logic level of the VBUS signal. If enabled, a VBUS interrupt will be generated when the VBUS signal matches the polarity selected by the VBPOL bit in register REGOCN. The VBUS interrupt is level-sensitive, and has no associated interrupt pending flag. The VBUS interrupt will be active as long as the VBUS signal matches the polarity selected by VBPOL. See Table 8.1 for VBUS input parameters.

**Important Note:** When USB is selected as a reset source, a system reset will be generated when the VBUS signal matches the polarity selected by the VBPOL bit. See **Section "11. Reset Sources" on page 100** for details on selecting USB as a reset source

#### Table 8.1. Voltage Regulator Electrical Specifications

-40 to +85	°C unless	otherwise s	specified.
------------	-----------	-------------	------------

Parameter	Conditions	Min	Тур	Max	Units
Input Voltage Range <sup>1</sup>		2.7		5.25	V
Output Voltage (V <sub>DD</sub> ) <sup>2</sup>	Output Current = 1 to 100 mA	3.0	3.3	3.6	V
Output Current <sup>2</sup>				100	mA
VBUS Detection Input Low Voltage				1.0	V
VBUS Detection Input High Voltage		3.0			V
Bias Current	Normal Mode (REGMOD = '0') Low Power Mode (REGMOD = '1')		65 35	111 61	μA
Dropout Voltage (V <sub>DO</sub> ) <sup>3</sup>			1		mV/mA

Notes:

1. Input range specified for regulation. When an external regulator is used, should be tied to  $V_{DD}$ .

- 2. Output current is total regulator output, including any current required by the C8051F34x.
- 3. The minimum input voltage is 2.70 V or VDD +  $V_{DO}$  (max load), whichever is greater.



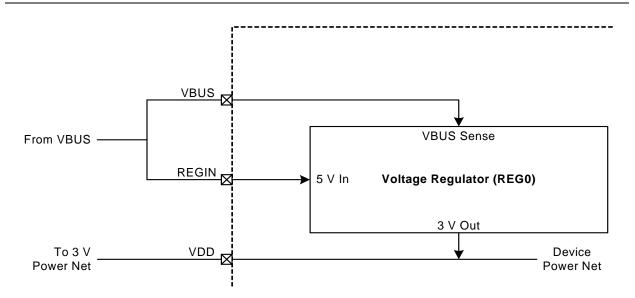


Figure 8.1. REG0 Configuration: USB Bus-Powered

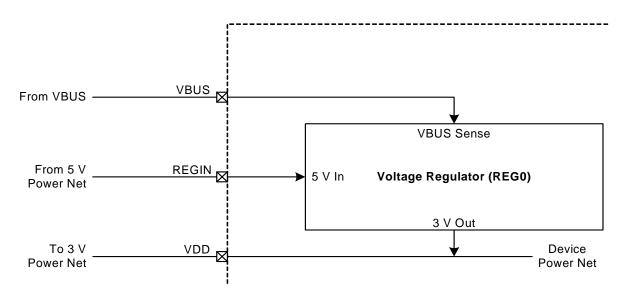


Figure 8.2. REG0 Configuration: USB Self-Powered



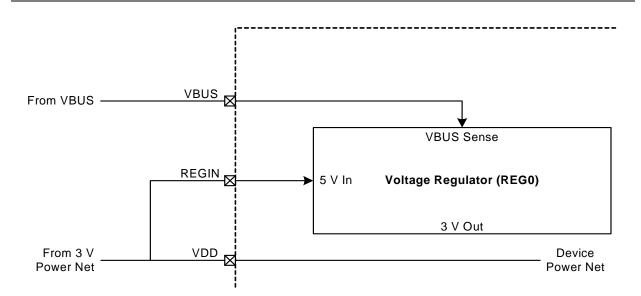


Figure 8.3. REG0 Configuration: USB Self-Powered, Regulator Disabled

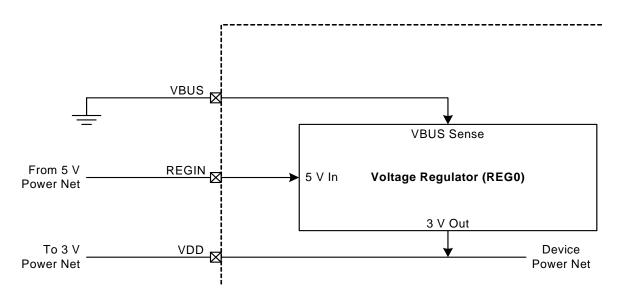


Figure 8.4. REG0 Configuration: No USB Connection



R/W	R	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value					
REGDIS		VBPOL	REGMOD	Reserved	Reserved	Reserved	Reserved	1					
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:					
Ditt	Bito	Bito	Bitt	Bito	DAL	Ditt	Bito	0xC9					
								0.00					
Bit7:	REGDIS: Vo	ltage Regu	lator Disable	e.									
	0: Voltage R												
	1: Voltage R	egulator Dis	sabled.										
Bit6:	VBSTAT: VB	US Signal	Status.										
	0: VBUS sig	nal currentl	y absent (de	evice not at	tached to U	SB network	x).						
	1: VBUS signal currently present (device attached to USB network).												
Bit5:	VBPOL: VBUS Interrupt Polarity Select.												
	This bit selects the VBUS interrupt polarity.												
0: VBUS interrupt active when VBUS is low.													
Dit 4		1: VBUS interrupt active when VBUS is high.											
Bit4:	REGMOD: Voltage Regulator Mode Select.												
		This bit selects the Voltage Regulator mode. When REGMOD is set to '1', the voltage regu-											
	lator operates in low power (suspend) mode. 0: USB0 Voltage Regulator in normal mode.												
	1: USB0 Vol	0 0											
Bits3-0:	Reserved, R	0 0	•										

### SFR Definition 8.1. REG0CN: Voltage Regulator Control



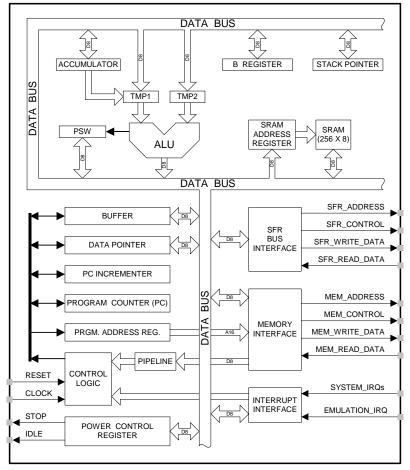
## 9. CIP-51 Microcontroller

The MCU system controller core is the CIP-51 microcontroller. The CIP-51 is fully compatible with the MCS-51<sup>™</sup> instruction set; standard 803x/805x assemblers and compilers can be used to develop software. The MCU family has a superset of all the peripherals included with a standard 8051. Included are four 16-bit counter/timers (see description in Section 21), an enhanced full-duplex UART (see description in Section 18), an Enhanced SPI (see description in Section 20), 256 bytes of internal RAM, 128 byte Special Function Register (SFR) address space (Section 9.2.6), and 25 Port I/O (see description in Section 15). The CIP-51 also includes on-chip debug hardware (see description in Section 23), and interfaces directly with the analog and digital subsystems providing a complete data acquisition or control-system solution in a single integrated circuit.

The CIP-51 Microcontroller core implements the standard 8051 organization and peripherals as well as additional custom peripherals and functions to extend its capability (see Figure 9.1 for a block diagram). The CIP-51 includes the following features:

- Fully Compatible with MCS-51 Instruction Set
- 0 to 48 MHz Clock Frequency
- 256 Bytes of Internal RAM
- 25 Port I/O

- Extended Interrupt Handler
- Reset Input
- Power Management Modes
- On-chip Debug Logic
- Program and Data Memory Security



#### Figure 9.1. CIP-51 Block Diagram



#### Performance

The CIP-51 employs a pipelined architecture that greatly increases its instruction throughput over the standard 8051 architecture. In a standard 8051, all instructions except for MUL and DIV take 12 or 24 system clock cycles to execute, and usually have a maximum system clock of 12 MHz. By contrast, the CIP-51 core executes 70% of its instructions in one or two system clock cycles, with no instructions taking more than eight system clock cycles.

With the CIP-51's maximum system clock at 25 MHz, it has a peak throughput of 25 MIPS. The CIP-51 has a total of 109 instructions. The table below shows the total number of instructions that for execution time.

Clocks to Execute	1	2	2/4	3	3/5	4	5	4/6	6	8
Number of Instructions	26	50	5	10	7	5	2	1	2	1

#### **Programming and Debugging Support**

In-system programming of the Flash program memory and communication with on-chip debug support logic is accomplished via the Silicon Labs 2-Wire Development Interface (C2). Note that the re-programmable Flash can also be read and changed a single byte at a time by the application software using the MOVC and MOVX instructions. This feature allows program memory to be used for non-volatile data storage as well as updating program code under software control.

The on-chip Silicon Labs 2-Wire (C2) Development Interface allows non-intrusive (uses no on-chip resources), full speed, in-circuit debugging using the production MCU installed in the final application. This debug logic supports inspection and modification of memory and registers, setting breakpoints, single stepping, run and halt commands. All analog and digital peripherals are fully functional while debugging using C2. The two C2 interface pins can be shared with user functions, allowing in-system debugging without occupying package pins. C2 details can be found in Section "23. C2 Interface" on page 271.

The CIP-51 is supported by development tools from Silicon Labs and third party vendors. Silicon Labs provides an integrated development environment (IDE) including editor, debugger, and programmer. The IDE's debugger and programmer interface to the CIP-51 via the C2 interface to provide fast and efficient in-system device programming and debugging. An 8051 assembler, linker and evaluation 'C' compiler are included in the Development Kit. Many third party macro assemblers and C compilers are also available, which can be used directly with the IDE.

#### 9.1. Instruction Set

The instruction set of the CIP-51 System Controller is fully compatible with the standard MCS-51<sup>™</sup> instruction set. Standard 8051 development tools can be used to develop software for the CIP-51. All CIP-51 instructions are the binary and functional equivalent of their MCS-51<sup>™</sup> counterparts, including opcodes, addressing modes and effect on PSW flags. However, instruction timing is different than that of the standard 8051.

#### 9.1.1. Instruction and CPU Timing

In many 8051 implementations, a distinction is made between machine cycles and clock cycles, with machine cycles varying from 2 to 12 clock cycles in length. However, the CIP-51 implementation is based solely on clock cycle timing. All instruction timings are specified in terms of clock cycles.

Due to the pipelined architecture of the CIP-51, most instructions execute in the same number of clock cycles as there are program bytes in the instruction. Conditional branch instructions take two fewer clock cycles to complete when the branch is not taken as opposed to when the branch is taken. Table 9.1 is the CIP-51 Instruction Set Summary, which includes the mnemonic, number of bytes, and number of clock cycles for each instruction.



#### 9.1.2. MOVX Instruction and Program Memory

In the CIP-51, the MOVX instruction serves three purposes: accessing on-chip XRAM, accessing off-chip data XRAM (only on C8051F340/1/4/5/8 devices), and accessing on-chip program Flash memory. The Flash access feature provides a mechanism for user software to update program code and use the program memory space for non-volatile data storage (see Section "12. Flash Memory" on page 107). The External Memory Interface (only on C8051F340/1/4/5/8 devices) provides a fast access interface to off-chip data XRAM (or memory-mapped peripherals) via the MOVX instruction. Refer to Section "13. External Data Memory Interface and On-Chip XRAM" on page 114. for details.

Mnemonic	Mnemonic Description		Clock Cycles	
	Arithmetic Operations			
ADD A, Rn	Add register to A	1	1	
ADD A, direct	Add direct byte to A	2	2	
ADD A, @Ri	Add indirect RAM to A	1	2	
ADD A, #data	Add immediate to A	2	2	
ADDC A, Rn	Add register to A with carry	1	1	
ADDC A, direct	Add direct byte to A with carry	2	2	
ADDC A, @Ri	Add indirect RAM to A with carry	1	2	
ADDC A, #data	Add immediate to A with carry	2	2	
SUBB A, Rn	Subtract register from A with borrow	1	1	
SUBB A, direct	Subtract direct byte from A with borrow	2	2	
SUBB A, @Ri	Subtract indirect RAM from A with borrow	1	2	
SUBB A, #data	Subtract immediate from A with borrow	2	2	
INC A	Increment A	1	1	
INC Rn	Increment register	1	1	
INC direct	Increment direct byte	2	2	
INC @Ri	Increment indirect RAM	1	2	
DEC A	Decrement A	1	1	
DEC Rn	Decrement register	1	1	
DEC direct	Decrement direct byte	2	2	
DEC @Ri	Decrement indirect RAM	1	2	
INC DPTR	Increment Data Pointer	1	1	
MUL AB	Multiply A and B	1	4	
DIV AB	Divide A by B	1	8	
DA A	Decimal adjust A	1	1	
	Logical Operations	•		
ANL A, Rn	AND Register to A	1	1	
ANL A, direct	AND direct byte to A	2	2	
ANL A, @Ri	AND indirect RAM to A	1	2	
ANL A, #data	AND immediate to A	2	2	
ANL direct, A	AND A to direct byte	2	2	
ANL direct, #data	AND immediate to direct byte	3	3	
ORL A, Rn	OR Register to A	1	1	
ORL A, direct	OR direct byte to A	2	2	
ORL A, @Ri	OR indirect RAM to A	1	2	

## Table 9.1. CIP-51 Instruction Set Summary



Table 9.1. CIP-51 Instruction Set Summary (	(Continued)
---	-------------

Mnemonic	Description	Bytes	Clock Cycles	
ORL A, #data	OR immediate to A	2	2	
ORL direct, A	OR A to direct byte	2	2	
ORL direct, #data	OR immediate to direct byte	3	3	
XRL A, Rn	Exclusive-OR Register to A	1	1	
XRL A, direct	Exclusive-OR direct byte to A	2	2	
XRL A, @Ri	Exclusive-OR indirect RAM to A	1	2	
XRL A, #data	Exclusive-OR immediate to A	2	2	
XRL direct, A	Exclusive-OR A to direct byte	2	2	
XRL direct, #data	Exclusive-OR immediate to direct byte	3	3	
CLR A	Clear A	1	1	
CPL A	Complement A	1	1	
RL A	Rotate A left	1	1	
RLC A	Rotate A left through Carry	1	1	
RR A	Rotate A right	1	1	
RRC A	Rotate A right through Carry	1	1	
SWAP A	Swap nibbles of A	1	1	
-	Data Transfer			
MOV A, Rn	Move Register to A	1	1	
MOV A, direct	Move direct byte to A	2	2	
MOV A, @Ri	Move indirect RAM to A	1	2	
MOV A, #data	Move immediate to A	2	2	
MOV Rn, A	Move A to Register	1	1	
MOV Rn, direct	Move direct byte to Register	2	2	
MOV Rn, #data	Move immediate to Register	2	2	
MOV direct, A	Move A to direct byte	2	2	
MOV direct, Rn	Move Register to direct byte	2	2	
MOV direct, direct	Move direct byte to direct byte	3	3	
MOV direct, @Ri	Move indirect RAM to direct byte	2	2	
MOV direct, #data	Move immediate to direct byte	3	3	
MOV @Ri, A	Move A to indirect RAM	1	2	
MOV @Ri, direct	Move direct byte to indirect RAM	2	2	
MOV @Ri, #data	Move immediate to indirect RAM	2	2	
MOV OPTR, #data16	Load DPTR with 16-bit constant	3	3	
MOVC A, @A+DPTR	Move code byte relative DPTR to A	1	3	
MOVC A, @A+DFTR MOVC A, @A+PC	Move code byte relative DC to A	1	3	
MOVC A, @A+PC MOVX A, @Ri		1	3	
	Move external data (8-bit address) to A			
MOVX @Ri, A	Move A to external data (8-bit address)	1	3	
MOVX A, @DPTR	Move external data (16-bit address) to A	1	3	
MOVX @DPTR, A	Move A to external data (16-bit address)	1	3	
PUSH direct	Push direct byte onto stack	2	2	
POP direct	Pop direct byte from stack	2	2	
XCH A, Rn	Exchange Register with A	1	1	
XCH A, direct	Exchange direct byte with A	2	2	
XCH A, @Ri	Exchange indirect RAM with A	1	2	
XCHD A, @Ri	Exchange low nibble of indirect RAM with A	1	2	



Table 9.1. CIP-51	Instruction	Set Summary	(Continued)

Mnemonic	Description	Bytes	Clock Cycles
	Boolean Manipulation		1
CLR C	Clear Carry	1	1
CLR bit	Clear direct bit	2	2
SETB C	Set Carry	1	1
SETB bit	Set direct bit	2	2
CPL C	Complement Carry	1	1
CPL bit	Complement direct bit	2	2
ANL C, bit	AND direct bit to Carry	2	2
ANL C, /bit	AND complement of direct bit to Carry	2	2
ORL C, bit	OR direct bit to carry	2	2
ORL C, /bit	OR complement of direct bit to Carry	2	2
MOV C, bit	Move direct bit to Carry	2	2
MOV bit, C	Move Carry to direct bit	2	2
JC rel	Jump if Carry is set	2	2/4
JNC rel	Jump if Carry is not set	2	2/4
JB bit, rel	Jump if direct bit is set	3	3/5
JNB bit, rel	Jump if direct bit is not set	3	3/5
JBC bit, rel	Jump if direct bit is set and clear bit	3	3/5
	Program Branching		1
ACALL addr11	Absolute subroutine call	2	4
LCALL addr16	Long subroutine call	3	5
RET	Return from subroutine	1	6
RETI	Return from interrupt	1	6
AJMP addr11	Absolute jump	2	4
LJMP addr16	Long jump	3	5
SJMP rel	Short jump (relative address)	2	4
JMP @A+DPTR	Jump indirect relative to DPTR	1	4
JZ rel	Jump if A equals zero	2	2/4
JNZ rel	Jump if A does not equal zero	2	2/4
CJNE A, direct, rel	Compare direct byte to A and jump if not equal	3	3/5
CJNE A, #data, rel	Compare immediate to A and jump if not equal	3	3/5
CJNE Rn, #data, rel	Compare immediate to Register and jump if not equal	3	3/5
CJNE @Ri, #data, rel	Compare immediate to indirect and jump if not equal	3	4/6
DJNZ Rn, rel	Decrement Register and jump if not zero	2	2/4
DJNZ direct, rel	Decrement direct byte and jump if not zero	3	3/5
NOP	No operation	1	1



#### Notes on Registers, Operands and Addressing Modes:

Rn - Register R0-R7 of the currently selected register bank.

@Ri - Data RAM location addressed indirectly through R0 or R1.

**rel** - 8-bit, signed (two's complement) offset relative to the first byte of the following instruction. Used by SJMP and all conditional jumps.

**direct** - 8-bit internal data location's address. This could be a direct-access Data RAM location (0x00-0x7F) or an SFR (0x80-0xFF).

#data - 8-bit constant

#data16 - 16-bit constant

bit - Direct-accessed bit in Data RAM or SFR

**addr11** - 11-bit destination address used by ACALL and AJMP. The destination must be within the same 2K-byte page of program memory as the first byte of the following instruction.

**addr16** - 16-bit destination address used by LCALL and LJMP. The destination may be anywhere within the 8K-byte program memory space.

There is one unused opcode (0xA5) that performs the same function as NOP. All mnemonics copyrighted © Intel Corporation 1980.



#### 9.2. Memory Organization

The memory organization of the CIP-51 System Controller is similar to that of a standard 8051. There are two separate memory spaces: program memory and data memory. Program and data memory share the same address space but are accessed via different instruction types. The CIP-51 memory organization is shown in Figure 9.2 and Figure 9.3.

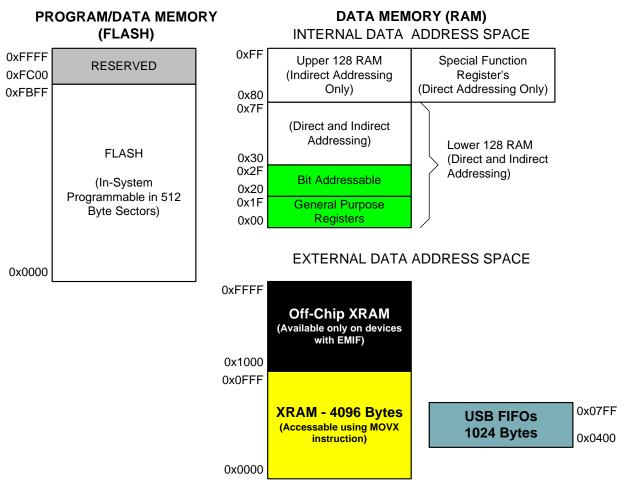


Figure 9.2. On-Chip Memory Map for 64 kB Devices



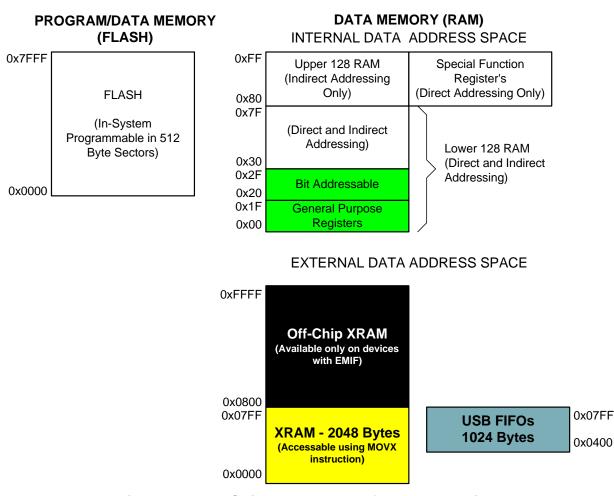


Figure 9.3. On-Chip Memory Map for 32 kB Devices

#### 9.2.1. Program Memory

The CIP-51 core has a 64k-byte program memory space. The C8051F34x implements 64k or 32k bytes of this program memory space as in-system, re-programmable Flash memory. Note that on the 64k versions of the C8051F34x, addresses above 0xFBFF are reserved.

Program memory is normally assumed to be read-only. However, the CIP-51 can write to program memory by setting the Program Store Write Enable bit (PSCTL.0) and using the MOVX instruction. This feature provides a mechanism for the CIP-51 to update program code and use the program memory space for non-volatile data storage. Refer to Section "12. Flash Memory" on page 107 for further details.



#### 9.2.2. Data Memory

The CIP-51 includes 256 of internal RAM mapped into the data memory space from 0x00 through 0xFF. The lower 128 bytes of data memory are used for general purpose registers and scratch pad memory. Either direct or indirect addressing may be used to access the lower 128 bytes of data memory. Locations 0x00 through 0x1F are addressable as four banks of general purpose registers, each bank consisting of eight byte-wide registers. The next 16 bytes, locations 0x20 through 0x2F, may either be addressed as bytes or as 128 bit locations accessible with the direct addressing mode.

The upper 128 bytes of data memory are accessible only by indirect addressing. This region occupies the same address space as the Special Function Registers (SFR) but is physically separate from the SFR space. The addressing mode used by an instruction when accessing locations above 0x7F determines whether the CPU accesses the upper 128 bytes of data memory space or the SFRs. Instructions that use direct addressing will access the SFR space. Instructions using indirect addressing above 0x7F access the upper 128 bytes of data memory organization of the CIP-51.

#### 9.2.3. General Purpose Registers

The lower 32 bytes of data memory, locations 0x00 through 0x1F, may be addressed as four banks of general-purpose registers. Each bank consists of eight byte-wide registers designated R0 through R7. Only one of these banks may be enabled at a time. Two bits in the program status word, RS0 (PSW.3) and RS1 (PSW.4), select the active register bank (see description of the PSW in SFR Definition 9.4). This allows fast context switching when entering subroutines and interrupt service routines. Indirect addressing modes use registers R0 and R1 as index registers.

#### 9.2.4. Bit Addressable Locations

In addition to direct access to data memory organized as bytes, the sixteen data memory locations at 0x20 through 0x2F are also accessible as 128 individually addressable bits. Each bit has a bit address from 0x00 to 0x7F. Bit 0 of the byte at 0x20 has bit address 0x00 while bit7 of the byte at 0x20 has bit address 0x07. Bit 7 of the byte at 0x2F has bit address 0x7F. A bit access is distinguished from a full byte access by the type of instruction used (bit source or destination operands as opposed to a byte source or destination).

The MCS-51<sup>™</sup> assembly language allows an alternate notation for bit addressing of the form XX.B where XX is the byte address and B is the bit position within the byte. For example, the instruction:

MOV C, 22h.3

moves the Boolean value at 0x13 (bit 3 of the byte at location 0x22) into the Carry flag.

#### 9.2.5. Stack

A programmer's stack can be located anywhere in the 256-byte data memory. The stack area is designated using the Stack Pointer (SP, 0x81) SFR. The SP will point to the last location used. The next value pushed on the stack is placed at SP+1 and then SP is incremented. A reset initializes the stack pointer to location 0x07. Therefore, the first value pushed on the stack is placed at location 0x08, which is also the first register (R0) of register bank 1. Thus, if more than one register bank is to be used, the SP should be initialized to a location in the data memory not being used for data storage. The stack depth can extend up to 256 bytes.



#### 9.2.6. Special Function Registers

The direct-access data memory locations from 0x80 to 0xFF constitute the special function registers (SFRs). The SFRs provide control and data exchange with the CIP-51's resources and peripherals. The CIP-51 duplicates the SFRs found in a typical 8051 implementation as well as implementing additional SFRs used to configure and access the sub-systems unique to the MCU. This allows the addition of new functionality while retaining compatibility with the MCS-51<sup>™</sup> instruction set. Table 9.2 lists the SFRs implemented in the CIP-51 System Controller.

The SFR registers are accessed anytime the direct addressing mode is used to access memory locations from 0x80 to 0xFF. SFRs with addresses ending in 0x0 or 0x8 (e.g. P0, TCON, SCON0, IE, etc.) are bit-addressable as well as byte-addressable. All other SFRs are byte-addressable only. Unoccupied addresses in the SFR space are reserved for future use. Accessing these areas will have an indeterminate effect and should be avoided. Refer to the corresponding pages of the datasheet, as indicated in Table 9.3, for a detailed description of each register.

F8	SPI0CN	PCA0L	PCA0H	PCA0CPL0	PCA0CPH0	PCA0CPL4	PCA0CPH4	VDM0CN
F0	В	POMDIN	P1MDIN	P2MDIN	P3MDIN	P4MDIN	EIP1	EIP2
E8	ADC0CN	PCA0CPL1	PCA0CPH1	PCA0CPL2	PCA0CPH2	PCA0CPL3	PCA0CPH3	RSTSRC
E0	ACC	XBR0	XBR1	XBR2	IT01CF	SMOD1	EIE1	EIE2
D8	PCA0CN	PCA0MD	PCA0CPM0	PCA0CPM1	PCA0CPM2	PCA0CPM3	PCA0CPM4	P3SKIP
D0	PSW	<b>REF0CN</b>	SCON1	SBUF1	P0SKIP	P1SKIP	P2SKIP	USB0XCN
C8	TMR2CN	REG0CN	TMR2RLL	TMR2RLH	TMR2L	TMR2H	-	-
C0	SMB0CN	SMB0CF	SMB0DAT	ADC0GTL	ADC0GTH	ADC0LTL	ADC0LTH	P4
B8	IP	CLKMUL	AMX0N	AMX0P	ADC0CF	ADC0L	ADC0H	-
B0	P3	OSCXCN	OSCICN	OSCICL	SBRLL1	SBRLH1	FLSCL	FLKEY
A8	IE	CLKSEL	EMI0CN	-	SBCON1	-	P4MDOUT	PFE0CN
A0	P2	SPI0CFG	SPI0CKR	SPI0DAT	POMDOUT	P1MDOUT	P2MDOUT	P3MDOUT
98	SCON0	SBUF0	CPT1CN	CPT0CN	CPT1MD	CPT0MD	CPT1MX	CPT0MX
90	P1	TMR3CN	TMR3RLL	TMR3RLH	TMR3L	TMR3H	USB0ADR	USB0DAT
88	TCON	TMOD	TL0	TL1	TH0	TH1	CKCON	PSCTL
80	P0	SP	DPL	DPH	EMI0TC	EMI0CF	OSCLCN	PCON
	0(8)	1(9)	2(A)	3(B)	4(C)	5(D)	6(E)	7(F)
	(bit addressable)							

#### Table 9.2. Special Function Register (SFR) Memory Map



## **Table 9.3. Special Function Registers**

Register	Address	Description	Page
ACC	0xE0	Accumulator	87
ADC0CF	0xBC	ADC0 Configuration	50
ADC0CN	0xE8	ADC0 Control	51
ADC0GTH	0xC4	ADC0 Greater-Than Compare High	52
ADC0GTL	0xC3	ADC0 Greater-Than Compare Low	52
ADC0H	0xBE	ADC0 High	50
ADC0L	0xBD	ADC0 Low	50
ADC0LTH	0xC6	ADC0 Less-Than Compare Word High	53
ADC0LTL	0xC5	ADC0 Less-Than Compare Word Low	53
AMX0N	0xBA	AMUX0 Negative Channel Select	49
AMX0P	0xBB	AMUX0 Positive Channel Select	48
В	0xF0	B Register	88
CKCON	0x8E	Clock Control	241
CLKMUL	0xB9	Clock Multiplier	138
CLKSEL	0xA9	Clock Select	140
CPT0CN	0x9B	Comparator0 Control	62
CPT0MD	0x9D	Comparator0 Mode Selection	64
CPT0MX	0x9F	Comparator0 MUX Selection	63
CPT1CN	0x9A	Comparator1 Control	65
CPT1MD	0x9C	Comparator1 Mode Selection	67
CPT1MX	0x9E	Comparator1 MUX Selection	66
DPH	0x83	Data Pointer High	86
DPL	0x82	Data Pointer Low	86
EIE1	0xE6	Extended Interrupt Enable 1	93
EIE2	0xE7	Extended Interrupt Enable 2	95
EIP1	0xF6	Extended Interrupt Priority 1	94
EIP2	0xF7	Extended Interrupt Priority 2	95
EMI0CN	0xAA	External Memory Interface Control	117
EMI0CF	0x85	External Memory Interface Configuration	118
EMIOTC	0x84	External Memory Interface Timing	123
FLKEY	0xB7	Flash Lock and Key	112
FLSCL	0xB6	Flash Scale	113
IE	0xA8	Interrupt Enable	91
IP	0xB8	Interrupt Priority	92
IT01CF	0xE4	INT0/INT1 Configuration	96
OSCICL	0xB3	Internal Oscillator Calibration	133
OSCICN	0xB2	Internal Oscillator Control	132
OSCLCN	0x86	Internal Low-Frequency Oscillator Control	134
OSCXCN	0xB1	External Oscillator Control	137
P0	0x80	Port 0 Latch	150
POMDIN	0xF1	Port 0 Input Mode Configuration	150
POMDOUT	0xA4	Port 0 Output Mode Configuration	151
POSKIP	0xD4	Port 0 Skip	151
P1	0x90	Port 1 Latch	152

SFRs are listed in alphabetical order. All undefined SFR locations are reserved.



### Table 9.3. Special Function Registers (Continued)

SFRs are listed in alphabetical order. All undefined SFR locations are reserved.

Register	Address	Description	Page
P1MDIN	0xF2	Port 1 Input Mode Configuration	152
P1MDOUT	0xA5	Port 1 Output Mode Configuration	152
P1SKIP	0xD5	Port 1 Skip	153
P2	0xA0	Port 2 Latch	153
P2MDIN	0xF3	Port 2 Input Mode Configuration	153
P2MDOUT	0xA6	Port 2 Output Mode Configuration	154
P2SKIP	0xD6	Port 2 Skip	154
P3	0xB0	Port 3 Latch	155
P3MDIN	0xF4	Port 3 Input Mode Configuration	155
P3MDOUT	0xA7	Port 3 Output Mode Configuration	155
P3SKIP	0xDF	Port 3Skip	156
P4	0xC7	Port 4 Latch	156
P4MDIN	0xF5	Port 4 Input Mode Configuration	157
P4MDOUT	0xAE	Port 4 Output Mode Configuration	157
PCA0CN	0xD8	PCA Control	266
PCA0CPH0	0xFC	PCA Capture 0 High	270
PCA0CPH1	0xEA	PCA Capture 1 High	270
PCA0CPH2	0xEC	PCA Capture 2 High	270
PCA0CPH3	0xEE	PCA Capture 3High	270
PCA0CPH4	0xFE	PCA Capture 4 High	270
PCA0CPL0	0xFB	PCA Capture 0 Low	269
PCA0CPL1	0xE9	PCA Capture 1 Low	269
PCA0CPL2	0xEB	PCA Capture 2 Low	269
PCA0CPL3	0xED	PCA Capture 3 Low	269
PCA0CPL4	0xFD	PCA Capture 4 Low	269
PCA0CPM0	0xDA	PCA Module 0 Mode Register	268
PCA0CPM1	0xDB	PCA Module 1 Mode Register	268
PCA0CPM2	0xDC	PCA Module 2 Mode Register	268
PCA0CPM3	0xDD	PCA Module 3 Mode Register	268
PCA0CPM4	0xDE	PCA Module 4 Mode Register	268
PCA0H	0xFA	PCA Counter High	269
PCA0L	0xF9	PCA Counter Low	269
PCA0MD	0xD9	PCA Mode	267
PCON	0x87	Power Control	98
PFE0CN	0xAF	Prefetch Engine Control	99
PSCTL	0x8F	Program Store R/W Control	112
PSW	0xD0	Program Status Word	87
REF0CN	0xD1	Voltage Reference Control	58
REG0CN	0xC9	Voltage Regulator Control	72
RSTSRC	0xEF	Reset Source Configuration/Status	105
SBCON1	0xAC	UART1 Baud Rate Generator Control	220
SBRLH1	0xB5	UART1 Baud Rate Generator High	221
SBRLL1	0xB4	UART1 Baud Rate Generator Low	221
SBUF1	0xD3	UART1 Data Buffer	220
SCON1	0xD2	UART1 Control	218



### Table 9.3. Special Function Registers (Continued)

Register	Address	Description	Page
SBUF0	0x99	UART0 Data Buffer	211
SCON0	0x98	UART0 Control	210
SMB0CF	0xC1	SMBus Configuration	194
SMB0CN	0xC0	SMBus Control	196
SMB0DAT	0xC2	SMBus Data	198
SMOD1	0xE5	UART1 Mode	219
SP	0x81	Stack Pointer	86
SPI0CFG	0xA1	SPI Configuration	229
SPI0CKR	0xA2	SPI Clock Rate Control	231
SPI0CN	0xF8	SPI Control	230
SPI0DAT	0xA3	SPI Data	231
TCON	0x88	Timer/Counter Control	239
TH0	0x8C	Timer/Counter 0 High	242
TH1	0x8D	Timer/Counter 1 High	242
TL0	0x8A	Timer/Counter 0 Low	242
TL1	0x8B	Timer/Counter 1 Low	242
TMOD	0x89	Timer/Counter Mode	240
TMR2CN	0xC8	Timer/Counter 2 Control	247
TMR2H	0xCD	Timer/Counter 2 High	248
TMR2L	0xCC	Timer/Counter 2 Low	248
TMR2RLH	0xCB	Timer/Counter 2 Reload High	248
TMR2RLL	0xCA	Timer/Counter 2 Reload Low	248
TMR3CN	0x91	Timer/Counter 3Control	253
TMR3H	0x95	Timer/Counter 3 High	254
TMR3L	0x94	Timer/Counter 3Low	254
TMR3RLH	0x93	Timer/Counter 3 Reload High	254
TMR3RLL	0x92	Timer/Counter 3 Reload Low	254
VDM0CN	0xFF	V <sub>DD</sub> Monitor Control	102
USB0ADR	0x96	USB0 Indirect Address Register	163
USB0DAT	0x97	USB0 Data Register	164
USB0XCN	0xD7	USB0 Transceiver Control	161
XBR0	0xE1	Port I/O Crossbar Control 0	148
XBR1	0xE2	Port I/O Crossbar Control 1	149
XBR2	0xE3	Port I/O Crossbar Control 2	149
All Other Ad	dresses	Reserved	

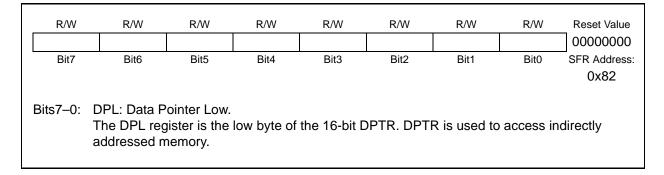
SFRs are listed in alphabetical order. All undefined SFR locations are reserved.



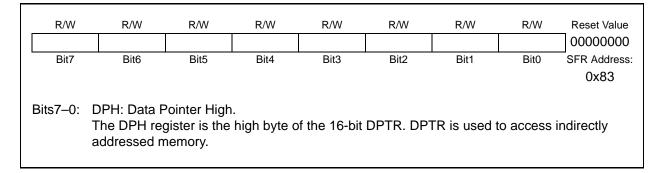
#### 9.2.7. Register Descriptions

Following are descriptions of SFRs related to the operation of the CIP-51 System Controller. Reserved bits should not be set to logic I. Future product versions may use these bits to implement new features in which case the reset value of the bit will be logic 0, selecting the feature's default state. Detailed descriptions of the remaining SFRs are included in the sections of the datasheet associated with their corresponding system function.

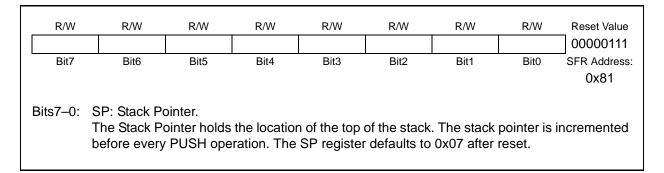
## SFR Definition 9.1. DPL: Data Pointer Low Byte



## SFR Definition 9.2. DPH: Data Pointer High Byte



## SFR Definition 9.3. SP: Stack Pointer





R/W	R/W	R/W	R/W	R/W	R/W	R/W	R	Reset Value
CY	AC	F0	RS1	RS0	OV	F1	PARITY	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
2	2.10	2.10		2.10	2.12		t addressable)	
						(2.	( 44410004510)	0/12 0
Bit7:	CY: Carry	Flag.						
	This bit is set when the last arithmetic operation resulted in a carry (addition) or a borrow							
			ared to logic 0					
Bit6:	AC: Auxilia	ary Carry Fl	ag					
			e last arithmeti					
			high order nibb	ole. It is clea	red to logi	c 0 by all othe	er arithmetic	operations.
Bit5:	F0: User F							
			ble, general p	urpose flag	for use ur	nder software	e control.	
Bits4–3:			Bank Select.	ادام برمما ا			-	
	These bits	select which	ch register bar	ik is used (	uting regis	ster accesse	s.	
	RS1	RS0 I	Register Banl	Addı	ess			
	0	0	0	0x00 -	0x07			
	0	1	1	0x08 -	0x0F			
	1	0	2	0x10 -	0x17			
	1	1	3	0x18 -	0x1F			
Dire								
Bit2:	OV: Overfl	•	der the followir		00000			
			SUBB instruct			ande overflov	N	
			esults in an over					
			uses a divide-					
						JL, and DIV	instructions	in all other
	The OV bit is cleared to 0 by the ADD, ADDC, SUBB, MUL, and DIV instructions in all other cases.							
Bit1:	F1: User F	lag 1.						
			ble, general p	urpose flag	for use ur	nder software	e control.	
Bit0:	PARITY: P							
	This bit is set to logic 1 if the sum of the eight bits in the accumulator is odd and cleared if the							
	sum is eve	n.						

## SFR Definition 9.5. ACC: Accumulator

R/W	R/W	R/W ACC.5	R/W ACC.4	R/W	R/W	R/W ACC.1	R/W	Reset Value
	ACC.6			ACC.3	ACC.2		ACC.0	0000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
						(bit	addressable)	0xE0
	ACC: Accum		mulator for	arithmetic o	operations.			



## SFR Definition 9.6. B: B Register

	R/W	Reset Value							
	B.7	B.6	B.5	B.4	B.3	B.2	B.1	B.0	00000000
	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
							(bit	t addressable	) 0xF0
Bits7–0: B: B Register. This register serves as a second accumulator for certain arithmetic operations.									

## 9.3. Interrupt Handler

The CIP-51 includes an extended interrupt system supporting multiple interrupt sources with two priority levels. The allocation of interrupt sources between on-chip peripherals and external inputs pins varies according to the specific version of the device. Each interrupt source has one or more associated interrupt-pending flag(s) located in an SFR. When a peripheral or external source meets a valid interrupt condition, the associated interrupt-pending flag is set to logic 1.

If interrupts are enabled for the source, an interrupt request is generated when the interrupt-pending flag is set. As soon as execution of the current instruction is complete, the CPU generates an LCALL to a predetermined address to begin execution of an interrupt service routine (ISR). Each ISR must end with an RETI instruction, which returns program execution to the next instruction that would have been executed if the interrupt request had not occurred. If interrupts are not enabled, the interrupt-pending flag is ignored by the hardware and program execution continues as normal. (The interrupt-pending flag is set to logic 1 regard-less of the interrupt's enable/disable state.)

Each interrupt source can be individually enabled or disabled through the use of an associated interrupt enable bit in an SFR (IE-EIE2). However, interrupts must first be globally enabled by setting the EA bit (IE.7) to logic 1 before the individual interrupt enables are recognized. Setting the EA bit to logic 0 disables all interrupt sources regardless of the individual interrupt-enable settings.

Some interrupt-pending flags are automatically cleared by the hardware when the CPU vectors to the ISR. However, most are not cleared by the hardware and must be cleared by software before returning from the ISR. If an interrupt-pending flag remains set after the CPU completes the return-from-interrupt (RETI) instruction, a new interrupt request will be generated immediately and the CPU will re-enter the ISR after the completion of the next instruction.

#### 9.3.1. MCU Interrupt Sources and Vectors

The MCU supports multiple interrupt sources. Software can simulate an interrupt by setting any interrupt-pending flag to logic 1. If interrupts are enabled for the flag, an interrupt request will be generated and the CPU will vector to the ISR address associated with the interrupt-pending flag. MCU interrupt sources, associated vector addresses, priority order and control bits are summarized in Table 9.4 on page 90. Refer to the datasheet section associated with a particular on-chip peripheral for information regarding valid interrupt conditions for the peripheral and the behavior of its interrupt-pending flag(s).

#### 9.3.2. External Interrupts

The INTO and INT1 external interrupt sources are configurable as active high or low, edge or level sensitive. The INOPL (INTO Polarity) and IN1PL (INT1 Polarity) bits in the IT01CF register select active high or active low; the IT0 and IT1 bits in TCON (Section "21.1. Timer 0 and Timer 1" on page 235) select level or edge sensitive. The following table lists the possible configurations.



IT0	IN0PL	INT0 Interrupt
1	0	Active low, edge sensitive
1	1	Active high, edge sensitive
0	0	Active low, level sensitive
0	1	Active high, level sensitive

IT1	IN1PL	INT1 Interrupt
1	0	Active low, edge sensitive
1	1	Active high, edge sensitive
0	0	Active low, level sensitive
0	1	Active high, level sensitive

INT0 and INT1 are assigned to Port pins as defined in the IT01CF register (see SFR Definition 9.13). Note that INT0 and INT0 Port pin assignments are independent of any Crossbar assignments. INT0 and INT1 will monitor their assigned Port pins without disturbing the peripheral that was assigned the Port pin via the Crossbar. To assign a Port pin only to INT0 and/or INT1, configure the Crossbar to skip the selected pin(s). This is accomplished by setting the associated bit in register XBR0 (see Section "15.1. Priority Crossbar Decoder" on page 144 for complete details on configuring the Crossbar). In the typical configuration, the external interrupt pin should be skipped in the crossbar and configured as open-drain with the pin latch set to '1'.

IE0 (TCON.1) and IE1 (TCON.3) serve as the interrupt-pending flags for the INT0 and INT1 external interrupts, respectively. If an INT0 or INT1 external interrupt is configured as edge-sensitive, the corresponding interrupt-pending flag is automatically cleared by the hardware when the CPU vectors to the ISR. When configured as level sensitive, the interrupt-pending flag remains logic 1 while the input is active as defined by the corresponding polarity bit (IN0PL or IN1PL); the flag remains logic 0 while the input is inactive. The external interrupt source must hold the input active until the interrupt request is recognized. It must then deactivate the interrupt request before execution of the ISR completes or another interrupt request will be generated.

#### 9.3.3. Interrupt Priorities

Each interrupt source can be individually programmed to one of two priority levels: low or high. A low priority interrupt service routine can be preempted by a high priority interrupt. A high priority interrupt cannot be preempted. Each interrupt has an associated interrupt priority bit in an SFR (IP or EIP2) used to configure its priority level. Low priority is the default. If two interrupts are recognized simultaneously, the interrupt with the higher priority is serviced first. If both interrupts have the same priority level, a fixed priority order is used to arbitrate, given in Table 9.4.

#### 9.3.4. Interrupt Latency

Interrupt response time depends on the state of the CPU when the interrupt occurs. Pending interrupts are sampled and priority decoded each system clock cycle. Therefore, the fastest possible response time is 6 system clock cycles: 1 clock cycle to detect the interrupt and 5 clock cycles to complete the LCALL to the ISR. If an interrupt is pending when a RETI is executed, a single instruction is executed before an LCALL is made to service the pending interrupt. Therefore, the maximum response time for an interrupt (when no other interrupt is currently being serviced or the new interrupt is of greater priority) occurs when the CPU is performing an RETI instruction followed by a DIV as the next instruction. In this case, the response time is 20 system clock cycles: 1 clock cycle to detect the interrupt, 6 clock cycles to execute the RETI, 8 clock cycles to complete the DIV instruction and 5 clock cycles to execute the LCALL to the ISR. If the CPU is executing an ISR for an interrupt with equal or higher priority, the new interrupt will not be serviced until the current ISR completes, including the RETI and following instruction.

Note that the CPU is stalled during Flash write/erase operations and USB FIFO MOVX accesses (see **Section "13.2. Accessing USB FIFO Space" on page 115**). Interrupt service latency will be increased for interrupts occurring while the CPU is stalled. The latency for these situations will be determined by the standard interrupt service procedure (as described above) and the amount of time the CPU is stalled.



Interrupt Source	Interrupt Vector	Priority Order	Pending Flag	Bit addressable?	Cleared by HW?	Enable Flag	Priority Control
Reset	0x0000	Тор	None	N/A	N/A	Always Enabled	Always Highest
External Interrupt 0 (INT0)	0x0003	0	IE0 (TCON.1)	Y	Y	EX0 (IE.0)	PX0 (IP.0)
Timer 0 Overflow	0x000B	1	TF0 (TCON.5)	Y	Y	ET0 (IE.1)	PT0 (IP.1)
External Interrupt 1 (INT1)	0x0013	2	IE1 (TCON.3)	Y	Y	EX1 (IE.2)	PX1 (IP.2)
Timer 1 Overflow	0x001B	3	TF1 (TCON.7)	Y	Y	ET1 (IE.3)	PT1 (IP.3)
UART0	0x0023	4	RI0 (SCON0.0) TI0 (SCON0.1)	Y	Ν	ES0 (IE.4)	PS0 (IP.4)
Timer 2 Overflow	0x002B	5	TF2H (TMR2CN.7) TF2L (TMR2CN.6)	Y	N	ET2 (IE.5)	PT2 (IP.5)
SPI0	0x0033	6	SPIF (SPI0CN.7) WCOL (SPI0CN.6) MODF (SPI0CN.5) RXOVRN (SPI0CN.4)	Y	N	ESPI0 (IE.6)	PSPI0 (IP.6)
SMB0	0x003B	7	SI (SMB0CN.0)	Y	N	ESMB0 (EIE1.0)	PSMB0 (EIP1.0)
USB0	0x0043	8	Special	Ν	Ν	EUSB0 (EIE1.1)	PUSB0 (EIP1.1)
ADC0 Window Compare	0x004B	9	ADOWINT (ADC0CN.3)	Y	Ν	EWADC0 (EIE1.2)	PWADC0 (EIP1.2)
ADC0 Conversion Complete	0x0053	10	AD0INT (ADC0CN.5)	Y	Ν	EADC0 (EIE1.3)	PADC0 (EIP1.3)
Programmable Counter Array	0x005B	11	CF (PCA0CN.7) CCFn (PCA0CN.n)	Y	N	EPCA0 (EIE1.4)	PPCA0 (EIP1.4)
Comparator0	0x0063	12	CP0FIF (CPT0CN.4) CP0RIF (CPT0CN.5)	Ν	Ν	ECP0 (EIE1.5)	PCP0 (EIP1.5)
Comparator1	0x006B	13	CP1FIF (CPT1CN.4) CP1RIF (CPT1CN.5)	Ν	Ν	ECP1 (EIE1.6)	PCP1 (EIP1.6)
Timer 3 Overflow	0x0073	14	TF3H (TMR3CN.7) TF3L (TMR3CN.6)	Ν	N	ET3 (EIE1.7)	PT3 (EIP1.7)
VBUS Level	0x007B	15	N/A	N/A	N/A	EVBUS (EIE2.0)	PVBUS (EIP2.0)
UART1	0x0083	16	RI1 (SCON1.0) TI1 (SCON1.1)	Ν	Ν	ES1 (EIE2.1)	PS1 (EIP2.1)

## Table 9.4. Interrupt Summary

#### 9.3.5. Interrupt Register Descriptions

The SFRs used to enable the interrupt sources and set their priority level are described below. Refer to the datasheet section associated with a particular on-chip peripheral for information regarding valid interrupt conditions for the peripheral and the behavior of its interrupt-pending flag(s).



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value				
EA	ESPI0	ET2	ES0	ET1	EX1	ET0	EX0	00000000				
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address				
						(bi	t addressable	e) 0xA8				
Bit7:	EA: Enable A	All Interrupt	S.									
	This bit globally enables/disables all interrupts. It overrides the individual interrupt mask set											
	tings.			•				•				
	0: Disable al	l interrupt s	ources.									
	1: Enable ea	ch interrup	t according	to its indivi	dual mask s	etting.						
Bit6:	ESPI0: Enab											
	This bit sets	the maskin	g of the SP	10 interrupt	s							
	0: Disable al		•	·								
	1: Enable int			ated by SPI	0.							
Bit5:	ET2: Enable		•	2								
	This bit sets the masking of the Timer 2 interrupt.											
	0: Disable Timer 2 interrupt.											
	1: Enable interrupt requests generated by the TF2L or TF2H flags.											
Bit4:	ES0: Enable UARTO Interrupt.											
	This bit sets the masking of the UART0 interrupt.											
	0: Disable UART0 interrupt.											
	1: Enable UART0 interrupt.											
Bit3:	ET1: Enable Timer 1 Interrupt.											
	This bit sets the masking of the Timer 1 interrupt.											
	0: Disable all Timer 1 interrupt.											
	1: Enable interrupt requests generated by the TF1 flag.											
Bit2:	EX1: Enable	External Ir	nterrupt 1.									
	This bit sets the masking of External Interrupt 1.											
	0: Disable external interrupt 1.											
	1: Enable int	errupt requ	lests genera	ated by the	INT1 input.							
Bit1:	ET0: Enable											
	This bit sets the masking of the Timer 0 interrupt.											
	0: Disable all	l Timer 0 in	terrupt.									
	1: Enable int			ated by the	TF0 flag.							
Bit0:	EX0: Enable	External Ir	nterrupt 0.									
	This bit sets			al Interrupt	0.							
	0: Disable ex											
	1: Enable int											



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value				
-	PSPI0	PT2	PS0	PT1	PX1	PT0	PX0	1000000				
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address				
						(bi	t addressabl	e) 0xB8				
Bit7:	UNUSED. R	,										
Bit6:	PSPI0: Seria	•		,	rupt Priority	Control.						
	This bit sets the priority of the SPI0 interrupt.											
	0: SPI0 inter	•										
<b>D</b>	1: SPI0 inter	•	• • •									
Bit5:	PT2: Timer 2	•										
	This bit sets			•	t.							
	0: Timer 2 int 1: Timer 2 int											
Bit4:			• •									
DI(4.	PS0: UART0 Interrupt Priority Control. This bit sets the priority of the UART0 interrupt.											
	0: UARTO interrupt set to low priority level.											
	1: UART0 interrupts set to high priority level.											
Bit3:	PT1: Timer 1 Interrupt Priority Control.											
	This bit sets the priority of the Timer 1 interrupt.											
	0: Timer 1 interrupt set to low priority level.											
	1: Timer 1 interrupts set to high priority level.											
Bit2:	PX1: Externa	al Interrupt	1 Priority Co	ontrol.								
	This bit sets	the priority	of the Exter	nal Interrup	ot 1 interrup	t.						
	0: External Ir	•										
	1: External Ir											
Bit1:	PT0: Timer 0 Interrupt Priority Control.											
	This bit sets the priority of the Timer 0 interrupt.											
	0: Timer 0 interrupt set to low priority level.											
D'10	1: Timer 0 interrupt set to high priority level. PX0: External Interrupt 0 Priority Control.											
Bit0:												
	This bit sets				ot 0 interrup	τ.						
	0: External Ir 1: External Ir	•										
	I. EXTERIAL I	iterrupt 0 s	ser to nigh pi	ionty level.								

## SFR Definition 9.8. IP: Interrupt Priority



SFR Definition 9.9. EIE1: Extended Interrupt Enable 1
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R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value				
ET3	ECP1	ECP0	EPCA0	EADC0	EWADC0	EUSB0	ESMB0	00000000				
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Addres 0xE6				
Bit7:	ET3: Enable	Timer 3 In	terrupt.									
	This bit sets	the maskin	g of the Tin	ner 3 interru	ipt.							
	0: Disable Ti											
	1: Enable int				TF3L or TF3	3H flags.						
Bit6:	ECP1: Enab	le Compara	ator1 (CP1)	Interrupt.								
	This bit sets the masking of the CP1 interrupt.											
	0: Disable C	P1 interrup	ts.									
	1: Enable int	terrupt requ	iests genera	ated by the	CP1RIF or (	CP1FIF flag	js.					
Bit5:	ECP0: Enab	le Compara	ator0 (CP0)	Interrupt.								
	This bit sets	the maskin	g of the CP	0 interrupt.								
	0: Disable C	0: Disable CP0 interrupts.										
	1: Enable int	terrupt requ	lests genera	ated by the	CP0RIF or (	CP0FIF flag	js.					
Bit4:	EPCA0: Ena	able Progra	mmable Co	unter Array	(PCA0) Inte	errupt.						
	This bit sets the masking of the PCA0 interrupts.											
	0: Disable all PCA0 interrupts.											
	1: Enable int	terrupt requ	lests genera	ated by PCA	۹0.							
Bit3:	EADC0: Ena	able ADC0	Conversion	Complete I	nterrupt.							
	This bit sets	the maskin	g of the AD	C0 Conver	sion Comple	ete interrupt	t.					
	0: Disable A	DC0 Conve	ersion Comp	olete interru	pt.							
	1: Enable int	terrupt requ	iests genera	ated by the	AD0INT flag	J.						
Bit2:	EWADC0: E	nable Wind	low Compa	rison ADC0	Interrupt.							
	This bit sets	the maskin	g of ADC0	Window Co	mparison in	terrupt.						
	0: Disable A	DC0 Windo	w Compari	son interrup	ot.							
	1: Enable interrupt requests generated by ADC0 Window Compare flag (AD0WINT).											
Bit1:	EUSB0: Ena	able USB0 I	nterrupt.									
	This bit sets the masking of the USB0 interrupt.											
	0: Disable all USB0 interrupts.											
	1: Enable int	terrupt requ	lests genera	ated by USE	30.							
Bit0:	ESMB0: Ena	able SMBus	s (SMB0) In	terrupt.								
	This bit sets	the maskin	g of the SN	1B0 interrup	ot.							
	0: Disable al											
	1: Enable int											



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value				
PT3	PCP1	PCP0	PPCA0	PADC0	PWADC0	PUSB0	PSMB0	0000000				
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Addres 0xF6				
Bit7:	PT3: Timer 3											
	This bit sets				ot.							
	0: Timer 3 in	•										
	1: Timer 3 in	•	• •									
Bit6:	PCP1: Com				ontrol.							
	This bit sets											
	0: CP1 interr											
	1: CP1 interr											
Bit5:	PCP0: Com	· ·	, i		ontrol.							
	This bit sets											
	0: CP0 interr											
	1: CP0 interrupt set to high priority level.											
Bit4:	PPCA0: Programmable Counter Array (PCA0) Interrupt Priority Control.											
	This bit sets the priority of the PCA0 interrupt.											
	0: PCA0 interrupt set to low priority level.											
	1: PCA0 inte											
Bit3:	PADC0 ADC											
	This bit sets											
	0: ADC0 Conversion Complete interrupt set to low priority level.											
	1: ADC0 Co		•		• • •							
Bit2:	PWADC0: A					ontrol.						
	This bit sets	the priority	of the ADC	0 Window	interrupt.							
	0: ADC0 Window interrupt set to low priority level.											
	1: ADC0 Wir		•	• • •	evel.							
Bit1:	PUSB0: USE	30 Interrup	t Priority Co	ntrol.								
	This bit sets the priority of the USB0 interrupt.											
	0: USB0 interrupt set to low priority level.											
	1: USB0 inte	errupt set to	high priorit	y level.								
Bit0:	PSMB0: SM	Bus (SMB0	)) Interrupt I	Priority Cor	ntrol.							
	This bit sets	the priority	of the SMB	0 interrupt.								
	0: SMB0 inte											
	1: SMB0 inte		المتعرفة والمتعاد									

## SFR Definition 9.10. EIP1: Extended Interrupt Priority 1



SFR Definition 9.11. EIE2: Extended Interrupt Enable 2
--

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
-	-	-	-	-	-	ES1	EVBUS	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
								0xE7
Bits7–2: Bit1: Bit0:	UNUSED. Re ES1: Enable This bit sets 0: Disable U/ 1: Enable UA EVBUS: Ena This bit sets 0: Disable all 1: Enable inte	UART1 In the maskin ART1 intern RT1 interr ble VBUS the maskin VBUS inte	errupt. g of the UA rupt. upt. Level Interru g of the VB errupts.	RT1 interru upt. US interrup	pt. t.	se.		

## SFR Definition 9.12. EIP2: Extended Interrupt Priority 2

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
-	-	-	-	-	-	PS1	PVBUS	0000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
								0xF7
Bits7–2: Bit1: Bit0:	UNUSED. R PS1: UART1 This bit sets 0: UART1 int 1: UART1 int PVBUS: VBU This bit sets 0: VBUS inte 1: VBUS inte	Interrupt F the priority terrupt set t terrupts set JS Level In the priority errupt set to	Priority Cont of the UAR o low priorit to high prior terrupt Prio of the VBU low priority	rol. T1 interrupt ty level. prity level. rity Control. S interrupt. y level.				



## SFR Definition 9.13. IT01CF: INT0/INT1 Configuration

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
IN1PL	IN1SL2	IN1SL1	IN1SL0	INOPL	IN0SL2	IN0SL1	INOSLO	00000001
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
								0xE4
Note: Re	fer to SFR Def	inition 21.	1 for INT0/1	edae- or le	vel-sensitiv	e interrupt s	selection.	-
				eage of le		o intorrapt.		
Bit7:	IN1PL: INT1	Polarity						
	0: INT1 input		ow.					
	1: INT1 input							
Bits6–4:	IN1SL2-0: IN	IT1 Port Pi	n Selection	Bits				
	These bits se	lect which	Port pin is	assigned to	INT1. Note	that this pi	n assignme	ent is inde-
	pendent of th							
	peripheral that							
	assign the Po					the selected	d pin (accor	nplished by
	setting to '1' t	he corresp	bonding bit i	in register P	POSKIP).			
	IN1SL2-0		1 Port Pin					
	000		P0.0					
	001		P0.1					
	010		P0.2					
	011		P0.3					
	100		P0.4					
	101		P0.5					
	110		P0.6					
	111		P0.7					
Bit3:	INOPL: INTO	Polaritv						
	0: INTO interr		ve low.					
	1: INT0 interr	upt is activ	ve high.					
Bits2–0:	INT0SL2-0: I							
	These bits se							
	pendent of th							
	peripheral that		•	•				
	assign the Po					the selected	a pin (accor	npiisned by
	setting to '1' t	ne conesp	bonding bit i	in register F	103KIP).			
	IN0SL2-0	INT	0 Port Pin					
	000		P0.0					
	001		P0.1					
	010		P0.2					
	011		P0.3					
	100		P0.4					
	101		P0.5					
	110		P0.6					
	111		P0.7					
	L							



#### 9.4. Power Management Modes

The CIP-51 core has two software programmable power management modes: Idle and Stop. Idle mode halts the CPU while leaving the peripherals and clocks active. In Stop mode, the CPU is halted, all interrupts, are inactive, and the internal oscillator is stopped (analog peripherals remain in their selected states; the external oscillator is not affected). Since clocks are running in Idle mode, power consumption is dependent upon the system clock frequency and the number of peripherals left in active mode before entering Idle. Stop mode consumes the least power. Figure 1.15 describes the Power Control Register (PCON) used to control the CIP-51's power management modes.

Although the CIP-51 has Idle and Stop modes built in (as with any standard 8051 architecture), power management of the entire MCU is better accomplished through system clock and individual peripheral management. Each analog peripheral can be disabled when not in use and placed in low power mode. Digital peripherals, such as timers or serial buses, draw little power when they are not in use. Turning off the oscillators lowers power consumption considerably; however a reset is required to restart the MCU.

The internal oscillator can be placed in Suspend mode (see **Section "14. Oscillators" on page 131**). In Suspend mode, the internal oscillator is stopped until a non-idle USB event is detected, or the VBUS input signal matches the polarity selected by the VBPOL bit in register REGOCN (SFR Definition 8.1).

#### 9.4.1. Idle Mode

Setting the Idle Mode Select bit (PCON.0) causes the CIP-51 to halt the CPU and enter Idle mode as soon as the instruction that sets the bit completes execution. All internal registers and memory maintain their original data. All analog and digital peripherals can remain active during Idle mode.

Idle mode is terminated when an enabled interrupt is asserted or a reset occurs. The assertion of an enabled interrupt will cause the Idle Mode Selection bit (PCON.0) to be cleared and the CPU to resume operation. The pending interrupt will be serviced and the next instruction to be executed after the return from interrupt (RETI) will be the instruction immediately following the one that set the Idle Mode Select bit. If Idle mode is terminated by an internal or external reset, the CIP-51 performs a normal reset sequence and begins program execution at address 0x0000.

If enabled, the Watchdog Timer (WDT) will eventually cause an internal watchdog reset and thereby terminate the Idle mode. This feature protects the system from an unintended permanent shutdown in the event of an inadvertent write to the PCON register. If this behavior is not desired, the WDT may be disabled by software prior to entering the Idle mode if the WDT was initially configured to allow this operation. This provides the opportunity for additional power savings, allowing the system to remain in the Idle mode indefinitely, waiting for an external stimulus to wake up the system. Refer to **Section "11.6. PCA Watchdog Timer Reset" on page 103** for more information on the use and configuration of the WDT.

#### 9.4.2. Stop Mode

Setting the Stop Mode Select bit (PCON.1) causes the CIP-51 to enter Stop mode as soon as the instruction that sets the bit completes execution. In Stop mode the internal oscillator, CPU, and all digital peripherals are stopped; the state of the external oscillator circuit is not affected. Each analog peripheral (including the external oscillator circuit) may be shut down individually prior to entering Stop Mode. Stop mode can only be terminated by an internal or external reset. On reset, the CIP-51 performs the normal reset sequence and begins program execution at address 0x0000.

If enabled, the Missing Clock Detector will cause an internal reset and thereby terminate the Stop mode. The Missing Clock Detector should be disabled if the CPU is to be put to in STOP mode for longer than the MCD timeout of 100 µsec.



R/W GF5	R/W GF4	R/W GF3	R/W GF2	R/W GF1	R/W GF0	R/W STOP	R/W	Reset Value				
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0x87				
Bits7–2:	Bits7–2: GF5–GF0: General Purpose Flags 5–0. These are general purpose flags for use under software control.											
Bit1:	STOP: Stop Mode Select. Setting this bit will place the CIP-51 in Stop mode. This bit will always be read as 0.											
Bit0:	<ol> <li>CPU goes into Stop mode (internal oscillator stopped).</li> <li>Bit0: IDLE: Idle Mode Select.</li> <li>Setting this bit will place the CIP-51 in Idle mode. This bit will always be read as 0.</li> <li>CPU goes into Idle mode. (Shuts off clock to CPU, but clock to Timers, Interrupts, Serial Ports, and Analog Peripherals are still active.)</li> </ol>											

## SFR Definition 9.14. PCON: Power Control



# 10. Prefetch Engine

The 48 MHz versions of the C8051F34x family of devices incorporate a 2-byte prefetch engine. Because the access time of the FLASH memory is 40 ns, and the minimum instruction time is roughly 20 ns, the prefetch engine is necessary for full-speed code execution. Instructions are read from FLASH memory two bytes at a time by the prefetch engine, and given to the CIP-51 processor core to execute. When running linear code (code without any jumps or branches), the prefetch engine allows instructions to be executed at full speed. When a code branch occurs, the processor may be stalled for up to two clock cycles while the next set of code bytes is retrieved from FLASH memory. The FLRT bit (FLSCL.4) determines how many clock cycles are used to read each set of two code bytes from FLASH. When operating from a system clock of 25 MHz or less, the FLRT bit should be set to '0' so that the prefetch engine takes only one clock cycle for each read. When operating with a system clock of greater than 25 MHz (up to 48 MHz), the FLRT bit should be set to '1', so that each prefetch code read lasts for two clock cycles.

R	R	R/W	R	R	R	R	R/W	Reset Value
		PFEN					FLBWE	00100000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	-
							SFR Address:	0xAF
Bits 7–6: Bit 5:	Unused. Rea PFEN: Prefe This bit enab 0: Prefetch e 1: Prefetch e	tch Enable bles the prei engine is dis	fetch engine abled.					
Bits 4–1: Bit 0:	Unused. Rea FLBWE: FLA This bit allow 0: Each byte 1: FLASH by	ASH Block N vs block wri of a softwa	Write Enabl tes to FLAS are FLASH	e. SH memory write is writt				

## SFR Definition 10.1. PFE0CN: Prefetch Engine Control



## 11. Reset Sources

Reset circuitry allows the controller to be easily placed in a predefined default condition. On entry to this reset state, the following occur:

- CIP-51 halts program execution
- Special Function Registers (SFRs) are initialized to their defined reset values
- External Port pins are forced to a known state
- Interrupts and timers are disabled.

All SFRs are reset to the predefined values noted in the SFR detailed descriptions. The contents of internal data memory are unaffected during a reset; any previously stored data is preserved. However, since the stack pointer SFR is reset, the stack is effectively lost even though the data on the stack is not altered.

The Port I/O latches are reset to 0xFF (all logic ones) in open-drain mode. Weak pull-ups are enabled during and after the reset. For  $V_{DD}$  Monitor and Power-On Resets, the RST pin is driven low until the device exits the reset state.

On exit from the reset state, the program counter (PC) is reset, and the system clock defaults to the internal oscillator. Refer to Section "14. Oscillators" on page 131 for information on selecting and configuring the system clock source. The Watchdog Timer is enabled with the system clock divided by 12 as its clock source (Section "22.3. Watchdog Timer Mode" on page 264 details the use of the Watchdog Timer). Program execution begins at location 0x0000.

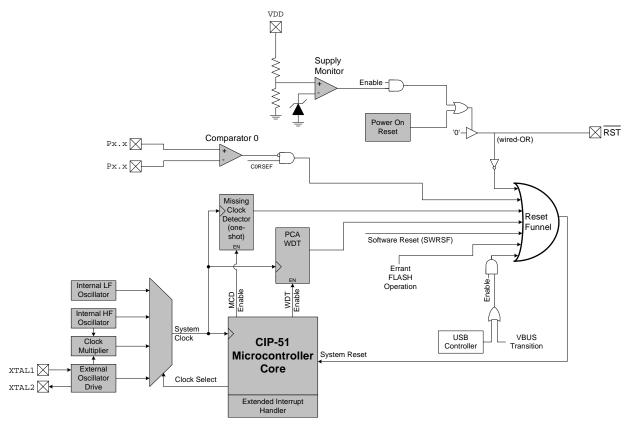


Figure 11.1. Reset Sources



### 11.1. Power-On Reset

During power-up, the device is held in a reset state and the  $\overline{RST}$  pin is driven low until V<sub>DD</sub> settles above V<sub>RST</sub>. A Power-On Reset delay (T<sub>PORDelay</sub>) occurs before the device is released from reset; this delay is typically less than 0.3 ms. Figure 11.2. plots the power-on and V<sub>DD</sub> monitor reset timing.

On exit from a power-on reset, the PORSF flag (RSTSRC.1) is set by hardware to logic 1. When PORSF is set, all of the other reset flags in the RSTSRC Register are indeterminate (PORSF is cleared by all other resets). Since all resets cause program execution to begin at the same location (0x0000) software can read the PORSF flag to determine if a power-up was the cause of reset. The content of internal data memory should be assumed to be undefined after a power-on reset. The V<sub>DD</sub> monitor is enabled following a power-on reset.

Software can force a power-on reset by writing '1' to the PINRSF bit in register RSTSRC.

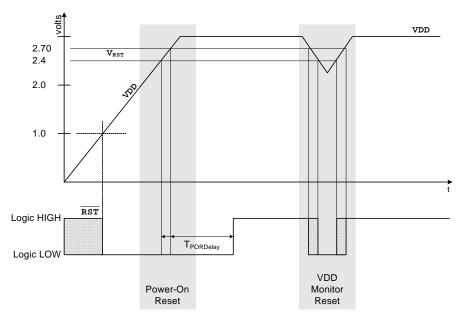


Figure 11.2. Power-On and  $V_{\text{DD}}$  Monitor Reset Timing



## 11.2. Power-Fail Reset / V<sub>DD</sub> Monitor

When a power-down transition or power irregularity causes  $V_{DD}$  to drop below  $V_{RST}$ , the power supply monitor will drive the  $\overline{RST}$  pin low and hold the CIP-51 in a reset state (see Figure 11.2). When  $V_{DD}$  returns to a level above  $V_{RST}$ , the CIP-51 will be released from the reset state. Note that even though internal data memory contents are not altered by the power-fail reset, it is impossible to determine if  $V_{DD}$  dropped below the level required for data retention. If the PORSF flag reads '1', the data may no longer be valid. The  $V_{DD}$  monitor is enabled after power-on resets; however its defined state (enabled/disabled) is not altered by any other reset source. For example, if the  $V_{DD}$  monitor is enabled and a software reset is performed, the  $V_{DD}$  monitor will still be enabled after the reset. It is strongly recommended that the  $V_{DD}$  monitor be left enabled at all times for any system that contains code to write to Flash memory.

**Important Note:** The V<sub>DD</sub> monitor must be enabled before it is selected as a reset source. Selecting the V<sub>DD</sub> monitor as a reset source before it is enabled and stabilized may cause a system reset. In applications where this reset is undesirable, a delay can be implemented between enabling the V<sub>DD</sub> monitor and selecting it as a reset source. The procedure for configuring the V<sub>DD</sub> monitor as a reset source is shown below:

- Step 1. Enable the  $V_{DD}$  monitor (VDM0CN.7 = '1').
- Step 2. If desired, wait for the V<sub>DD</sub> monitor to stabilize (see Table 11.1 for the V<sub>DD</sub> Monitor turn-on time).
- Step 3. Select the  $V_{DD}$  monitor as a reset source (RSTSRC.1 = '1').

See Figure 11.2 for  $V_{DD}$  monitor timing. See Table 11.1 for complete electrical characteristics of the  $V_{DD}$  monitor.

R/W	R	R	R	R	R	R	R	Reset Value
VDMEN	VDDSTAT	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Variable
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
								0xFF
Bit7:	VDMEN: V <sub>DI</sub>	o Monitor E	nable.					
	This bit turns	-		it on/off. The	e V חס Monit	or cannot q	enerate sv	stem resets
	until it is also					-	-	
	Monitor must			-				
	V <sub>DD</sub> monitor							-
	See Table 11					-	-	
	lowing all PC					00		
	0: V <sub>DD</sub> Monit		ł.					
	1: V <sub>DD</sub> Monit							
Bit6:	V <sub>DD</sub> STAT: V <sub>I</sub>							
	This bit indic	_	rrent power	supply stat	us (Vnn Mo	nitor outpu	t).	
	0: V <sub>DD</sub> is at c		•					
	1: $V_{DD}$ is abo							
Bits5–0:	Reserved. R							
2.00 0.								

## SFR Definition 11.1. VDM0CN: V<sub>DD</sub> Monitor Control



## 11.3. External Reset

The external RST pin provides a means for external circuitry to force the device into a reset state. Asserting an active-low signal on the RST pin generates a reset; an external pull-up and/or decoupling of the RST pin may be necessary to avoid erroneous noise-induced resets. See Table 11.1 for complete RST pin specifications. The PINRSF flag (RSTSRC.0) is set on exit from an external reset.

#### 11.4. Missing Clock Detector Reset

The Missing Clock Detector (MCD) is a one-shot circuit that is triggered by the system clock. If more than 100  $\mu$ s pass between rising edges on the system clock, the one-shot will time out and generate a reset. After a MCD reset, the MCDRSF flag (RSTSRC.2) will read '1', signifying the MCD as the reset source; otherwise, this bit reads '0'. Writing a '1' to the MCDRSF bit enables the Missing Clock Detector; writing a '0' disables it. The state of the RST pin is unaffected by this reset.

#### 11.5. Comparator0 Reset

Comparator0 can be configured as a reset source by writing a '1' to the CORSEF flag (RSTSRC.5). Comparator0 should be enabled and allowed to settle prior to writing to CORSEF to prevent any turn-on chatter on the output from generating an unwanted reset. The Comparator0 reset is active-low: if the non-inverting input voltage (on CP0+) is less than the inverting input voltage (on CP0-), a system reset is generated. After a Comparator0 reset, the CORSEF flag (RSTSRC.5) will read '1' signifying Comparator0 as the reset source; otherwise, this bit reads '0'. The state of the RST pin is unaffected by this reset.

#### 11.6. PCA Watchdog Timer Reset

The programmable Watchdog Timer (WDT) function of the Programmable Counter Array (PCA) can be used to prevent software from running out of control during a system malfunction. The PCA WDT function can be enabled or disabled by software as described in Section "22.3. Watchdog Timer Mode" on page 264; the WDT is enabled and clocked by SYSCLK / 12 following any reset. If a system malfunction prevents user software from updating the WDT, a reset is generated and the WDTRSF bit (RSTSRC.5) is set to '1'. The state of the RST pin is unaffected by this reset.

## 11.7. Flash Error Reset

If a Flash read/write/erase or program read targets an illegal address, a system reset is generated. This may occur due to any of the following:

- A Flash write or erase is attempted above user code space. This occurs when PSWE is set to "1", and a MOVX write operation is attempted above address 0x7FFF (32 kB Flash devices) or 0xFBFF (64 kB Flash devices).
- A Flash read is attempted above user code space. This occurs when a MOVC operation is attempted above address 0x7FFF (32 kB Flash devices) or 0xFBFF (64 kB Flash devices).
- A Program read is attempted above user code space. This occurs when user code attempts to branch to an address above 0x7FFF (32 kB Flash devices) or 0xFBFF (64 kB Flash devices).
- A Flash read, write or erase attempt is restricted due to a Flash security setting (see Section "12.3. Security Options" on page 109).
- A Flash Write or Erase is attempted when the V<sub>DD</sub> monitor is not enabled.

The FERROR bit (RSTSRC.6) is set following a Flash error reset. The state of the  $\overline{RST}$  pin is unaffected by this reset.



#### 11.8. Software Reset

Software may force a reset by writing a '1' to the SWRSF bit (RSTSRC.4). The SWRSF bit will read '1' following a software forced reset. The state of the RST pin is unaffected by this reset.

#### 11.9. USB Reset

Writing '1' to the USBRSF bit in register RSTSRC selects USB0 as a reset source. With USB0 selected as a reset source, a system reset will be generated when either of the following occur:

- RESET signaling is detected on the USB network. The USB Function Controller (USB0) must be enabled for RESET signaling to be detected. See Section "16. Universal Serial Bus Controller (USB0)" on page 159 for information on the USB Function Controller.
- The voltage on the VBUS pin matches the polarity selected by the VBPOL bit in register REGOCN. See Section "8. Voltage Regulator (REG0)" on page 69 for details on the VBUS detection circuit.

The USBRSF bit will read '1' following a USB reset. The state of the  $\overline{RST}$  pin is unaffected by this reset.



R/W	R	R/W	R/W	R	R/W	R/W	R	Reset Value		
USBRS	F FERROR	CORSEF	SWRSF	WDTRSF	MCDRSF	PORSF	PINRSF	Variable		
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:		
								0xEF		
								•••=•		
Bit7:	USBRSF: U	SB Reset F	lag							
2	0: Read: Las		-	reset: Writ	e: USB rese	ets disabled	ł.			
	1: Read: Las									
Bit6:	FERROR: F									
	0: Source of			lash read/w	rite/erase er	ror.				
	1: Source of	last reset w	as a Flash	read/write/	erase error.					
Bit5:	CORSEF: Co	omparator0	Reset Ena	ble and Flag	<b>j</b> .					
	0: Read: So	•				: Compara	tor0 is not a	a reset		
	source.			-		-				
	1: Read: So	urce of last	reset was	Comparator	0; Write: Co	omparator0	is a reset s	source		
	(active-low).									
Bit4:	SWRSF: So	ftware Rese	et Force an	d Flag.						
	0: Read: So									
	1: Read: So				RSF bit; <b>Wr</b> i	ite: Forces	a system r	eset.		
Bit3:	WDTRSF: W	-		-						
	0: Source of									
	1: Source of									
Bit2:	MCDRSF: M									
	0: <b>Read:</b> So			not a Missin	g Clock Det	ector timec	out; Write: I	Vissing		
	Clock Detec				la als Data at					
	1: Read: So			-				sing Clock		
Bit1:	Detector ena			-			sied.			
DILI.	PORSF: Po	_	-	-	\\ <i>\\</i> :::::::::::::::::::::::::::::::::		- / -    4 -	$\frac{1}{2}$		
	This bit is se	-	-		-					
	monitor as a			-		_	-			
	and stabiliz									
	0: <b>Read:</b> Las	st reset was	not a pow	er-on or V <sub>DI</sub>	<sub>D</sub> monitor re	set; Write:	V <sub>DD</sub> monit	or is not a		
	reset source									
	1: Read: Las	st reset was	a power-or	n or V <sub>DD</sub> mo	nitor reset; a	all other res	et flags ind	eterminate;		
	Write: V <sub>DD</sub> r	monitor is a	reset source	ce.						
Bit0:	PINRSF: HV	V Pin Reset	Flag.							
	0: Source of	last reset w	as <u>not RS</u>	T pin.						
	1: Source of	last reset w	as RST pi	า.						
						_		. ,		
	r bits that ac									
	read), read-modify-write instructions read and modify the source enable only. This applies to bits: USBRSF, C0RSEF, SWRSF, MCDRSF, PORSF.									
DITS: US	SKSF, CURSE	EF, SWRSF,	MCDRSF,	PURSF.						

## SFR Definition 11.2. RSTSRC: Reset Source



## **Table 11.1. Reset Electrical Characteristics**

#### -40 to +85 °C unless otherwise specified.

Parameter	Conditions	Min	Тур	Max	Units
RST Output Low Voltage	$I_{OL}$ = 8.5 mA, $V_{DD}$ = 2.7 to 3.6 V			0.6	V
RST Input High Voltage		$0.7 \times V_{DD}$			V
RST Input Low Voltage				$0.3 \times V_{DD}$	
RST Input Pull-Up Current	RST = 0.0 V		25	40	μA
V <sub>DD</sub> POR Threshold (V <sub>RST</sub> )		2.40	2.55	2.70	V
Missing Clock Detector Tim- eout	Time from last system clock ris- ing edge to reset initiation	100	220	500	μs
Reset Time Delay	Delay between release of any reset source and code execution at location 0x0000	5.0			μs
Minimum RST Low Time to Generate a System Reset		15			μs
V <sub>DD</sub> Monitor Turn-on Time		100			μs
V <sub>DD</sub> Monitor Supply Current			20	50	μA



## 12. Flash Memory

On-chip, re-programmable Flash memory is included for program code and non-volatile data storage. The Flash memory can be programmed in-system through the C2 interface or by software using the MOVX instruction. Once cleared to logic 0, a Flash bit must be erased to set it back to logic 1. Flash bytes would typically be erased (set to 0xFF) before being reprogrammed. The write and erase operations are automatically timed by hardware for proper execution; data polling to determine the end of the write/erase operation is not required. Code execution is stalled during a Flash write/erase operation. Refer to Table 12.1 for complete Flash memory electrical characteristics.

## 12.1. Programming The Flash Memory

The simplest means of programming the Flash memory is through the C2 interface using programming tools provided by Silicon Labs or a third party vendor. This is the only means for programming a non-initialized device. For details on the C2 commands to program Flash memory, see **Section "23. C2 Interface"** on page 271.

To ensure the integrity of Flash contents, it is strongly recommended that the  $V_{DD}$  monitor be left enabled in any system which writes or erases Flash memory from code. It is also crucial to ensure that the FLRT bit in register FLSCL be set to '1' if a clock speed higher than 25 MHz is being used for the device.

#### 12.1.1. Flash Lock and Key Functions

Flash writes and erases by user software are protected with a lock and key function. The Flash Lock and Key Register (FLKEY) must be written with the correct key codes, in sequence, before Flash operations may be performed. The key codes are: 0xA5, 0xF1. The timing does not matter, but the codes must be written in order. If the key codes are written out of order, or the wrong codes are written, Flash writes and erases will be disabled until the next system reset. Flash writes and erases will also be disabled if a Flash write or erase is attempted before the key codes have been written properly. The Flash lock resets after each write or erase; the key codes must be written again before a following Flash operation can be performed. The FLKEY register is detailed in SFR Definition 12.2.

#### 12.1.2. Flash Erase Procedure

The Flash memory can be programmed by software using the MOVX write instruction with the address and data byte to be programmed provided as normal operands. Before writing to Flash memory using MOVX, Flash write operations must be enabled by: (1) Writing the Flash key codes in sequence to the Flash Lock register (FLKEY); and (2) Setting the PSWE Program Store Write Enable bit (PSCTL.0) to logic 1 (this directs the MOVX writes to target Flash memory). The PSWE bit remains set until cleared by software.

A write to Flash memory can clear bits to logic 0 but cannot set them; only an erase operation can set bits to logic 1 in Flash. A byte location to be programmed must be erased before a new value is written. The Flash memory is organized in 512-byte pages. The erase operation applies to an entire page (setting all bytes in the page to 0xFF). To erase an entire 512-byte page, perform the following steps:

- Step 1. Disable interrupts (recommended).
- Step 2. Write the first key code to FLKEY: 0xA5.
- Step 3. Write the second key code to FLKEY: 0xF1.
- Step 4. Set the PSEE bit (register PSCTL).
- Step 5. Set the PSWE bit (register PSCTL).
- Step 6. Using the MOVX instruction, write a data byte to any location within the 512-byte page to be erased.
- Step 7. Clear the PSWE bit (register PSCTL).
- Step 8. Clear the PSEE bit (register PSCTI).



#### 12.1.3. Flash Write Procedure

Bytes in Flash memory can be written one byte at a time, or in groups of two. The FLBWE bit in register PFE0CN (SFR Definition 10.1) controls whether a single byte or a block of two bytes is written to Flash during a write operation. When FLBWE is cleared to '0', the Flash will be written one byte at a time. When FLBWE is set to '1', the Flash will be written in two-byte blocks. Block writes are performed in the same amount of time as single-byte writes, which can save time when storing large amounts of data to Flash memory.During a single-byte write to Flash, bytes are written individually, and a Flash write will be performed after each MOVX write instruction. The recommended procedure for writing Flash in single bytes is:

- Step 1. Disable interrupts.
- Step 2. Clear the FLBWE bit (register PFE0CN) to select single-byte write mode.
- Step 3. Set the PSWE bit (register PSCTL).
- Step 4. Clear the PSEE bit (register PSCTL).
- Step 5. Write the first key code to FLKEY: 0xA5.
- Step 6. Write the second key code to FLKEY: 0xF1.
- Step 7. Using the MOVX instruction, write a single data byte to the desired location within the 512-byte sector.
- Step 8. Clear the PSWE bit.
- Step 9. Re-enable interrupts.

Steps 5-7 must be repeated for each byte to be written.

For block Flash writes, the Flash write procedure is only performed after the last byte of each block is written with the MOVX write instruction. A Flash write block is two bytes long, from even addresses to odd addresses. Writes must be performed sequentially (i.e. addresses ending in 0b and 1b must be written in order). The Flash write will be performed following the MOVX write that targets the address ending in 1b. If a byte in the block does not need to be updated in Flash, it should be written to 0xFF. The recommended procedure for writing Flash in blocks is:

- Step 1. Disable interrupts.
- Step 2. Set the FLBWE bit (register PFE0CN) to select block write mode.
- Step 3. Set the PSWE bit (register PSCTL).
- Step 4. Clear the PSEE bit (register PSCTL).
- Step 5. Write the first key code to FLKEY: 0xA5.
- Step 6. Write the second key code to FLKEY: 0xF1.
- Step 7. Using the MOVX instruction, write the first data byte to the even block location (ending in 0b).
- Step 8. Write the first key code to FLKEY: 0xA5.
- Step 9. Write the second key code to FLKEY: 0xF1.
- Step 10. Using the MOVX instruction, write the second data byte to the odd block location (ending in 1b).
- Step 11. Clear the PSWE bit.
- Step 12. Re-enable interrupts.

Steps 5–10 must be repeated for each block to be written.



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Parameter	Conditions	Min	Тур	Max	Units
Flash Size	C8051F340/2/4/6/A/C/D*	65536*			Bytes
Flash Size	C8051F341/3/5/7/8/9/B	32768		20	Bytes
Endurance		20k	100k		Erase/Write
Erase Cycle Time	25 MHz System Clock	10	15	20	ms
Write Cycle Time	25 MHz System Clock	40	55	70	μs

### Table 12.1. Flash Electrical Characteristics

\*Note: 1024 bytes at location 0xFC00 to 0xFFFF are reserved.

### 12.2. Non-Volatile Data Storage

The Flash memory can be used for non-volatile data storage as well as program code. This allows data such as calibration coefficients to be calculated and stored at run time. Data is written using the MOVX write instruction and read using the MOVC instruction. Note: MOVX read instructions always target XRAM.

## 12.3. Security Options

The CIP-51 provides security options to protect the Flash memory from inadvertent modification by software as well as to prevent the viewing of proprietary program code and constants. The Program Store Write Enable (bit PSWE in register PSCTL) and the Program Store Erase Enable (bit PSEE in register PSCTL) bits protect the Flash memory from accidental modification by software. PSWE must be explicitly set to '1' before software can modify the Flash memory; both PSWE and PSEE must be set to '1' before software can erase Flash memory. Additional security features prevent proprietary program code and data constants from being read or altered across the C2 interface.

A Security Lock Byte located at the last byte of Flash user space offers protection of the Flash program memory from access (reads, writes, or erases) by unprotected code or the C2 interface. The Flash security mechanism allows the user to lock n 512-byte Flash pages, starting at page 0 (addresses 0x0000 to 0x01FF), where n is the 1's complement number represented by the Security Lock Byte. Note that the page containing the Flash Security Lock Byte is also locked when any other Flash pages are locked. See example below.

Security Lock Byte:	1111101b
1's Complement:	0000010b
Flash pages locked:	3 (2 + Flash Lock Byte Page)
	First two pages of Flash: 0x0000 to 0x03FF
Addresses locked:	Flash Lock Byte Page: (0xFA00 to 0xFBFF for 64k devices; 0x7E00 to 0x7FFF for 32k devices)



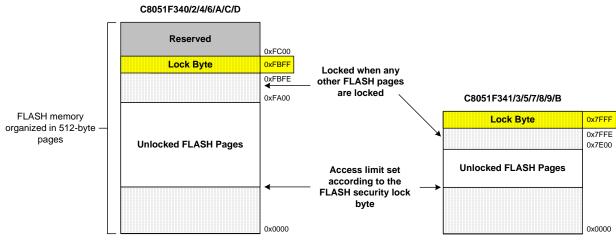


Figure 12.1. Flash Program Memory Map and Security Byte



The level of FLASH security depends on the FLASH access method. The three FLASH access methods that can be restricted are reads, writes, and erases from the C2 debug interface, user firmware executing on unlocked pages, and user firmware executing on locked pages.

#### Accessing FLASH from the C2 debug interface:

- 1. Any unlocked page may be read, written, or erased.
- 2. Locked pages cannot be read, written, or erased.
- 3. The page containing the Lock Byte may be read, written, or erased if it is unlocked.
- 4. Reading the contents of the Lock Byte is always permitted.
- 5. Locking additional pages (changing '1's to '0's in the Lock Byte) is not permitted.
- 6. Unlocking FLASH pages (changing '0's to '1's in the Lock Byte) requires the C2 Device Erase command, which erases all FLASH pages including the page containing the Lock Byte and the Lock Byte itself.
- 7. The Reserved Area cannot be read, written, or erased.

#### Accessing FLASH from user firmware executing on an unlocked page:

- 1. Any unlocked page except the page containing the Lock Byte may be read, written, or erased.
- 2. Locked pages cannot be read, written, or erased.
- 3. The page containing the Lock Byte cannot be erased. It may be read or written only if it is unlocked.
- 4. Reading the contents of the Lock Byte is always permitted.
- 5. Locking additional pages (changing '1's to '0's in the Lock Byte) is not permitted.
- 6. Unlocking FLASH pages (changing '0's to '1's in the Lock Byte) is not permitted.
- 7. The Reserved Area cannot be read, written, or erased. Any attempt to access the reserved area, or any other locked page, will result in a FLASH Error device reset.

#### Accessing FLASH from user firmware executing on a locked page:

- 1. Any unlocked page except the page containing the Lock Byte may be read, written, or erased.
- 2. Any locked page except the page containing the Lock Byte may be read, written, or erased.
- 3. The page containing the Lock Byte cannot be erased. It may only be read or written.
- 4. Reading the contents of the Lock Byte is always permitted.
- 5. Locking additional pages (changing '1's to '0's in the Lock Byte) is not permitted.
- 6. Unlocking FLASH pages (changing '0's to '1's in the Lock Byte) is not permitted.
- 7. The Reserved Area cannot be read, written, or erased. Any attempt to access the reserved area, or any other locked page, will result in a FLASH Error device reset.



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
-	-	-	-	-	Reserved	PSEE	PSWE	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
								0x8F
Bits7–3: Bit2: Bit1: Bit0:	Unused: Rea Reserved. R PSEE: Progr Setting this b to be erased Flash memo tion address 0: Flash prog 1: Flash prog PSWE: Prog Setting this b write instruct 0: Writes to I 1: Writes to I memory.	lead = 0b. N ram Store E bit (in combi I. If this bit is ry using the ed by the N gram memo gram memo gram Store N bit allows wittion. The Fla Flash program	Aust Write = rase Enabl nation with s logic 1 an MOVX instru- ory erasure ory erasure Vrite Enabl riting a byte ash locatior am memory	= 0b. e PSWE) allo d Flash writ truction will iction. The v disabled. enabled. e of data to to n should be v disabled.	tes are enab erase the e value of the the Flash pro erased befo	oled (PSWE ntire page data byte w ogram men ore writing o	is logic 1) that contain written does nory using data.	, a write to no the loca- not matter.

## SFR Definition 12.2. FLKEY: Flash Lock and Key

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xB7
Bits-0:	FLKEY: Flas Write: This register remains lock timing of the must be writt system reset codes have l Read: When read, 00: Flash is 01: The first 10: Flash is 11: Flash writ	must be w ked until this writes does ten for each t if the wror been writte bits 1-0 ind write/erase key code h unlocked (v	ritten to before seregister is a not matter, a Flash write an correctly. in cate the cu- locked. as been wri- writes/erases	ore Flash w written to w as long as e or erase o written or rrent Flash tten (0xA5) s allowed).	vith the follow the codes a operation. Fla if a Flash op lock state.	wing key coo re written in ash will be l	des: 0xA5 order. Th locked un	5, 0xF1. The e key codes til the next



## SFR Definition 12.3. FLSCL: Flash Scale

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value		
FOSE	Reserved	Reserved	FLRT	Reserved	Reserved	Reserved	Reserved	10000000		
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:		
								0xB6		
Bits7:	<ul> <li>Bits7: FOSE: Flash One-shot Enable</li> <li>This bit enables the Flash read one-shot. When the Flash one-shot disabled, the Flash sense amps are enabled for a full clock cycle during Flash reads. At system clock frequencies below 10 MHz, disabling the Flash one-shot will increase system power consumption.</li> <li>0: Flash one-shot disabled.</li> <li>1: Flash one-shot enabled.</li> </ul>									
Bits6–5: Bit 4:	RESERVED			rite 00b.						
Dit <b>4</b> .	This bit shou speed. 0: SYSCLK • 1: SYSCLK •	ild be progra		he smallest	allowed val	ue, accordi	ng to the sy	stem clock		
Bits3–0:	RESERVED	. Read = 00	00b. Must	Write 0000k	).					



## 13. External Data Memory Interface and On-Chip XRAM

4k Bytes (C8051F340/2/4/6/A/C/D) or 2k Bytes (C8051F341/3/5/7/8/9/B) of RAM are included on-chip, and mapped into the external data memory space (XRAM). The 1k Bytes of USB FIFO space can also be mapped into XRAM address space for additional general-purpose data storage. Additionally, an External Memory Interface (EMIF) is available on the C8051F340/1/4/5/8/C devices, which can be used to access off-chip data memories and memory-mapped devices connected to the GPIO ports. The external memory space may be accessed using the external move instruction (MOVX) and the data pointer (DPTR), or using the MOVX indirect addressing mode using R0 or R1. If the MOVX instruction is used with an 8-bit address operand (such as @R1), then the high byte of the 16-bit address is provided by the External Memory Interface Control Register (EMIOCN, shown in SFR Definition 13.1). Note: the MOVX instruction can also be used for writing to the FLASH memory. See Section "12. Flash Memory" on page 107 for details. The MOVX instruction accesses XRAM by default.

## 13.1. Accessing XRAM

The XRAM memory space is accessed using the MOVX instruction. The MOVX instruction has two forms, both of which use an indirect addressing method. The first method uses the Data Pointer, DPTR, a 16-bit register which contains the effective address of the XRAM location to be read from or written to. The second method uses R0 or R1 in combination with the EMI0CN register to generate the effective XRAM address. Examples of both of these methods are given below.

### 13.1.1. 16-Bit MOVX Example

The 16-bit form of the MOVX instruction accesses the memory location pointed to by the contents of the DPTR register. The following series of instructions reads the value of the byte at address 0x1234 into the accumulator A:

MOV	DPTR, #1234h	; load DPTR with 16-bit address to read (0x1234)
MOVX	A, @DPTR	; load contents of 0x1234 into accumulator A

The above example uses the 16-bit immediate MOV instruction to set the contents of DPTR. Alternately, the DPTR can be accessed through the SFR registers DPH, which contains the upper 8-bits of DPTR, and DPL, which contains the lower 8-bits of DPTR.

#### 13.1.2. 8-Bit MOVX Example

The 8-bit form of the MOVX instruction uses the contents of the EMI0CN SFR to determine the upper 8-bits of the effective address to be accessed and the contents of R0 or R1 to determine the lower 8-bits of the effective address to be accessed. The following series of instructions read the contents of the byte at address 0x1234 into the accumulator A.

MOV	EMIOCN, #12h	; load high byte of address into EMIOCN	
MOV	R0, #34h	; load low byte of address into R0 (or R1)	
MOVX	a, @R0	; load contents of 0x1234 into accumulator a	Ŧ



## 13.2. Accessing USB FIFO Space

The C8051F34x devices include 1k of RAM which functions as USB FIFO space. Figure 13.1 shows an expanded view of the FIFO space and user XRAM. FIFO space is normally accessed via USB FIFO registers; see **Section "16.5. FIFO Management" on page 167** for more information on accessing these FIFOs. The MOVX instruction should not be used to load or modify USB data in the FIFO space.

Unused areas of the USB FIFO space may be used as general purpose XRAM if necessary. The FIFO block operates on the USB clock domain; thus the USB clock must be active when accessing FIFO space. Note that the number of SYSCLK cycles required by the MOVX instruction is increased when accessing USB FIFO space.

To access the FIFO RAM directly using MOVX instructions, the following conditions must be met: (1) the USBFAE bit in register EMI0CF must be set to '1', and (2) the USB clock must be greater than or equal to twice the SYSCLK (USBCLK  $\geq$  2 x SYSCLK). When this bit is set, the USB FIFO space is mapped into XRAM space at addresses 0x0400 to 0x07FF. The normal XRAM (on-chip or external) at the same addresses cannot be accessed when the USBFAE bit is set to '1'.

#### Important Note: The USB clock must be active when accessing FIFO space.

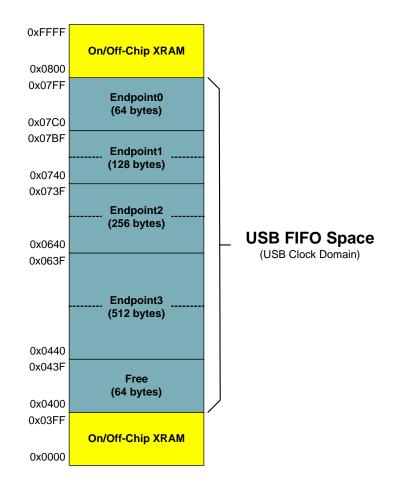


Figure 13.1. USB FIFO Space and XRAM Memory Map with USBFAE set to '1'



## **13.3.** Configuring the External Memory Interface

Configuring the External Memory Interface consists of five steps:

- 1. Configure the Output Modes of the associated port pins as either push-pull or open-drain (push-pull is most common), and skip the associated pins in the crossbar.
- 2. Configure Port latches to "park" the EMIF pins in a dormant state (usually by setting them to logic '1').
- 3. Select Multiplexed mode or Non-multiplexed mode.
- 4. Select the memory mode (on-chip only, split mode without bank select, split mode with bank select, or off-chip only).
- 5. Set up timing to interface with off-chip memory or peripherals.

Each of these five steps is explained in detail in the following sections. The Port selection, Multiplexed mode selection, and Mode bits are located in the EMI0CF register shown in SFR Definition 13.2.

#### 13.4. Port Configuration

The External Memory Interface appears on Ports 4, 3, 2, and 1 when it is used for off-chip memory access. When the EMIF is used, the Crossbar should be configured to skip over the control lines P1.7 (WR), P1.6 ( $\overline{\text{RD}}$ ), and if multiplexed mode is selected P1.3 (ALE) using the P1SKIP register. For more information about configuring the Crossbar, see Section "Figure 15.1. Port I/O Functional Block Diagram (Port 0 through Port 3)" on page 142.

The External Memory Interface claims the associated Port pins for memory operations ONLY during the execution of an off-chip MOVX instruction. Once the MOVX instruction has completed, control of the Port pins reverts to the Port latches or to the Crossbar settings for those pins. See Section "15. Port Input/ Output" on page 142 for more information about the Crossbar and Port operation and configuration. The Port latches should be explicitly configured to 'park' the External Memory Interface pins in a dormant state, most commonly by setting them to a logic 1.

During the execution of the MOVX instruction, the External Memory Interface will explicitly disable the drivers on all Port pins that are acting as Inputs (Data[7:0] during a READ operation, for example). The Output mode of the Port pins (whether the pin is configured as Open-Drain or Push-Pull) is unaffected by the External Memory Interface operation, and remains controlled by the PnMDOUT registers. In most cases, the output modes of all EMIF pins should be configured for push-pull mode.



## SFR Definition 13.1. EMI0CN: External Memory Interface Control

	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
P	GSEL7	PGSEL6	PGSEL5	PGSEL4	PGSEL3	PGSEL2	PGSEL1	PGSEL0	0000000
	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
								SFR Address:	0xAA
Bit	T a R 0 0 0 0	ddress whe AM. x00: 0x000 x01: 0x010	Page Select on using an 0 to 0x00FF 0 to 0x01FF 00 to 0xFEF	Bits provid 8-bit MOV> <del>-</del> F	its. le the high t ( command				



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
-	USBFAE	-	EMD2	EMD1	EMD0	EALE1	EALE0	00000011
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
							SFR Address	: 0x85
Bit7:	Unused. Rea	ad = 0b. Wr	rite = don't d	care.				
Bit6:	USBFAE: US	SB FIFO Ad	cess Enab	le.				
	0: USB FIFO							
	1: USB FIFO							
	in XRAM spa							
	greater than	-		e SYSCLK	(USBCLK <u>&gt;</u>	<u>&gt;</u> 2 x SYSC	LK) to acc	ess this
D:45	area with MO							
Bit5:	Unused. Rea							
Bit4:	EMD2: EMIF 0: EMIF oper	•			mode			
	1: EMIF oper		•			ress and da	ta nins)	
Bits3-2:	EMD1-0: EM				parate add		ta pinoj.	
2100 2.	These bits co		•		xternal Mer	morv Interfa	ce.	
	00: Internal C							alias to
	on-chip mem			•	,			
	01: Split Mod	le without E	Bank Select	: Accesses	below the o	on-chip XRA	M bounda	ry are
	directed on-c	hip. Acces	ses above	the on-chip	XRAM bou	ndary are di	irected off-	chip. 8-bit
	off-chip MOV	•						
	resolve uppe						ce, EMI0CI	N must be
	set to a page				•			
	10: Split Mod							
	on-chip. Acce							
	MOVX opera 11: External (					•		
	CPU.		A dilesses		CAIVI OTIIY. C			
Bits1–0:	EALE1–0: AL	E Pulse-V	Vidth Select	Bits (only h	as effect w	hen EMD2	= 0).	
Dito i oi	00: ALE high						0).	
	01: ALE high							
	10: ALE high							
	11: ALE high	and ALE I	ow pulse w	idth = 4 SYS	SCLK cycle	S.		



## 13.5. Multiplexed and Non-multiplexed Selection

The External Memory Interface is capable of acting in a Multiplexed mode or a Non-multiplexed mode, depending on the state of the EMD2 (EMI0CF.4) bit.

#### 13.5.1. Multiplexed Configuration

In Multiplexed mode, the Data Bus and the lower 8-bits of the Address Bus share the same Port pins: AD[7:0]. In this mode, an external latch (74HC373 or equivalent logic gate) is used to hold the lower 8-bits of the RAM address. The external latch is controlled by the ALE (Address Latch Enable) signal, which is driven by the External Memory Interface logic. An example of a Multiplexed Configuration is shown in Figure 13.2.

In Multiplexed mode, the external MOVX operation can be broken into two phases delineated by the state of the ALE signal. During the first phase, ALE is high and the lower 8-bits of the Address Bus are presented to AD[7:0]. During this phase, the address latch is configured such that the 'Q' outputs reflect the states of the 'D' inputs. When ALE falls, signaling the beginning of the second phase, the address latch outputs remain fixed and are no longer dependent on the latch inputs. Later in the second phase, the Data Bus controls the state of the AD[7:0] port at the time RD or WR is asserted.

See Section "13.7.2. Multiplexed Mode" on page 127 for more information.

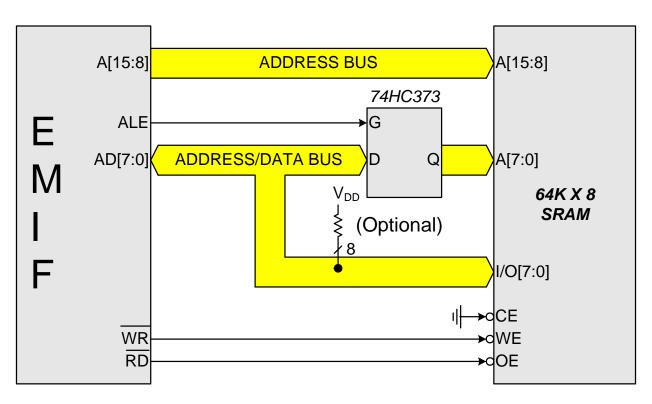


Figure 13.2. Multiplexed Configuration Example



#### 13.5.2. Non-multiplexed Configuration

In Non-multiplexed mode, the Data Bus and the Address Bus pins are not shared. An example of a Non-multiplexed Configuration is shown in Figure 13.3. See **Section "13.7.1. Non-multiplexed Mode" on page 124** for more information about Non-multiplexed operation.

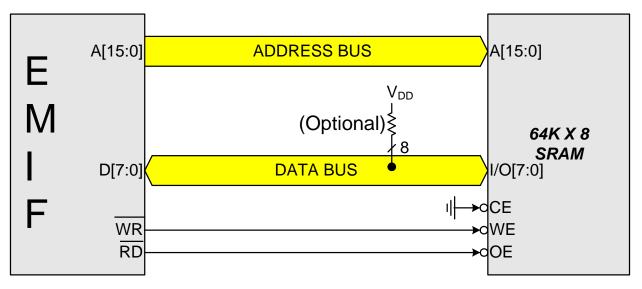


Figure 13.3. Non-multiplexed Configuration Example

## 13.6. Memory Mode Selection

The external data memory space can be configured in one of four modes, shown in Figure 13.4, based on the EMIF Mode bits in the EMIOCF register (SFR Definition 13.2). These modes are summarized below. More information about the different modes can be found in **Section "13.7. Timing" on page 122**.

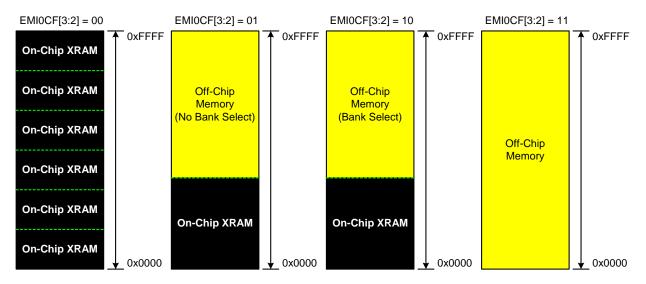


Figure 13.4. EMIF Operating Modes



#### 13.6.1. Internal XRAM Only

When EMI0CF.[3:2] are set to '00', all MOVX instructions will target the internal XRAM space on the device. Memory accesses to addresses beyond the populated space will wrap on 2k or 4k boundaries (depending on the RAM available on the device). As an example, the addresses 0x1000 and 0x2000 both evaluate to address 0x0000 in on-chip XRAM space.

- 8-bit MOVX operations use the contents of EMI0CN to determine the high-byte of the effective address and R0 or R1 to determine the low-byte of the effective address.
- 16-bit MOVX operations use the contents of the 16-bit DPTR to determine the effective address.

#### 13.6.2. Split Mode without Bank Select

When EMI0CF.[3:2] are set to '01', the XRAM memory map is split into two areas, on-chip space and off-chip space.

- Effective addresses below the internal XRAM size boundary will access on-chip XRAM space.
- Effective addresses above the internal XRAM size boundary will access off-chip space.
- 8-bit MOVX operations use the contents of EMI0CN to determine whether the memory access is on-chip or off-chip. However, in the "No Bank Select" mode, an 8-bit MOVX operation will not drive the upper 8-bits A[15:8] of the Address Bus during an off-chip access. This allows the user to manipulate the upper address bits at will by setting the Port state directly via the port latches. This behavior is in contrast with "Split Mode with Bank Select" described below. The lower 8-bits of the Address Bus A[7:0] are driven, determined by R0 or R1.
- 16-bit MOVX operations use the contents of DPTR to determine whether the memory access is on-chip or off-chip, and unlike 8-bit MOVX operations, the full 16-bits of the Address Bus A[15:0] are driven during the off-chip transaction.



#### 13.6.3. Split Mode with Bank Select

When EMI0CF.[3:2] are set to '10', the XRAM memory map is split into two areas, on-chip space and off-chip space.

- Effective addresses below the internal XRAM size boundary will access on-chip XRAM space.
- Effective addresses above the internal XRAM size boundary will access off-chip space.
- 8-bit MOVX operations use the contents of EMI0CN to determine whether the memory access is on-chip or off-chip. The upper 8-bits of the Address Bus A[15:8] are determined by EMI0CN, and the lower 8-bits of the Address Bus A[7:0] are determined by R0 or R1. All 16-bits of the Address Bus A[15:0] are driven in "Bank Select" mode.
- 16-bit MOVX operations use the contents of DPTR to determine whether the memory access is on-chip or off-chip, and the full 16-bits of the Address Bus A[15:0] are driven during the off-chip transaction.

#### 13.6.4. External Only

When EMI0CF[3:2] are set to '11', all MOVX operations are directed to off-chip space. On-chip XRAM is not visible to the CPU. This mode is useful for accessing off-chip memory located between 0x0000 and the internal XRAM size boundary.

- 8-bit MOVX operations ignore the contents of EMI0CN. The upper Address bits A[15:8] are not driven (identical behavior to an off-chip access in "Split Mode without Bank Select" described above). This allows the user to manipulate the upper address bits at will by setting the Port state directly. The lower 8-bits of the effective address A[7:0] are determined by the contents of R0 or R1.
- 16-bit MOVX operations use the contents of DPTR to determine the effective address A[15:0]. The full 16-bits of the Address Bus A[15:0] are driven during the off-chip transaction.

#### 13.7. Timing

The timing parameters of the External Memory Interface can be configured to enable connection to devices having different setup and hold time requirements. The Address Setup time, Address Hold time, RD and WR strobe widths, and in multiplexed mode, the width of the ALE pulse are all programmable in units of SYSCLK periods through EMI0TC, shown in SFR Definition 13.3, and EMI0CF[1:0].

The timing for an off-chip MOVX instruction can be calculated by adding 4 SYSCLK cycles to the timing parameters defined by the EMI0TC register. Assuming non-multiplexed operation, the minimum execution time for an off-chip XRAM operation is 5 SYSCLK cycles (1 SYSCLK for RD or WR pulse + 4 SYSCLKs). For multiplexed operations, the Address Latch Enable signal will require a minimum of 2 additional SYSCLK cycles. Therefore, the minimum execution time for an off-chip XRAM operation in multiplexed mode is 7 SYSCLK cycles (2 for ALE + 1 for RD or WR + 4). The programmable setup and hold times default to the maximum delay settings after a reset. Table 13.1 lists the AC parameters for the External Memory Interface, and Figure 13.5 through Figure 13.10 show the timing diagrams for the different External Memory Interface modes and MOVX operations.



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
EAS1	EAS0	EWR3	EWR2	EWR1	EWR0	EAH1	EAH0	11111111
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
							SFR Address	: 0x84
Bits7–6:	EAS1-0: EM	1IF Address	Setup Tim	e Bits.				
	00: Address							
	01: Address							
	10: Address							
	11: Address							
Bits5–2:	EWR3 <u>–0:</u> EN							
	0000: <u>WR</u> ar							
	0001: <u>WR</u> ar							
	0010: <u>WR</u> ar							
	0011: WR ar							
	0100: <u>WR</u> ar							
	0101: <u>WR</u> ar							
	0110: <u>WR</u> ar							
	0111: <u>WR</u> ar							
	1000: <u>WR</u> ar							
	1001: <u>WR</u> ar							
	1010: <u>WR</u> ar							
	1011: <u>WR</u> ar							
	1100: <u>WR</u> ar							
	1101: <u>WR</u> ar							
	1110: WR ar							
	1111:WR and				cles.			
Bits1–0:	EAH1-0: EN							
	00: Address							
	01: Address							
	10: Address							
	11: Address	1010  time =	3 3 1 30 LK	cycles.				

## SFR Definition 13.3. EMI0TC: External Memory Timing Control



### 13.7.1. Non-multiplexed Mode

13.7.1.1.16-bit MOVX: EMI0CF[4:2] = '101', '110', or '111'.

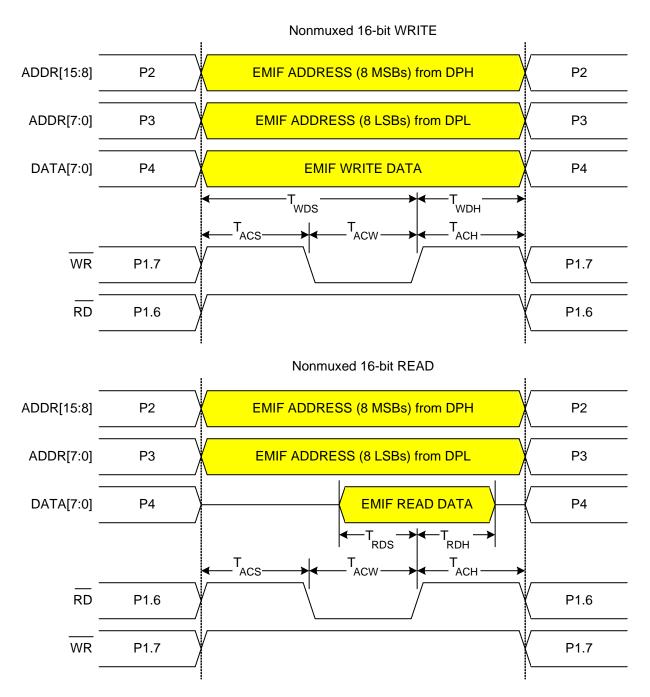
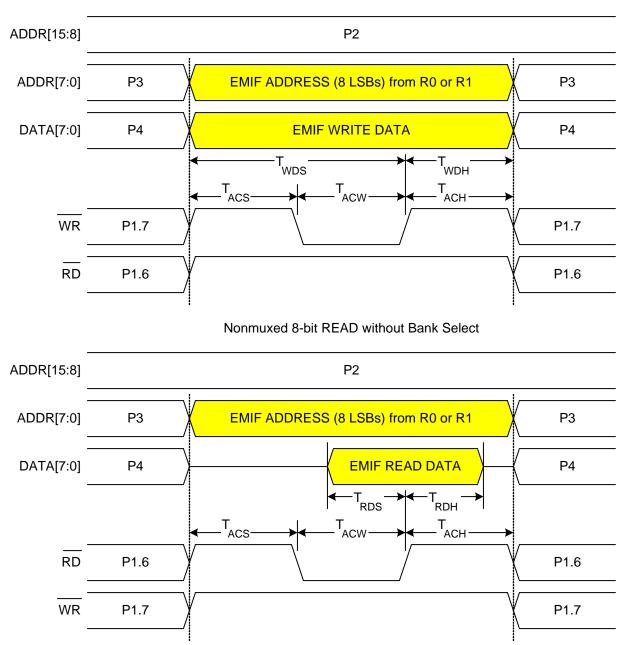


Figure 13.5. Non-multiplexed 16-bit MOVX Timing



### 13.7.1.2.8-bit MOVX without Bank Select: EMI0CF[4:2] = '101' or '111'.

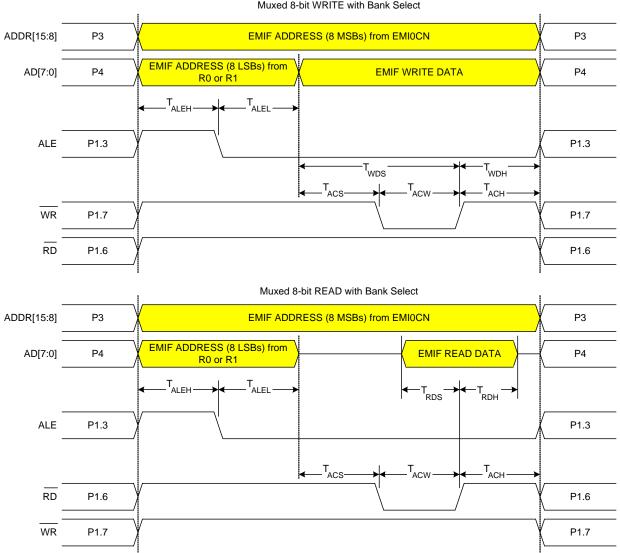


Nonmuxed 8-bit WRITE without Bank Select

Figure 13.6. Non-multiplexed 8-bit MOVX without Bank Select Timing



#### 13.7.1.3.8-bit MOVX with Bank Select: EMI0CF[4:2] = '110'.



Muxed 8-bit WRITE with Bank Select

Figure 13.7. Non-multiplexed 8-bit MOVX with Bank Select Timing



## 13.7.2. Multiplexed Mode

## 13.7.2.1.16-bit MOVX: EMI0CF[4:2] = '001', '010', or '011'.

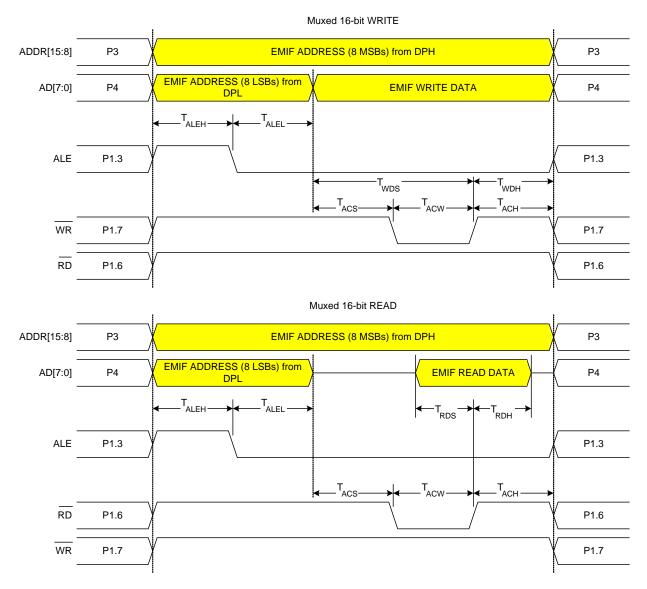
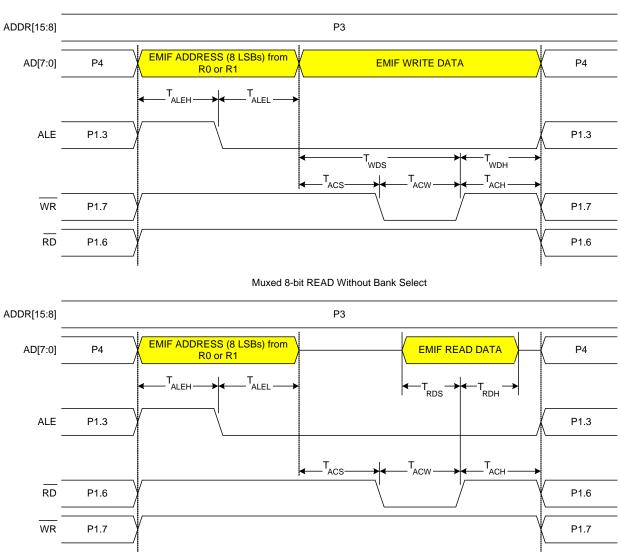


Figure 13.8. Multiplexed 16-bit MOVX Timing



#### 13.7.2.2.8-bit MOVX without Bank Select: EMI0CF[4:2] = '001' or '011'.

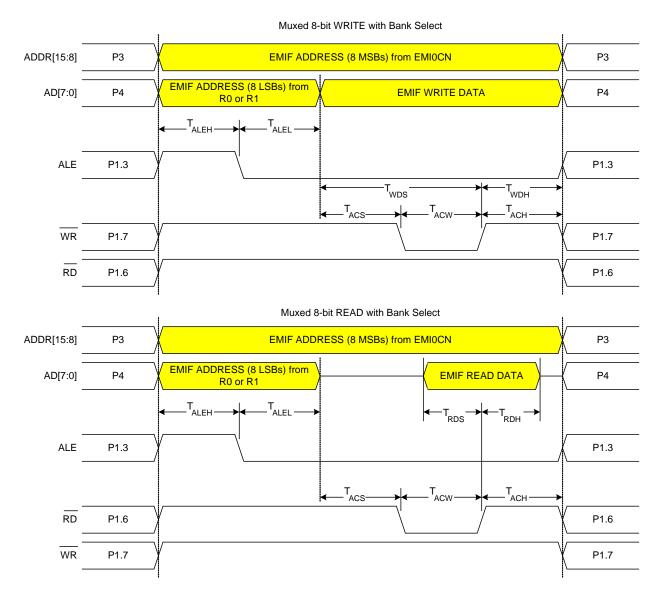


Muxed 8-bit WRITE Without Bank Select





#### 13.7.2.3.8-bit MOVX with Bank Select: EMI0CF[4:2] = '010'.







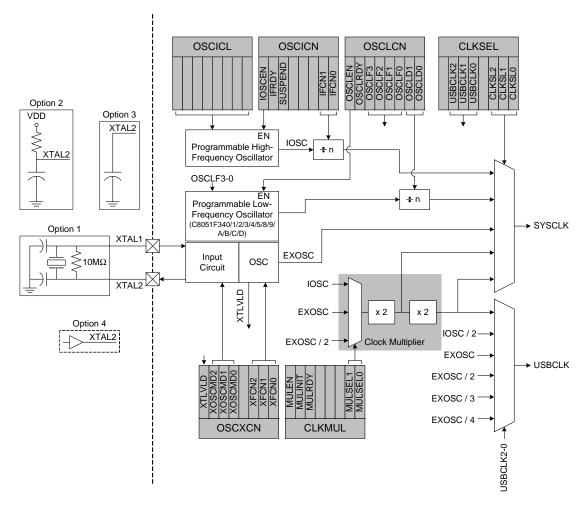
Parameter	Description	Min*	Max*	Units				
T <sub>ACS</sub>	Address / Control Setup Time	0	3 x T <sub>SYSCLK</sub>	ns				
T <sub>ACW</sub>	Address / Control Pulse Width	1 x T <sub>SYSCLK</sub>	16 x T <sub>SYSCLK</sub>	ns				
T <sub>ACH</sub>	Address / Control Hold Time	0	3 x T <sub>SYSCLK</sub>	ns				
T <sub>ALEH</sub>	Address Latch Enable High Time	1 x T <sub>SYSCLK</sub>	4 x T <sub>SYSCLK</sub>	ns				
T <sub>ALEL</sub>	Address Latch Enable Low Time	1 x T <sub>SYSCLK</sub>	4 x T <sub>SYSCLK</sub>	ns				
T <sub>WDS</sub>	Write Data Setup Time	1 x T <sub>SYSCLK</sub>	19 x T <sub>SYSCLK</sub>	ns				
T <sub>WDH</sub>	Write Data Hold Time	0	3 x T <sub>SYSCLK</sub>	ns				
T <sub>RDS</sub>	Read Data Setup Time	20		ns				
T <sub>RDH</sub>	Read Data Hold Time	0		ns				
Note: T <sub>SYSCLK</sub> is equal to one period of the device system clock (SYSCLK).								

Table 13.1. AC Parameters for External Memory Interface



## 14. Oscillators

C8051F34x devices include a programmable internal high-frequency oscillator, a programmable internal low-frequency oscillator (C8051F340/1/2/3/4/5/8/9/A/B/C/D), an external oscillator drive circuit, and a 4x Clock Multiplier. The internal high-frequency and low-frequency oscillators can be enabled/disabled and adjusted using the special function registers, as shown in Figure 14.1. The system clock (SYSCLK) can be derived from either of the internal oscillators, the external oscillator circuit, or the 4x Clock Multiplier divided by 2. The USB clock (USBCLK) can be derived from the internal oscillator, external oscillator, or 4x Clock Multiplier. Oscillator electrical specifications are given in Table 14.1.







## 14.1. Programmable Internal High-Frequency (H-F) Oscillator

All C8051F34x devices include a programmable internal oscillator that defaults as the system clock after a system reset. The internal oscillator period can be programmed via the OSCICL register shown in SFR Definition 14.2. The OSCICL register is factory calibrated to obtain a 12 MHz internal oscillator frequency. Electrical specifications for the precision internal oscillator are given in Table 14.1 on page 141. Note that the system clock may be derived from the programmed internal oscillator divided by 1, 2, 4, or 8, as defined by the IFCN bits in register OSCICN. The divide value defaults to 8 following a reset.

#### 14.1.1. Internal H-F Oscillator Suspend Mode

The internal high-frequency oscillator may be placed in Suspend mode by writing '1' to the SUSPEND bit in register OSCICN. In Suspend mode, the internal H-F oscillator is stopped until a non-idle USB event is detected (Section 16) or VBUS matches the polarity selected by the VBPOL bit in register REGOCN (Section 8.2). Note that the USB transceiver can still detect USB events when it is disabled.

## SFR Definition 14.1. OSCICN: Internal H-F Oscillator Control

	-	-	5.44	_	-	5.44	-	5.44	5
	R/W	R	R/W	R	R/W	R/W	R/W	R/W	Reset Value
	IOSCEN	IFRDY	SUSPEND	-	-	-	IFCN1	IFCN0	10000000
	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
									0xB2
В	it7:	IOSCEN: In	ternal H-F O	scillator En	able Bit.				
		0: Internal H	I-F Oscillator	Disabled.					
		1: Internal H	I-F Oscillator	Enabled.					
В	it6:	IFRDY: Inter	rnal H-F Osc	illator Freq	uency Read	dy Flag.			
		0: Internal H	I-F Oscillator	is not runr	ning at prog	rammed fre	quency.		
		1: Internal H	I-F Oscillator	is running	at program	med freque	ency.		
В	it5:	SUSPEND:	Force Suspe	end					
		Writing a '1'	to this bit will	force the i	nternal H-F	oscillator to	be stopped	d. The osci	lator will be
		re-started or	n the next no	n-idle USB	event (i.e.,	RESUME s	signaling) or	· VBUS inte	errupt event
		(see SFR D	efinition 8.1).						
В	its4–2:	UNUSED. R	Read = 000b,	Write = do	on't care.				
В	its1–0:	IFCN1-0: In	iternal H-F O	scillator Fr	equency Co	ontrol.			
		00: SYSCL	K derived fror	n Internal I	H-F Oscillat	or divided b	oy 8.		
		01: SYSCL	K derived fror	n Internal I	H-F Oscillat	or divided b	oy 4.		
		10: SYSCL	K derived fror	n Internal I	H-F Oscillat	or divided b	oy 2.		
		11: SYSCLK	C derived fror	n Internal H	H-F Oscillat	or divided b	y 1.		



SFR Definition 14.2.	<b>OSCICL: Internal H-F</b>	<b>Oscillator Calibration</b>
----------------------	-----------------------------	-------------------------------

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value	
-	-	-			OSCCAL			Variable	
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xB3	
Bits4–0: OSCCAL: Oscillator Calibration Value These bits determine the internal H-F oscillator period. When set to 00000b, the oscillator operates at its fastest setting. When set to 11111b, the oscillator operates at is slowest set- ting. The contents of this register are factory calibrated to produce a 12 MHz internal oscilla- tor frequency.									
tor frequency. Note: The contents of this register are undefined when Clock Recovery is enabled. See Section "16.4. USB Clock Configuration" on page 166 for details on Clock Recovery.									

## 14.2. Programmable Internal Low-Frequency (L-F) Oscillator

The C8051F340/1/2/3/4/5/8/9/C/D devices include a programmable internal oscillator which operates at a nominal frequency of 80 kHz. The low-frequency oscillator circuit includes a divider that can be changed to divide the clock by 1, 2, 4, or 8, using the OSCLD bits in the OSCLCN register (see SFR Definition 14.3). Additionally, the OSCLF bits (OSCLCN5:2) can be used to adjust the oscillator's output frequency.

### 14.2.1. Calibrating the Internal L-F Oscillator

Timers 2 and 3 include capture functions that can be used to capture the oscillator frequency, when running from a known time base. When either Timer 2 or Timer 3 is configured for L-F Oscillator Capture Mode, a falling edge (Timer 2) or rising edge (Timer 3) of the low-frequency oscillator's output will cause a capture event on the corresponding timer. As a capture event occurs, the current timer value (TMRnH:TMRnL) is copied into the timer reload registers (TMRnRLH:TMRnRLL). By recording the difference between two successive timer capture values, the low-frequency oscillator's period can be calculated. The OSCLF bits can then be adjusted to produce the desired oscillator period.



## SFR Definition 14.3. OSCLCN: Internal L-F Oscillator Control

R/W	R	R/W	R	R/W	R/W	R/W	R/W	Reset Value	
-									
OSCLEN			OSCLF2	OSCLF1	OSCLF0	OSCLD1	OSCLD0	00vvvv00	
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:	
								0x86	
Bit7: OSCLEN: Internal L-F Oscillator Enable. 0: Internal L-F Oscillator Disabled.									
Bit6:	1: Internal L-F Oscillator Enabled. Bit6: OSCLRDY: Internal L-F Oscillator Ready Flag. 0: Internal L-F Oscillator frequency not stabilized. 1: Internal L-F Oscillator frequency stabilized.								
Bits5–2:	Bits5–2: OSCLF[3:0]: Internal L-F Oscillator Frequency Control bits. Fine-tune control bits for the internal L-F Oscillator frequency. When set to 0000b, the L-F oscillator operates at its fastest setting. When set to 1111b, the L-F oscillator operates at its								
Bits1–0:	<ul> <li>slowest setting.</li> <li>OSCLD[1:0]: Internal L-F Oscillator Divider Select.</li> <li>00: Divide by 8 selected.</li> <li>01: Divide by 4selected.</li> <li>10: Divide by 2 selected.</li> <li>11: Divide by 1 selected.</li> </ul>								



## 14.3. External Oscillator Drive Circuit

The external oscillator circuit may drive an external crystal, ceramic resonator, capacitor, or RC network. A CMOS clock may also provide a clock input. For a crystal or ceramic resonator configuration, the crystal/ resonator must be wired across the XTAL1 and XTAL2 pins as shown in Option 1 of Figure 14.1. A 10 M $\Omega$  resistor also must be wired across the XTAL1 and XTAL2 pins for the crystal/resonator configuration. In RC, capacitor, or CMOS clock configuration, the clock source should be wired to the XTAL2 pin as shown in Option 2, 3, or 4 of Figure 14.1. The type of external oscillator must be selected in the OSCXCN register, and the frequency control bits (XFCN) must be selected appropriately (see SFR Definition 14.4)

**Important Note on External Oscillator Usage:** Port pins must be configured when using the external oscillator circuit. When the external oscillator drive circuit is enabled in crystal/resonator mode, Port pins P0.6 and P0.7 (C8051F340/1/4/5/8) or P0.2 and P0.3 (C8051F342/3/6/7/9/A/B) are used as XTAL1 and XTAL2 respectively. When the external oscillator drive circuit is enabled in capacitor, RC, or CMOS clock mode, Port pin P0.7 (C8051F340/1/4/5/8) or P0.3 (C8051F342/3/6/7/9/A/B) is used as XTAL2. The Port I/ O Crossbar should be configured to skip the Port pins used by the oscillator circuit; see Section "15.1. Priority Crossbar Decoder" on page 144 for Crossbar configuration. Additionally, when using the external oscillator circuit in crystal/resonator, capacitor, or RC mode, the associated Port pins should be configured as a digital input. See Section "15.2. Port I/O Initialization" on page 147 for details on Port input mode selection.

#### 14.3.1. Clocking Timers Directly Through the External Oscillator

The external oscillator source divided by eight is a clock option for the timers (Section "21. Timers" on page 235) and the Programmable Counter Array (PCA) (Section "22. Programmable Counter Array (PCA0)" on page 255). When the external oscillator is used to clock these peripherals, but is not used as the system clock, the external oscillator frequency must be less than or equal to the system clock frequency. In this configuration, the clock supplied to the peripheral (external oscillator / 8) is synchronized with the system clock; the jitter associated with this synchronization is limited to ±0.5 system clock cycles.

#### 14.3.2. External Crystal Example

If a crystal or ceramic resonator is used as an external oscillator source for the MCU, the circuit should be configured as shown in Figure 14.1, Option 1. The External Oscillator Frequency Control value (XFCN) should be chosen from the Crystal column of the table in SFR Definition 14.4 (OSCXCN register). For example, a 12 MHz crystal requires an XFCN setting of 111b.

When the crystal oscillator is first enabled, the oscillator amplitude detection circuit requires a settling time to achieve proper bias. Introducing a delay of 1 ms between enabling the oscillator and checking the XTLVLD bit will prevent a premature switch to the external oscillator as the system clock. Switching to the external oscillator before the crystal oscillator has stabilized can result in unpredictable behavior. The recommended procedure is:

- Step 1. Enable the external oscillator.
- Step 2. Wait at least 1 ms.
- Step 3. Poll for XTLVLD => '1'.
- Step 4. Switch the system clock to the external oscillator.

**Important Note on External Crystals:** Crystal oscillator circuits are quite sensitive to PCB layout. The crystal should be placed as close as possible to the XTAL pins on the device. The traces should be as short as possible and shielded with ground plane from any other traces which could introduce noise or interference.



#### 14.3.3. External RC Example

If an RC network is used as an external oscillator source for the MCU, the circuit should be configured as shown in Figure 14.1, Option 2. The capacitor should be no greater than 100 pF; however for very small capacitors, the total capacitance may be dominated by parasitic capacitance in the PCB layout. To determine the required External Oscillator Frequency Control value (XFCN) in the OSCXCN Register, first select the RC network value to produce the desired frequency of oscillation. If the frequency desired is 100 kHz, let R = 246 k $\Omega$  and C = 50 pF:

$$f = \frac{1.23(10^3)}{\text{RC}} = \frac{1.23(10^3)}{[246 \times 50]} = 0.1 \text{ MHz} = 100 \text{ kHz}$$

Referring to the table in SFR Definition 14.4, the required XFCN setting is 010b. Programming XFCN to a higher setting in RC mode will improve frequency accuracy at an increased external oscillator supply current.

#### 14.3.4. External Capacitor Example

If a capacitor is used as an external oscillator for the MCU, the circuit should be configured as shown in Figure 14.1, Option 3. The capacitor should be no greater than 100 pF; however for very small capacitors, the total capacitance may be dominated by parasitic capacitance in the PCB layout. To determine the required External Oscillator Frequency Control value (XFCN) in the OSCXCN Register, select the capacitor to be used and find the frequency of oscillation from the equations below. Assume  $V_{DD} = 3.0$  V and C = 50 pF:

$$f = \frac{KF}{(C \times V_{DD})} = \frac{KF}{(50 \text{ x } 3)\text{MHz}}$$

$$f = \frac{KF}{150 \text{ MHz}}$$

If a frequency of roughly 150 kHz is desired, select the K Factor from the table in SFR Definition 14.4 as KF = 22:

$$f = \frac{22}{150} = 0.146$$
 MHz, or 146 kHz

Therefore, the XFCN value to use in this example is 011b.



SFR Definition 14.4	. OSCXCN: External	<b>Oscillator Control</b>
---------------------	--------------------	---------------------------

R	R/W	R/W	R/W	R	R/W	R/W	R/W	Reset Value
XTLVLD	XOSCN	1D2 XOSCMD1	XOSCMD0	-	XFCN2	XFCN1	XFCN0	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
								0xB1
Bit7:		Crystal Oscillat		ıg.				
	•	ly when XOSCI	,	t vot stable				
		I Oscillator is ur I Oscillator is ru						
Bits6–4:		D2–0: External (	•					
Dito0 4.		ernal Oscillator		ouc bito.				
		ernal CMOS Clo						
		ernal CMOS Clo		th divide by	/ 2 stage.			
	100: RC	Oscillator Mode			-			
		acitor Oscillator						
	•	stal Oscillator M						
Dire		stal Oscillator M			age.			
Bit3:		ED. Read = 0, V			al Dita			
Bits2–0:		0: External Osci See table belov	•	iency Conti	OI BITS.			
	000-111.	See lable below	v.					
	XFCN	Crystal (XOSC	,		CMD = 10x		CMD = 10	<)
	000	f ≤ 32 kł			5 kHz		tor = 0.87	
	001	32 kHz < f ≤			f ≤ 50 kHz		ctor = 2.6	
	010	84 kHz < f ≤ 2			≤ 100 kHz		ctor = 7.7	
	011	225 kHz < f ≤			f ≤ 200 kHz		ctor = 22	
	100	590 kHz < f ≤			f ≤ 400 kHz		ctor = 65	
	101	1.5 MHz < f ≤			f ≤ 800 kHz		tor = 180	
	110	4 MHz < f ≤ 1			$f \le 1.6 \text{ MHz}$		tor = 664	
	111	10 MHz < f ≤	30 MHz	1.6 MHz <	f ≤ 3.2 MHz	K Fac	tor = 1590	
CRYSTA		Circuit from Fig	ure 14.1, O	ption 1; XC	SCMD = 11	x)		
		XFCN value to r				,		
RC MOD	•	from Figure 14.	· · ·		,			
		XFCN value to r	•	ency range				
		10 <sup>3</sup> ) / (R x C), w						
		ency of clock in						
		citor value in pF						
	R = Pull-	up resistor value	5 III KS2					
	(Circuit fr	om Figure 14.1,	Option 3: X	(OSCMD =	10x)			
		K Factor (KF) fo						
		C x V <sub>DD</sub> ), where		1	,			
		ency of clock in						
		citor value the >		n pF				
		ower Supply on						



### 14.4. 4x Clock Multiplier

The 4x Clock Multiplier allows a 12 MHz oscillator to generate the 48 MHz clock required for Full Speed USB communication (see Section "16.4. USB Clock Configuration" on page 166). A divided version of the Multiplier output can also be used as the system clock. C8051F340/1/2/3 devices can use the 48 MHz Clock Multiplier output as system clock. See Table 3.1, "Global DC Electrical Characteristics," on page 25 for system clock frequency specifications. See Section 14.5 for details on system clock and USB clock source selection.

The 4x Clock Multiplier is configured via the CLKMUL register. The procedure for configuring and enabling the 4x Clock Multiplier is as follows:

- 1. Reset the Multiplier by writing 0x00 to register CLKMUL.
- 2. Select the Multiplier input source via the MULSEL bits.
- 3. Enable the Multiplier with the MULEN bit (CLKMUL | = 0x80).
- 4. Delay for  $>5 \ \mu s$ .
- 5. Initialize the Multiplier with the MULINIT bit (CLKMUL | = 0xC0).
- 6. Poll for MULRDY = '1'.

Important Note: When using an external oscillator as the input to the 4x Clock Multiplier, the external source must be enabled and stable before the Multiplier is initialized. See Section 14.5 for details on selecting an external oscillator source.

#### SFR Definition 14.5. CLKMUL: Clock Multiplier Control

R/W	R/W	R	R/W	R/W	R/W	R/W	R/W	Reset Value
MULEN		MULRDY	-	-	_	MUL		00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address
	2.10	2.10	2	2.10	2.12	2	2.10	0xB9
								0,05
Bit7:	MULEN: Clo	ck Multiplier	Enable					
	0: Clock Mul							
	1: Clock Mul							
Bit6:	MULINIT: CI	ock Multiplie	r Initialize					
	This bit shou	ıld be a 'Ò' w	hen the Cl	lock Multipli	er is enable	ed. Once en	abled, wri	ting a '1' to
	this bit will in	itialize the C	lock Multip	lier. The Ml	JLRDY bit ı	reads '1' wh	en the Clo	ock Multiplier
	is stabilized.							
Bit5:	MULRDY: C	lock Multiplie	er Ready					
	This read-on	ly bit indicat	es the stat	us of the Cl	ock Multipli	er.		
	0: Clock Mul	tiplier not re	ady.					
	1: Clock Mul		` '					
	Unused. Rea							
Bits1–0:	MULSEL: CI		•					
	These bits se	elect the clo	ck supplied	d to the Cloo	ck Multiplier	r.		
			-		-			
		LSEL	_	elected Clo				
	(	00	In	ternal Oscil	ator			
	(	01	Ex	ternal Oscil	lator			
		10	Exte	ernal Oscilla	tor / 2			
		11		RESERVE	D			



### 14.5. System and USB Clock Selection

The internal oscillator requires little start-up time and may be selected as the system or USB clock immediately following the OSCICN write that enables the internal oscillator. External crystals and ceramic resonators typically require a start-up time before they are settled and ready for use. The Crystal Valid Flag (XTLVLD in register OSCXCN) is set to '1' by hardware when the external oscillator is settled. **To avoid reading a false XTLVLD, in crystal mode software should delay at least 1 ms between enabling the external oscillator and checking XTLVLD.** RC and C modes typically require no startup time.

#### 14.5.1. System Clock Selection

The CLKSL[1:0] bits in register CLKSEL select which oscillator source is used as the system clock. CLKSL[1:0] must be set to 01b for the system clock to run from the external oscillator; however the external oscillator may still clock certain peripherals (timers, PCA, USB) when the internal oscillator is selected as the system clock. The system clock may be switched on-the-fly between the internal oscillator, external oscillator, and 4x Clock Multiplier so long as the selected oscillator is enabled and has settled. C8051F340/ 1/2/3 devices can use the 48 MHz Clock Multiplier output as system clock. See Table 3.1, "Global DC Electrical Characteristics," on page 25 for system clock frequency specifications. When operating with a system clock of greater than 25 MHz (up to 48 MHz), the FLRT bit (FLSCL.4) should be set to '1'. See Section "10. Prefetch Engine" on page 99 for more details.

#### 14.5.2. USB Clock Selection

The USBCLK[2:0] bits in register CLKSEL select which oscillator source is used as the USB clock. The USB clock may be derived from the 4x Clock Multiplier output, a divided version of the internal oscillator, or a divided version of the external oscillator. Note that the USB clock must be 48 MHz when operating USB0 as a Full Speed Function; the USB clock must be 6 MHz when operating USB0 as a Low Speed Function. See SFR Definition 14.6 for USB clock selection options.

Some example USB clock configurations for Full and Low Speed mode are given below:

Internal Oscillator								
Clock Signal Input Source Selection Register Bit Setting								
USB Clock	Clock Multiplier	USBCLK = 000b						
Clock Multiplier Input	Internal Oscillator*	MULSEL = 00b						
Internal Oscillator	Divide by 1	IFCN = 11b						
	External Oscillator							
Clock Signal	Input Source Selection	Register Bit Settings						
USB Clock	Clock Multiplier	USBCLK = 000b						
Clock Multiplier Input	External Oscillator	MULSEL = 01b						
External Oscillator	Crystal Oscillator Mode 12 MHz Crystal	XOSCMD = 110b XFCN = 111b						

\*Note: Clock Recovery must be enabled for this configuration.

Internal Oscillator								
Clock Signal	Input Source Selection	Register Bit Settings						
USB Clock	Internal Oscillator / 2	USBCLK = 001b						
Internal Oscillator	Divide by 1	IFCN = 11b						
	External Oscillator	•						
Clock Signal Input Source Selection Register Bit Settings								



Internal Oscillator									
Clock Signal	Input Source Selection	Register Bit Settings							
USB Clock	External Oscillator / 4	USBCLK = 101b							
External Oscillator	Crystal Oscillator Mode 24 MHz Crystal	XOSCMD = 110b XFCN = 111b							

## SFR Definition 14.6. CLKSEL: Clock Select

	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
	-		USBCLK		-		CLKSL		00000000
-	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address
									0xA9

Bit 7: Unused. Read = 0b; Write = don't care.

Bits6–4: USBCLK2–0: USB Clock Select

These bits select the clock supplied to USB0. When operating USB0 in full-speed mode, the selected clock should be 48 MHz. When operating USB0 in low-speed mode, the selected clock should be 6 MHz.

USBCLK	Selected Clock			
000	4x Clock Multiplier			
001	Internal Oscillator / 2			
010	External Oscillator			
011	External Oscillator / 2			
100	External Oscillator / 3			
101	External Oscillator / 4			
110	RESERVED			
111	RESERVED			

Bit3: Unused. Read = 0b; Write = don't care.

Bits2–0: CLKSL2–0: System Clock Select

These bits select the system clock source. When operating from a system clock of 25 MHz or less, the FLRT bit should be set to '0'. When operating with a system clock of greater than 25 MHz (up to 48 MHz), the FLRT bit (FLSCL.4) should be set to '1'. See Section "10. Prefetch Engine" on page 99 for more details.

CLKSL	Selected Clock			
000	Internal Oscillator (as determined by the IFCN bits in register OSCICN)			
001	External Oscillator			
010	4x Clock Multiplier / 2			
011*	4x Clock Multiplier*			
100	Low-Frequency Oscillator			
101-111	RESERVED			
*Note: This option is only available on 48 MHz devices.				



## **Table 14.1. Oscillator Electrical Characteristics**

#### $V_{DD}$ = 2.7 to 3.6 V; -40 to +85 °C unless otherwise specified

Parameter	Conditions	Min	Тур	Max	Units		
Internal High-Frequency Oscillator (Using Factory-Calibrated Settings)							
Oscillator Frequency	IFCN = 11b	11.82	12.00	12.18	MHz		
Oscillator Supply Current (from V <sub>DD</sub> )	24 °C, V <sub>DD</sub> = 3.0 V, OSCICN.7 = 1	_	685	—	μA		
Internal Low-Frequency Oscillator (Using Factory-Calibrated Settings)							
Oscillator Frequency	OSCLD = 11b	72	80	99	kHz		
Oscillator Supply Current (from V <sub>DD</sub> )	24 °C, V <sub>DD</sub> = 3.0 V, OSCLCN.7 = 1	_	7.0	_	μA		
External USB Clock Requirements							
USB Clock Frequency*	Full Speed Mode	47.88	48	48.12	N 41 I		
	Low Speed Mode	5.91	6	6.09	MHz		

\*Note: Applies only to external oscillator sources.

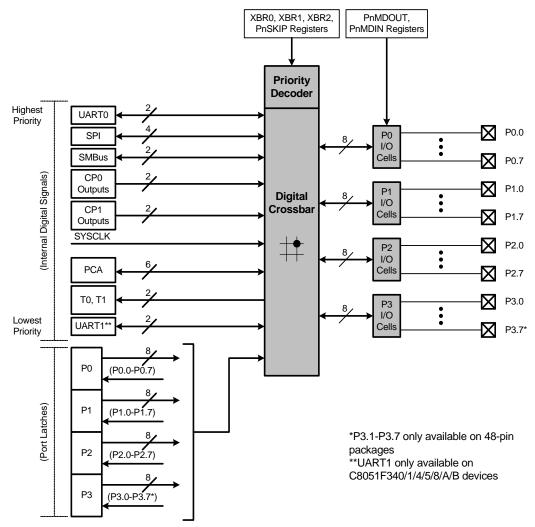


## **15. Port Input/Output**

Digital and analog resources are available through 40 I/O pins (48-pin packages) or 25 I/O pins (32-pin packages). Port pins are organized as shown in Figure 15.1. Each of the Port pins can be defined as general-purpose I/O (GPIO) or analog input; Port pins P0.0-P3.7 can be assigned to one of the internal digital resources as shown in Figure 15.3. The designer has complete control over which functions are assigned, limited only by the number of physical I/O pins. This resource assignment flexibility is achieved through the use of a Priority Crossbar Decoder. Note that the state of a Port I/O pin can always be read in the corresponding Port latch, regardless of the Crossbar settings.

The Crossbar assigns the selected internal digital resources to the I/O pins based on the Priority Decoder (Figure 15.3 and Figure 15.4). The registers XBR0, XBR1, and XBR2 defined in SFR Definition 15.1, SFR Definition 15.2, and SFR Definition 15.3, are used to select internal digital functions.

All Port I/Os are 5 V tolerant (refer to Figure 15.2 for the Port cell circuit). The Port I/O cells are configured as either push-pull or open-drain in the Port Output Mode registers (PnMDOUT, where n = 0, 1, 2, 3, 4). Complete Electrical Specifications for Port I/O are given in Table 15.1 on page 158.







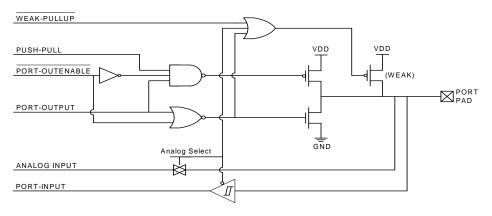


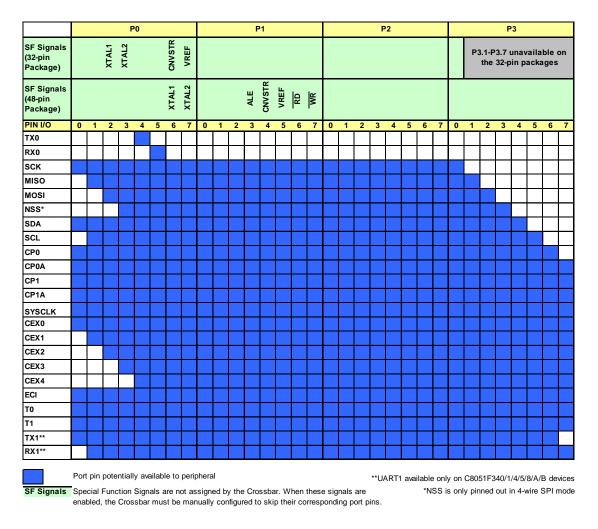
Figure 15.2. Port I/O Cell Block Diagram



### **15.1.** Priority Crossbar Decoder

The Priority Crossbar Decoder (Figure 15.3) assigns a priority to each I/O function, starting at the top with UART0. When a digital resource is selected, the least-significant unassigned Port pin is assigned to that resource (excluding UART0, which is always at pins 4 and 5). If a Port pin is assigned, the Crossbar skips that pin when assigning the next selected resource. Additionally, the Crossbar will skip Port pins whose associated bits in the PnSKIP registers are set. The PnSKIP registers allow software to skip Port pins that are to be used for analog input, dedicated functions, or GPIO.

**Important Note on Crossbar Configuration:** If a Port pin is claimed by a peripheral without use of the Crossbar, its corresponding PnSKIP bit should be set. This applies to the VREF signal, external oscillator pins (XTAL1, XTAL2), the ADC's external conversion start signal (CNVSTR), EMIF control signals, and any selected ADC or Comparator inputs. The PnSKIP registers may also be used to skip pins to be used as GPIO. The Crossbar skips selected pins as if they were already assigned, and moves to the next unassigned pin. Figure 15.3 shows all the possible pins available to each peripheral. Figure 15.4 shows the Crossbar Decoder priority with no Port pins skipped. Figure 15.5 shows a Crossbar example with pins P0.2, P0.3, and P1.0 skipped.







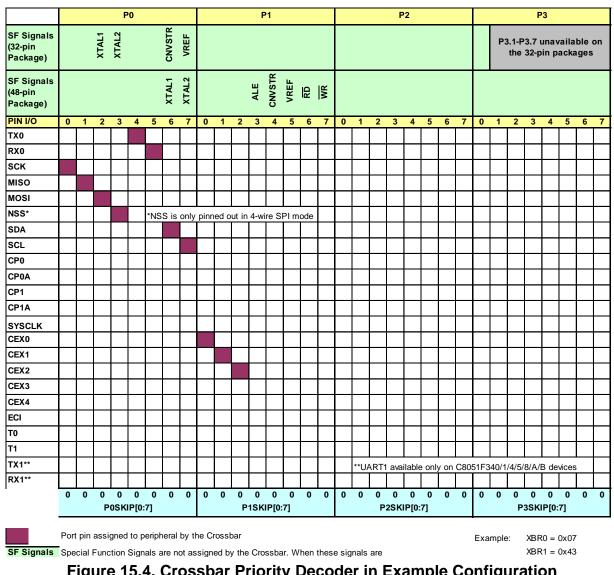
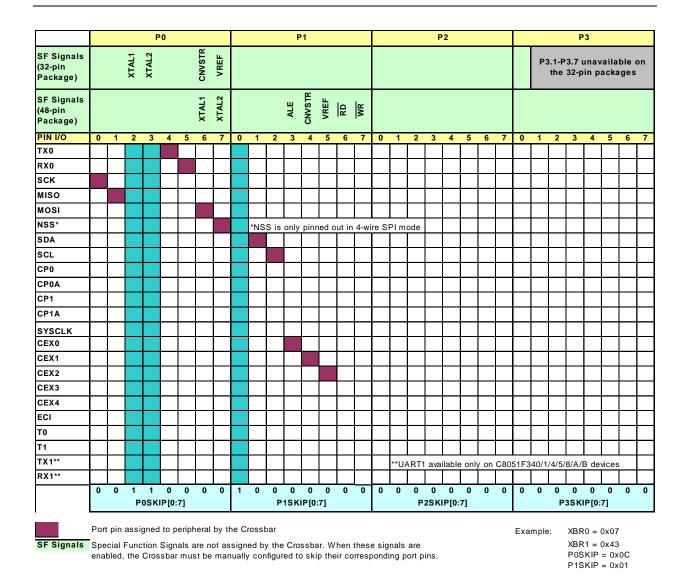


Figure 15.4. Crossbar Priority Decoder in Example Configuration (No Pins Skipped)





# Figure 15.5. Crossbar Priority Decoder in Example Configuration (3 Pins Skipped)

Registers XBR0, XBR1, and XBR2 are used to assign the digital I/O resources to the physical I/O Port pins. Note that when the SMBus is selected, the Crossbar assigns both pins associated with the SMBus (SDA and SCL); when either UART is selected, the Crossbar assigns both pins associated with the UART (TX and RX). UART0 pin assignments are fixed for bootloading purposes: UART TX0 is always assigned to P0.4; UART RX0 is always assigned to P0.5. Standard Port I/Os appear contiguously after the prioritized functions have been assigned.

**Important Note:** The SPI can be operated in either 3-wire or 4-wire modes, depending on the state of the NSSMD1-NSSMD0 bits in register SPI0CN. According to the SPI mode, the NSS signal may or may not be routed to a Port pin.



#### 15.2. Port I/O Initialization

Port I/O initialization consists of the following steps:

- Step 1. Select the input mode (analog or digital) for all Port pins, using the Port Input Mode register (PnMDIN).
- Step 2. Select the output mode (open-drain or push-pull) for all Port pins, using the Port Output Mode register (PnMDOUT).
- Step 3. Select any pins to be skipped by the I/O Crossbar using the Port Skip registers (PnSKIP).
- Step 4. Assign Port pins to desired peripherals (XBR0, XBR1).
- Step 5. Enable the Crossbar (XBARE = '1').

All Port pins must be configured as either analog or digital inputs. Any pins to be used as Comparator or ADC inputs should be configured as an analog inputs. When a pin is configured as an analog input, its weak pull-up, digital driver, and digital receiver are disabled. This process saves power and reduces noise on the analog input. Pins configured as digital inputs may still be used by analog peripherals; however this practice is not recommended. To configure a Port pin for digital input, write '0' to the corresponding bit in register PnMDOUT, and write '1' to the corresponding Port latch (register Pn).

Additionally, all analog input pins should be configured to be skipped by the Crossbar (accomplished by setting the associated bits in PnSKIP). Port input mode is set in the PnMDIN register, where a '1' indicates a digital input, and a '0' indicates an analog input. All pins default to digital inputs on reset.

The output driver characteristics of the I/O pins are defined using the Port Output Mode registers (PnMD-OUT). Each Port Output driver can be configured as either open drain or push-pull. This selection is required even for the digital resources selected in the XBRn registers, and is not automatic. The only exception to this is the SMBus (SDA, SCL) pins, which are configured as open-drain regardless of the PnMDOUT settings. When the WEAKPUD bit in XBR1 is '0', a weak pull-up is enabled for all Port I/O configured as open-drain. WEAKPUD does not affect the push-pull Port I/O. Furthermore, the weak pull-up is turned off on an output that is driving a '0' to avoid unnecessary power dissipation.

Registers XBR0 and XBR1 must be loaded with the appropriate values to select the digital I/O functions required by the design. Setting the XBARE bit in XBR1 to '1' enables the Crossbar. Until the Crossbar is enabled, the external pins remain as standard Port I/O (in input mode), regardless of the XBRn Register settings. For given XBRn Register settings, one can determine the I/O pin-out using the Priority Decode Table; as an alternative, the Configuration Wizard utility of the Silicon Labs IDE software will determine the Port I/O pin-assignments based on the XBRn Register settings.

**Important Note:** The Crossbar must be enabled to use Ports P0, P1, P2, and P3 as standard Port I/O in output mode. These Port output drivers are disabled while the Crossbar is disabled. Port 4 always functions as standard GPIO.



5.44	544	5.44	<b>5</b> 4 4	<b>B</b> 444	544	5.44	5	<b>D</b>				
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value				
CP1AE		CP0AE	CP0E	SYSCKE	SMB0E	SPI0E	URT0E	00000000				
Bit7	Bit6 Bit5 Bit4 Bit3 Bit2 Bit1 Bit0							SFR Address:				
								0xE1				
Bit7:		CP1AE: Comparator1 Asynchronous Output Enable										
	0: Asynchro			•								
	1: Asynchro			•								
Bit6:	CP1E: Com		•	e								
	0: CP1 unav		•									
	1: CP1 route			_								
Bit5:	CP0AE: Cor	•		•	nable							
	0: Asynchro											
	1: Asynchro			•								
Bit4:	CP0E: Com		•	e								
	0: CP0 unav		•									
	1: CP0 route											
Bit3:	SYSCKE: /S		•									
	0: /SYSCLK											
	1: /SYSCLK	•		oin.								
Bit2:	SMB0E: SM											
	0: SMBus I/0			ins.								
	1: SMBus I/0		Port pins.									
Bit1:	SPI0E: SPI I											
	0: SPI I/O ur		•									
	1: SPI I/O ro		•									
Bit0:	URTOE: UAR		•									
	0: UARTO I/(											
	1: UART0 T	XU, RXU rol	ited to Port	pins P0.4 a	nd P0.5.							

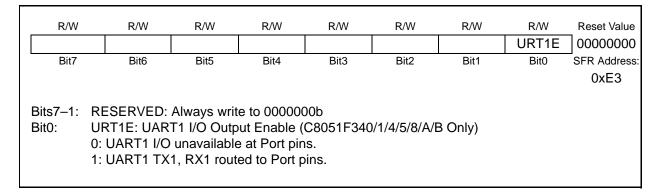
## SFR Definition 15.1. XBR0: Port I/O Crossbar Register 0



SFR Definition 15.2. XBR1: Port I/O Crossbar Register	1
---	---

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value				
WEAKP	JD XBARE	T1E	T0E	ECIE		PCA0ME		00000000				
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xE2				
Bit7:	WEAKPUD:   0: Weak Pull- push-pull out	ups enable			se I/O are	configured as	s analog	input or				
	1: Weak Pull-ups disabled.											
Bit6:	XBARE: Crossbar Enable.											
	0: Crossbar disabled; all Port drivers disabled.											
	1: Crossbar enabled.											
Bit5:	T1E: T1 Enable											
	0: T1 unavailable at Port pin.											
	1: T1 routed to Port pin.											
Bit4:	T0E: T0 Enable											
	0: T0 unavailable at Port pin.											
Dire	1: T0 routed to Port pin.											
Bit3:	ECIE: PCA0 External Counter Input Enable											
	0: ECI unavailable at Port pin.											
D:+=0_0.	1: ECI routed to Port pin. PCA0ME: PCA Module I/O Enable Bits.											
Bits2–0:												
	000: All PCA			pins.								
	001: CEX0 ro		•	<u> </u>								
	010: CEX0, 0 011: CEX0, 0											
	100: CEX0, C				aina							
	100. CEX0, C					<b>^</b>						
	110: Reserve	,	, CEA3, CE		o Port pina	5.						
	111: Reserve											
	III. Reserve	u.										

## SFR Definition 15.3. XBR2: Port I/O Crossbar Register 2





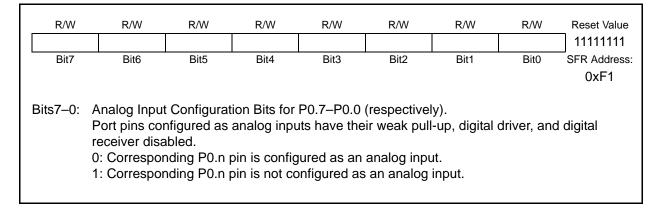
### 15.3. General Purpose Port I/O

Port pins that remain unassigned by the Crossbar and are not used by analog peripherals can be used for general purpose I/O. Ports 3-0 are accessed through corresponding special function registers (SFRs) that are both byte addressable and bit addressable. Port 4 (48-pin packages only) uses an SFR which is byte-addressable. When writing to a Port, the value written to the SFR is latched to maintain the output data value at each pin. When reading, the logic levels of the Port's input pins are returned regardless of the XBRn settings (i.e., even when the pin is assigned to another signal by the Crossbar, the Port register can always read its corresponding Port I/O pin). The exception to this is the execution of the read-modify-write instructions. The read-modify-write instructions when operating on a Port SFR are the following: ANL, ORL, XRL, JBC, CPL, INC, DEC, DJNZ and MOV, CLR or SETB, when the destination is an individual bit in a Port SFR. For these instructions, the value of the register (not the pin) is read, modified, and written back to the SFR.

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value		
P0.7	P0.6	P0.5	P0.4	P0.3	P0.2	P0.1	P0.0	11111111		
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:		
	(bit addressable) 0x80									
Bits7–0:	P0.[7:0] Write - Outp 0: Logic Low 1: Logic Higł Read - Alwa pin when cou 0: P0.n pin is 1: P0.n pin is	y Output. n Output (hi ys reads '0' nfigured as s logic low.	gh impedar if selected digital input	nce if corres as analog i	ponding P(	)MDOUT.n l	, bit = 0).			

#### SFR Definition 15.4. P0: Port0 Latch

### SFR Definition 15.5. P0MDIN: Port0 Input Mode





R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value 0000000			
Bit7	Bit6 Bit5 Bit4 Bit3 Bit2 Bit1 Bit0 SFR Addre 0xA4										
Bits7–0:	Output Confi ter P0MDIN 0: Correspor 1: Correspor (Note: When of the value of	s logic 0. Iding P0.n Iding P0.n SDA and S	Output is op Output is pu SCL appear	ben-drain. Jsh-pull.							

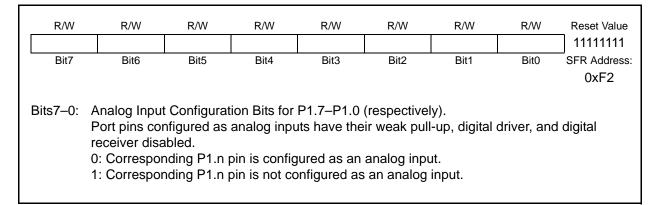
# SFR Definition 15.7. P0SKIP: Port0 Skip

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
								0xD4
	These bits se log inputs (fo lator circuit, 0 0: Correspon	r ADC or C CNVSTR ir	Comparator)	or used as be skipped	special fund I by the Cros	ctions (VRE ssbar.		

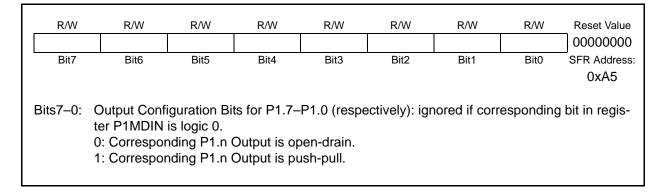


R/W P1.7	R/W P1.6	R/W P1.5	R/W P1.4	R/W P1.3	R/W P1.2	R/W P1.1	R/W P1.0	Reset Value 11111111
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1 (bit	Bit0 addressable	SFR Address: 0x90
Bits7–0:	P1.[7:0] Write - Outp 0: Logic Low 1: Logic Higl Read - Alwa pin when co 0: P1.n pin is 1: P1.n pin is	<ul> <li>Output.</li> <li>Output (hi ys reads '0' nfigured as s logic low.</li> </ul>	gh impedar if selected digital inpu	nce if corres as analog i	ponding P1	IMDOUT.n I	oit = 0).	reads Port

#### SFR Definition 15.9. P1MDIN: Port1 Input Mode



### SFR Definition 15.10. P1MDOUT: Port1 Output Mode





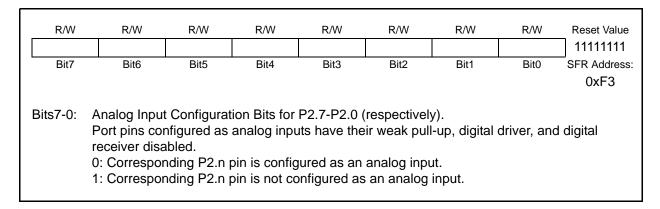
# SFR Definition 15.11. P1SKIP: Port1 Skip

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value 0000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xD5
	P1SKIP[7:0]: These bits se log inputs (fo lator circuit, ( 0: Correspor 1: Correspor	elect Port p r ADC or C CNVSTR in iding P1.n p	ins to be sk omparator) put) should pin is not sk	ipped by the or used as be skipped ipped by the	e Crossbar special fund by the Cro e Crossbar.	ctions (VRE ssbar.		sed as ana- xternal oscil-

# SFR Definition 15.12. P2: Port2 Latch

R/W P2.7	R/W P2.6	R/W P2.5	R/W P2.4	R/W P2.3	R/W P2.2	R/W P2.1	R/W P2.0	Reset Value 11111111			
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:			
	(bit addressable) 0xA0										
Bits7–0:	P2.[7:0] Write - Outpu 0: Logic Low 1: Logic High Read - Alwa pin when cou 0: P2.n pin is 1: P2.n pin is	o Output. In Output (hi ys reads '0' Infigured as Is logic low.	gh impedar if selected digital input	nce if corres as analog i	ponding P2	2MDOUT.n l	bit = 0).				

### SFR Definition 15.13. P2MDIN: Port2 Input Mode





### SFR Definition 15.14. P2MDOUT: Port2 Output Mode

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xA6
t (	Output Confi er P2MDIN 0: Correspor 1: Correspor	is logic 0. Iding P2.n (	Output is op	en-drain.	ctively): igr	nored if corre	espondinç	g bit in regis-

# SFR Definition 15.15. P2SKIP: Port2 Skip

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xD6
Bits7–0:	P2SKIP[7:0] These bits se log inputs (fc lator circuit, 0 0: Correspor 1: Correspor	elect Port p or ADC or C CNVSTR in nding P2.n	ins to be sk comparator) put) should pin is not sk	ipped by the or used as l be skipped kipped by th	e Crossbar special fund by the Cro e Crossbar.	ctions (VRE ssbar.		



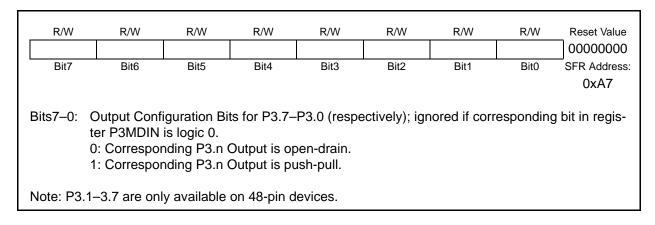
R/W P3.7	R/W P3.6	R/W P3.5	R/W P3.4	R/W P3.3	R/W P3.2	R/W P3.1	R/W P3.0	Reset Value	
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:	
	(bit addressable) 0xB0								
Bits7–0:	P3.[7:0] Write - Outpu 0: Logic Low 1: Logic High Read - Alway pin when cor 0: P3.n pin is 1: P3.n pin is	output. Output (hi ys reads '0' ofigured as s logic low.	gh impedar if selected digital input	as analog iı			,	reads Port	
Note: P3.	.1–3.7 are onl	y available	on 48-pin d	evices.					

# SFR Definition 15.16. P3: Port3 Latch

# SFR Definition 15.17. P3MDIN: Port3 Input Mode

R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
							11111111
Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address
							0xF4
receiver disa	•			ii weak pui	up, uigitai	unver, and	a digital
	Analog Input	Analog Input Configura	Analog Input Configuration Bits for	Analog Input Configuration Bits for P3.7–P3.0	Analog Input Configuration Bits for P3.7–P3.0 (respective)	Analog Input Configuration Bits for P3.7–P3.0 (respectively).	

## SFR Definition 15.18. P3MDOUT: Port3 Output Mode





## SFR Definition 15.19. P3SKIP: Port3 Skip

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xDF
	P3SKIP[3:0]: These bits se log inputs (fo lator circuit, ( 0: Correspor 1: Correspor	elect Port p or ADC or C CNVSTR in nding P3.n	ins to be sk comparator) put) should pin is not sk	ipped by the or used as be skipped ipped by the	e Crossbar special fund by the Cro e Crossbar.	ctions (VRE ssbar.	•	sed as ana- xternal oscil-
Note: P3.1	–3.7 are only	y available	on 48-pin d	evices.				

SFR Definition 15.20. P4: Port4 Latch

P4.7P4.6P4.5P4.4P4.3P4.2P4.1P4.01111111Bit7Bit6Bit5Bit4Bit3Bit2Bit1Bit0SFR Address: 0xC7Bits7-0:P4.[7:0] Write - Output appears on I/O pins. 0: Logic Low Output. 1: Logic High Output (high impedance if corresponding P4MDOUT.n bit = 0). Read - Always reads '0' if selected as analog input in register P4MDIN. Directly reads Port pin when configured as digital input. 0: P4.n pin is logic low. 1: P4.n pin is logic high.	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
0xC7 Bits7–0: P4.[7:0] Write - Output appears on I/O pins. 0: Logic Low Output. 1: Logic High Output (high impedance if corresponding P4MDOUT.n bit = 0). Read - Always reads '0' if selected as analog input in register P4MDIN. Directly reads Port pin when configured as digital input. 0: P4.n pin is logic low.	P4.7	P4.6	P4.5	P4.4	P4.3	P4.2	P4.1	P4.0	11111111
<ul> <li>Bits7–0: P4.[7:0]</li> <li>Write - Output appears on I/O pins.</li> <li>0: Logic Low Output.</li> <li>1: Logic High Output (high impedance if corresponding P4MDOUT.n bit = 0).</li> <li>Read - Always reads '0' if selected as analog input in register P4MDIN. Directly reads Port pin when configured as digital input.</li> <li>0: P4.n pin is logic low.</li> </ul>	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
<ul> <li>Write - Output appears on I/O pins.</li> <li>0: Logic Low Output.</li> <li>1: Logic High Output (high impedance if corresponding P4MDOUT.n bit = 0).</li> <li>Read - Always reads '0' if selected as analog input in register P4MDIN. Directly reads Port pin when configured as digital input.</li> <li>0: P4.n pin is logic low.</li> </ul>									0xC7
	Bits7–0:	Write - Outpu 0: Logic Low 1: Logic High Read - Alway pin when cor 0: P4.n pin is	Output. n Output (hi ys reads '0' nfigured as s logic low.	gh impedar if selected digital input	nce if corres as analog i				reads Port



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address 0xF5
F	Analog Input Port pins con receiver disat ): Correspon	figured as pled.	analog inpu	ts have thei ured as an a	r weak pull	-up, digital (	driver, and	d digital

# SFR Definition 15.21. P4MDIN: Port4 Input Mode

# SFR Definition 15.22. P4MDOUT: Port4 Output Mode

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xAE
Bits7–0:	Output Confi ter P4MDIN 0: Correspor 1: Correspor	is logic 0. nding P4.n (	Output is op	pen-drain.	ectively); igr	nored if corr	esponding	g bit in regis-
Note: P4	is only availa	ble on 48-p	in devices.					



# Table 15.1. Port I/O DC Electrical Characteristics

#### $V_{DD}$ = 2.7 to 3.6 V, -40 to +85 °C unless otherwise specified

Parameters	Conditions	Min	Тур	Max	Units
	I <sub>OH</sub> = –3 mA, Port I/O push-pull	V <sub>DD</sub> – 0.7			
Output High Voltage	$I_{OH} = -10 \ \mu A$ , Port I/O push-pull	V <sub>DD</sub> – 0.1			V
	I <sub>OH</sub> = -10 mA, Port I/O push-pull		V <sub>DD</sub> – 0.8		
	I <sub>OL</sub> = 8.5 mA			0.6	
Output Low Voltage	I <sub>OL</sub> = 10 μA			0.1	V
	I <sub>OL</sub> = 25 mA		1.0		
Input High Voltage		2.0			V
Input Low Voltage				0.8	V
Input Leakage Current	Weak Pull-up Off			±1	
mput Leakage Current	Weak Pull-up On, $V_{IN} = 0 V$		25	50	μA



# 16. Universal Serial Bus Controller (USB0)

C8051F34x devices include a complete Full/Low Speed USB function for USB peripheral implementations\*. The USB Function Controller (USB0) consists of a Serial Interface Engine (SIE), USB Transceiver (including matching resistors and configurable pull-up resistors), 1k FIFO block, and clock recovery mechanism for crystal-less operation. No external components are required. The USB Function Controller and Transceiver is Universal Serial Bus Specification 2.0 compliant.

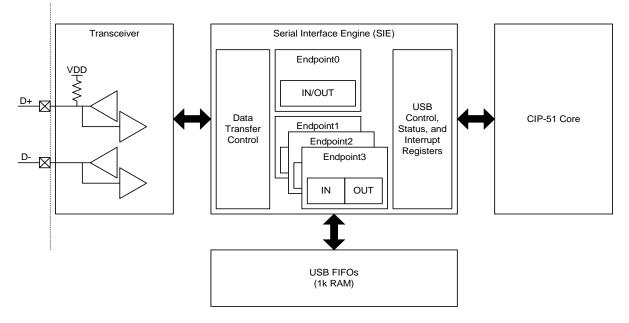


Figure 16.1. USB0 Block Diagram

*Important Note:* This document assumes a comprehensive understanding of the USB Protocol. Terms and abbreviations used in this document are defined in the USB Specification. We encourage you to review the latest version of the USB Specification before proceeding.

\*Note: The C8051F34x cannot be used as a USB Host device.



#### 16.1. Endpoint Addressing

A total of eight endpoint pipes are available. The control endpoint (Endpoint0) always functions as a bi-directional IN/OUT endpoint. The other endpoints are implemented as three pairs of IN/OUT endpoint pipes:

Endpoint	Associated Pipes	USB Protocol Address		
Endpoint0	Endpoint0 IN	0x00		
Lindpolitio	Endpoint0 OUT	0x00		
Endpoint1	Endpoint1 IN	0x81		
LIndpoint	Endpoint1 OUT	0x01		
Endpoint2	Endpoint2 IN	0x82		
LIndpointz	Endpoint2 OUT	0x02		
Endpoint3	Endpoint3 IN	0x83		
LIndpoints	Endpoint3 OUT	0x03		

 Table 16.1. Endpoint Addressing Scheme

#### 16.2. USB Transceiver

The USB Transceiver is configured via the USB0XCN register shown in SFR Definition 16.1. This configuration includes Transceiver enable/disable, pull-up resistor enable/disable, and device speed selection (Full or Low Speed). When bit SPEED = '1', USB0 operates as a Full Speed USB function, and the on-chip pull-up resistor (if enabled) appears on the D+ pin. When bit SPEED = '0', USB0 operates as a Low Speed USB function, and the on-chip pull-up resistor (if enabled) appears on the D- pin. Bits4-0 of register USB0XCN can be used for Transceiver testing as described in SFR Definition 16.1. The pull-up resistor is enabled only when VBUS is present (see Section "8.2. VBUS Detection" on page 69 for details on VBUS detection).

Important Note: The USB clock should be active before the Transceiver is enabled.



#### SFR Definition 16.1. USB0XCN: USB0 Transceiver Control

R/W	R/W	R/W	R/W	R/W	R	R	R	Reset Value
PREN	PHYEN	SPEED	PHYTST1	PHYTST0	DFREC	Dp	Dn	0000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Addres 0xD7
Bit7:	PREN: Interr	•						
	The location							
	0: Internal pu							
	1: Internal pu	ili-up resis	tor enabled	when VBUS	is present	(device a	ittached to t	the USB net
Bit6:	work). PHYEN: Phy	sical Lave	r Enable					
Dito.	This bit enab			) physical lay	ver transcei	ver.		
	0: Transceive							
	1: Transceive		• • •					
Bit5:	SPEED: USE							
	This bit selec							
	0: USB0 ope		Low Speed	device. If en	abled, the	internal p	oull-up resis	tor appears
	on the D– lin 1: USB0 ope	-	Full Speed o	dovido Ifono	blad thair	tornal pu	ll un rogiete	r oppoore o
			ruii Speeu u	levice. Il ella	ibieu, irie ir	itemai pu	II-up resisio	JI appears 0
			•			•	•	
Bits4–3 <sup>.</sup>	the D+ line.		·			·	·	
Bits4–3:		: Physical	Layer Test			·	·	
Bits4–3:	the D+ line. PHYTST1–0 These bits ca	: Physical an be used	Layer Test to test the l	USB0 transc	eiver.		·	
Bits4–3:	the D+ line. PHYTST1-0 These bits ca	: Physical an be used <b>0]</b>	Layer Test I to test the I	USB0 transc <b>de</b>	eiver.	D-	·	
Bits4–3:	the D+ line. PHYTST1-0 These bits ca PHYTST[1: 00b	: Physical an be used <b>0]</b> Mode	Layer Test to test the <b>Mo</b> 0: Normal (r	USB0 transc <b>de</b> non-test moo	eiver. D+ le) X	<b>D-</b> X		
Bits4–3:	the D+ line. PHYTST1-0 These bits ca PHYTST[1: 00b 01b	: Physical an be used 0] Mode Mode	Layer Test to test the <b>Mo</b> 0: Normal (r 1: Differenti	USB0 transc de non-test moc al '1' Forced	eiver. D+ de) X 1	<b>D</b> – X 0		
Bits4–3:	the D+ line. PHYTST1-0 These bits ca PHYTST[1: 00b 01b 10b	: Physical an be used 0] Mode Mode Mode	Layer Test to test the <b>Mo</b> 0: Normal (r 1: Differenti 2: Differenti	USB0 transc de non-test mod al '1' Forced al '0' Forced	eiver. D+ de) X 1 0	<b>D-</b> X 0 1		
Bits4–3:	the D+ line. PHYTST1-0 These bits ca PHYTST[1: 00b 01b	: Physical an be used 0] Mode Mode Mode	Layer Test to test the <b>Mo</b> 0: Normal (r 1: Differenti 2: Differenti	USB0 transc de non-test moc al '1' Forced	eiver. D+ de) X 1 0	<b>D</b> – X 0		
Bits4–3: Bit2:	the D+ line. PHYTST1-0 These bits ca PHYTST[1: 00b 01b 10b	: Physical an be used 0] Mode Mode Mode Mode	Layer Test to test the I 0: Normal (r 1: Differenti 2: Differenti 3: Single-Er	USB0 transc de non-test mod al '1' Forced al '0' Forced	eiver. D+ de) X 1 0	<b>D-</b> X 0 1		
	the D+ line. PHYTST1-0 These bits ca <b>PHYTST[1:</b> 00b 01b 10b 11b	: Physical an be used 0] Mode Mode Mode Mode	Layer Test to test the l 0: Normal (r 1: Differenti 2: Differenti 3: Single-Er	USB0 transc de non-test moc al '1' Forced al '0' Forced nded '0' Forc	eiver. <b>D+</b> le) X 1 0 sed 0	<b>D</b>		
	the D+ line. PHYTST1-0 These bits ca PHYTST[1: 00b 01b 10b 11b DFREC: Diffe The state of t when PHYEN	: Physical an be used 0] Mode Mode Mode Mode erential Re this bit indi N = '1'.	Layer Test to test the l Mo 0: Normal (r 1: Differenti 2: Differenti 3: Single-Er eceiver icates the cu	USB0 transc de non-test mod al '1' Forced al '0' Forced nded '0' Force	eiver. <b>D+</b> le) X 1 0 sed 0	<b>D</b>		
	the D+ line. PHYTST1-0 These bits ca PHYTST[1: 00b 01b 10b 11b DFREC: Diffe The state of twhen PHYEN 0: Differentia	: Physical an be used 0] Mode Mode Mode Mode erential Re this bit indi N = '1'. I '0' signali	Layer Test to test the l Mo 0: Normal (n 1: Differenti 2: Differenti 3: Single-Er eceiver icates the cu	USB0 transc de non-test moc al '1' Forced al '0' Forced nded '0' Forc urrent differe us.	eiver. <b>D+</b> le) X 1 0 sed 0	<b>D</b>		
Bit2:	the D+ line. PHYTST1-0 These bits ca PHYTST[1: 00b 01b 10b 11b DFREC: Diffe The state of twhen PHYEN 0: Differentia 1: Differentia	Physical an be used 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0	Layer Test to test the l Mo 0: Normal (n 1: Differenti 2: Differenti 3: Single-Er eceiver icates the cu	USB0 transc de non-test moc al '1' Forced al '0' Forced nded '0' Forc urrent differe us.	eiver. <b>D+</b> le) X 1 0 sed 0	<b>D</b>		
	the D+ line. PHYTST1-0 These bits ca PHYTST[1: 00b 01b 10b 11b DFREC: Diffe The state of the sta	Physical an be used Mode Mode Mode Mode erential Re this bit indi N = '1'. I '0' signali I '1' signali al Status	Layer Test to test the l Mo 0: Normal (n 1: Differenti 2: Differenti 3: Single-Er eceiver icates the cu	USB0 transo de non-test moo al '1' Forced al '0' Forced nded '0' Force urrent differe us. us.	eiver. de) X 1 0 ed 0 ntial value	<b>D</b>		
Bit2:	the D+ line. PHYTST1-0 These bits ca PHYTST[1: 00b 01b 10b 11b DFREC: Diffe The state of twhen PHYEN 0: Differentia Dp: D+ Signa This bit indica	Physical an be used 0] Mode Mode Mode Mode Mode 100 signali 1 '1' signali al Status ates the cu	Layer Test to test the l <u>Mo</u> 0: Normal (r 1: Differenti 2: Differenti 3: Single-Er eceiver icates the cu ing on the bu urrent logic l	USB0 transo de non-test moo al '1' Forced al '0' Forced nded '0' Force urrent differe us. us.	eiver. de) X 1 0 ed 0 ntial value	<b>D</b>		
Bit2:	the D+ line. PHYTST1-0 These bits ca PHYTST[1: 00b 01b 10b 11b DFREC: Diffe The state of twee PHYEN 0: Differentia 1: Differentia Dp: D+ Signa This bit indica 0: D+ signal	Physical an be used 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0	Layer Test to test the l <u>Mo</u> 0: Normal (r 1: Differenti 2: Differenti 3: Single-Er eceiver icates the cu ing on the bu urrent logic lu t logic 0.	USB0 transo de non-test moo al '1' Forced al '0' Forced nded '0' Force urrent differe us. us.	eiver. de) X 1 0 ed 0 ntial value	<b>D</b>		
Bit2: Bit1:	the D+ line. PHYTST1-0 These bits ca PHYTST[1: 00b 01b 10b 11b DFREC: Diffe The state of t when PHYEN 0: Differentia 1: Differentia Dp: D+ Signa This bit indica 0: D+ signal 1: D+ signal	Physical an be used 0] Mode Mode Mode Mode Mode (1' N = '1'. I '0' signali I '1' signali al Status ates the cu currently a currently a	Layer Test to test the l <u>Mo</u> 0: Normal (r 1: Differenti 2: Differenti 3: Single-Er eceiver icates the cu ing on the bu urrent logic lu t logic 0.	USB0 transo de non-test moo al '1' Forced al '0' Forced nded '0' Force urrent differe us. us.	eiver. de) X 1 0 ed 0 ntial value	<b>D</b>		
Bit2:	the D+ line. PHYTST1-0 These bits ca <b>PHYTST[1:</b> 00b 01b 10b 11b DFREC: Diffe The state of t when PHYEN 0: Differentia 1: Differentia Dp: D+ Signa 0: D+ signal Dn: D- Signa	Physical an be used 0] Mode Mode Mode Mode erential Re this bit indi N = '1'. I '0' signali I '1' signali al Status ates the cu currently a I Status	Layer Test to test the l Mo 0: Normal (n 1: Differenti 2: Differenti 3: Single-Er eceiver icates the cu ing on the bu ing on the bu urrent logic la t logic 0. t logic 1.	USB0 transc de non-test moc al '1' Forced al '0' Forced nded '0' Forc urrent differe us. us. evel of the D	eiver.       D+       le)     X       1     0       ced     0       ntial value       + pin.	<b>D</b>		
Bit2: Bit1:	the D+ line. PHYTST1-0 These bits ca PHYTST[1: 00b 01b 10b 11b DFREC: Diffe The state of t when PHYEN 0: Differentia 1: Differentia Dp: D+ Signa This bit indica 0: D+ signal 1: D+ signal	Physical an be used 0] Mode Mode Mode Mode Mode Mode Mode (1' signali 1'1' signali	Layer Test to test the l <u>Mo</u> 0: Normal (r 1: Differenti 2: Differenti 3: Single-Er eceiver icates the cu ing on the bu urrent logic lu t logic 0. t logic 1.	USB0 transc de non-test moc al '1' Forced al '0' Forced nded '0' Forc urrent differe us. us. evel of the D	eiver.       D+       le)     X       1     0       ced     0       ntial value       + pin.	<b>D</b>		

#### 16.3. USB Register Access

The USB0 controller registers listed in Table 16.2 are accessed through two SFRs: USB0 Address (USB0ADR) and USB0 Data (USB0DAT). The USB0ADR register selects which USB register is targeted



by reads/writes of the USB0DAT register. See Figure 16.2.

Endpoint control/status registers are accessed by first writing the USB register INDEX with the target endpoint number. Once the target endpoint number is written to the INDEX register, the control/status registers associated with the target endpoint may be accessed. See the "Indexed Registers" section of Table 16.2 for a list of endpoint control/status registers.

Important Note: The USB clock must be active when accessing USB registers.

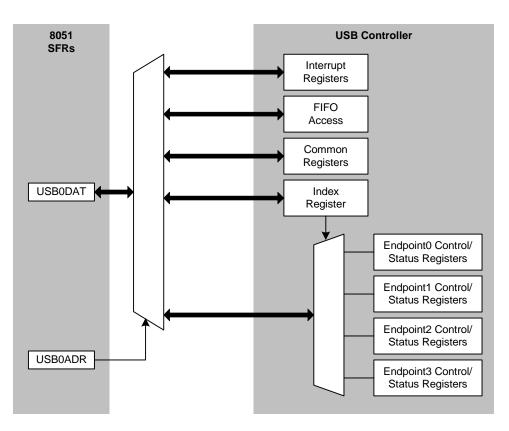


Figure 16.2. USB0 Register Access Scheme



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
BUSY	AUTORD			USBA	DDR			00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0x96
Bits7: Bit6: Bits5–0:	BUSY: USB0 This bit is user initiate a read target address set to '1', hard USB0DAT reg Write: 0: No effect. 1: A USB0 inco Read: 0: USB0DAT in 1: USB0 is bu AUTORD: US This bit is use 0: BUSY musi 1: The next in USB0DAT (US USBADDR: U These bits hol lists the USB0 will target the	d during in of the US s and BUS ware will jister. Soft lirect regis register da sy access B0 Regist d for block t be writte direct regi SBADDR I SB0 Indire d a 6-bit a core regi	direct USB B0 register SY bit may b clear BUSY ware should ster read is ata is valid. ing an indir er Auto-read FIFO read n manually ster read w bits will not ect Registe ddress use sters and th	0 register ac targeted by be written in ' when the t d check BU initiated at t ect register id Flag ls. for each US ill automatic be changed r Address d to indirect heir indirect	the USBAI the same v argeted reg SY for '0' be he address ; USB0DAT GB0 indirect cally be initia d).	DDR bits (U write to USE jister data is efore writing specified b register da register rea ated when s	ISB0ADR. Af B0ADR. Af s ready in g to USB0 by the USB ta is invali ad. software re	[5-0]). The ter BUSY is the DAT. ADDR bits. d. eads s. Table 16.2



USB0DAT       00000         Bit7       Bit6       Bit5       Bit4       Bit3       Bit2       Bit1       Bit0       SFR Add 0x9         This SFR is used to indirectly read and write USB0 registers.         Write Procedure:       1.       Poll for BUSY (USB       0ADR.7) => '0'.       2.       Load the target USB0 register address into the USBADDR bits in register USB0ADR.         3.       Write data to USB0DAT.       4.       Repeat (Step 2 may be skipped when writing to the same USB0 register).         Read Procedure:         1.       Poll for BUSY (USB       0ADR.7) => '0'.         2.       Load the target USB0 register address into the USBADDR bits in register USB0ADR.         Read Procedure:         1.       Poll for BUSY (USB       0ADR.7) => '0'.         2.       Load the target USB0 register address into the USBADDR bits in register USB0ADR.         3.       Write '1' to the BUSY bit in register USB0ADR (steps 2 and 3 can be performed in the same write).         4.       Poll for BUSY (USB       0ADR.7) => '0'.         5.       Read data from USB0DAT.         6.       Repeat from Step 2 (Step 2 may be skipped when reading the same USB0 register; S													
Bit7       Bit6       Bit5       Bit4       Bit3       Bit2       Bit1       Bit0       SFR Add 0x9         This SFR is used to indirectly read and write USB0 registers.         Write Procedure:       1. Poll for BUSY (USB 0ADR.7) => '0'.         2. Load the target USB0 register address into the USBADDR bits in register USB0ADR.         3. Write data to USB0DAT.         4. Repeat (Step 2 may be skipped when writing to the same USB0 register).         Read Procedure:         1. Poll for BUSY (USB 0ADR.7) => '0'.         2. Load the target USB0 register address into the USBADDR bits in register USB0ADR.         3. Write data to USB0DAT.         4. Repeat (Step 2 may be skipped when writing to the same USB0 register).         Read Procedure:         1. Poll for BUSY (USB 0ADR.7) => '0'.         2. Load the target USB0 register address into the USBADDR bits in register USB0ADR.         3. Write '1' to the BUSY bit in register USB0ADR (steps 2 and 3 can be performed in the same write).         4. Poll for BUSY (USB 0ADR.7) => '0'.         5. Read data from USB0DAT.         6. Repeat from Step 2 (Step 2 may be skipped when reading the same USB0 register; S		R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value			
<ul> <li>Ox9</li> <li>This SFR is used to indirectly read and write USB0 registers.</li> <li>Write Procedure: <ol> <li>Poll for BUSY (USB 0ADR.7) =&gt; '0'.</li> <li>Load the target USB0 register address into the USBADDR bits in register USB0ADR.</li> <li>Write data to USB0DAT.</li> <li>Repeat (Step 2 may be skipped when writing to the same USB0 register).</li> </ol> </li> <li>Read Procedure: <ol> <li>Poll for BUSY (USB 0ADR.7) =&gt; '0'.</li> <li>Load the target USB0 register address into the USBADDR bits in register USB0ADR.</li> <li>Write '1' to the BUSY bit in register USB0ADR (steps 2 and 3 can be performed in the same write).</li> <li>Poll for BUSY (USB 0ADR.7) =&gt; '0'.</li> <li>Read data from USB0DAT.</li> <li>Repeat from Step 2 (Step 2 may be skipped when reading the same USB0 register; S</li> </ol> </li> </ul>					USB	DAT				00000000			
<ul> <li>This SFR is used to indirectly read and write USB0 registers.</li> <li>Write Procedure: <ol> <li>Poll for BUSY (USB 0ADR.7) =&gt; '0'.</li> <li>Load the target USB0 register address into the USBADDR bits in register USB0ADR.</li> <li>Write data to USB0DAT.</li> <li>Repeat (Step 2 may be skipped when writing to the same USB0 register).</li> </ol> </li> <li>Read Procedure: <ol> <li>Poll for BUSY (USB 0ADR.7) =&gt; '0'.</li> <li>Load the target USB0 register address into the USBADDR bits in register USB0ADR.</li> <li>Write '1' to the BUSY bit in register USB0ADR (steps 2 and 3 can be performed in the same write).</li> <li>Poll for BUSY (USB 0ADR.7) =&gt; '0'.</li> </ol> </li> <li>Read data from USB0DAT.</li> <li>Repeat from Step 2 (Step 2 may be skipped when reading the same USB0 register; S</li> </ul>		Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:			
<ul> <li>Write Procedure: <ol> <li>Poll for BUSY (USB 0ADR.7) =&gt; '0'.</li> <li>Load the target USB0 register address into the USBADDR bits in register USB0ADR.</li> <li>Write data to USB0DAT.</li> <li>Repeat (Step 2 may be skipped when writing to the same USB0 register).</li> </ol> </li> <li>Read Procedure: <ol> <li>Poll for BUSY (USB 0ADR.7) =&gt; '0'.</li> <li>Load the target USB0 register address into the USBADDR bits in register USB0ADR.</li> <li>Write '1' to the BUSY bit in register USB0ADR (steps 2 and 3 can be performed in the same write).</li> <li>Poll for BUSY (USB 0ADR.7) =&gt; '0'.</li> <li>Read data from USB0DAT.</li> <li>Repeat from Step 2 (Step 2 may be skipped when reading the same USB0 register; S</li> </ol> </li> </ul>													
			Write Proceed 1. Poll for B 2. Load the 3. Write data 4. Repeat (S Read Proceed 1. Poll for B 2. Load the 3. Write '1' to same write 4. Poll for B 5. Read data	dure: USY (USB target USB a to USB0D Step 2 may dure: USY (USB target USB to the BUSY te). USY (USB a from USB	0ADR.7) = 0 register a 0AT. be skipped 0ADR.7) = 0 register a ′ bit in regis 0ADR.7) = 0DAT.	<ul> <li>'0'.</li> <li>ddress into</li> <li>when writir</li> <li>'0'.</li> <li>ddress into</li> <li>iter USB0A</li> <li>&gt; '0'.</li> </ul>	the USBAE og to the sau the USBAE DR (steps 2	DDR bits in r me USB0 re DDR bits in r 2 and 3 can	egister). register U be perfori	SB0ADR. med in the			
	<ol><li>Repeat from Step 2 (Step 2 may be skipped when reading the same USB0 reg may be skipped when the AUTORD bit (USB0ADR.6) is logic 1).</li></ol>												

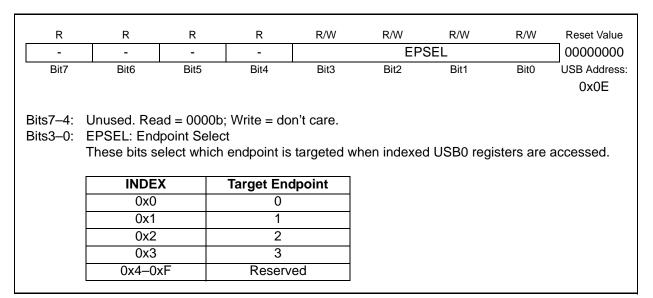
#### SFR Definition 16.3. USB0DAT: USB0 Data



USB Register Name	USB Register Address	Description	Page Number	
Name	Address	Interrupt Registers		
IN1INT	0x02	Endpoint0 and Endpoints1-3 IN Interrupt Flags	173	
OUT1INT	0x02	Endpoints1-3 OUT Interrupt Flags	173	
CMINT	0x06	Common USB Interrupt Flags	174	
IN1IE	0x07	Endpoint0 and Endpoints1-3 IN Interrupt Enables	175	
OUT1IE	0x09	Endpoints1-3 OUT Interrupt Enables	175	
CMIE				
		Common Registers		
FADDR	0x00	Function Address	169	
POWER	0x01	Power Management	171	
FRAMEL	0x0C	Frame Number Low Byte	172	
FRAMEH	0x0D	Frame Number High Byte	172	
INDEX	0x0E	Endpoint Index Selection	165	
CLKREC	0x0F	Clock Recovery Control	166	
FIFOn	0x20-0x23	Endpoints0-3 FIFOs	168	
		Indexed Registers		
E0CSR	0x11	Endpoint0 Control / Status	179	
EINCSRL	UXII	Endpoint IN Control / Status Low Byte	182	
EINCSRH	0x12	Endpoint IN Control / Status High Byte	183	
EOUTCSRL	0x14	Endpoint OUT Control / Status Low Byte	185	
EOUTCSRH	0x15	Endpoint OUT Control / Status High Byte	186	
E0CNT	0x16	Number of Received Bytes in Endpoint0 FIFO	180	
EOUTCNTL		Endpoint OUT Packet Count Low Byte	186	
EOUTCNTH	0x17	Endpoint OUT Packet Count High Byte	186	

#### Table 16.2. USB0 Controller Registers

### USB Register Definition 16.4. INDEX: USB0 Endpoint Index





#### 16.4. USB Clock Configuration

USB0 is capable of communication as a Full or Low Speed USB function. Communication speed is selected via the SPEED bit in SFR USB0XCN. When operating as a Low Speed function, the USB0 clock must be 6 MHz. When operating as a Full Speed function, the USB0 clock must be 48 MHz. Clock options are described in **Section "14. Oscillators" on page 131**. The USB0 clock is selected via SFR CLKSEL (see SFR Definition 14.6).

Clock Recovery circuitry uses the incoming USB data stream to adjust the internal oscillator; this allows the internal oscillator (and 4x Clock Multiplier) to meet the requirements for USB clock tolerance. Clock Recovery should be used in the following configurations:

Communication Speed	USB Clock	4x Clock Multiplier Input		
Full Speed	4x Clock Multiplier	Internal Oscillator		
Low Speed	Internal Oscillator / 2	N/A		

When operating USB0 as a Low Speed function with Clock Recovery, software must write '1' to the CRLOW bit to enable Low Speed Clock Recovery. Clock Recovery is typically not necessary in Low Speed mode.

Single Step Mode can be used to help the Clock Recovery circuitry to lock when high noise levels are present on the USB network. This mode is not required (or recommended) in typical USB environments.

#### **USB Register Definition 16.5. CLKREC: Clock Recovery Control**

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value			
CRE	CRSSEN	CRLOW			Reserved			00001001			
Bit7	Bit6	Bit6 Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address: 0x0F			
Bit7:	CRE: Clock This bit enab 0: Clock reco 1: Clock reco	oles/disables overy disabl	s the USB ( ed.	clock recov	ery feature.						
Bit6:	•										
Bit5:											
Bits4–0:	Reserved. R		ble. Must V	Vrite = 0100	)1b.						
Note: Th	Note: The USB transceiver must be enabled before enabling Clock Recovery.										



#### 16.5. FIFO Management

1024 bytes of on-chip XRAM are used as FIFO space for USB0. This FIFO space is split between Endpoints0-3 as shown in Figure 16.3. FIFO space allocated for Endpoints1-3 is configurable as IN, OUT, or both (Split Mode: half IN, half OUT).

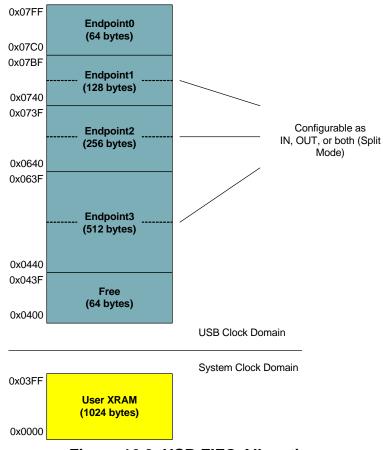


Figure 16.3. USB FIFO Allocation

#### 16.5.1. FIFO Split Mode

The FIFO space for Endpoints1-3 can be split such that the upper half of the FIFO space is used by the IN endpoint, and the lower half is used by the OUT endpoint. For example: if the Endpoint3 FIFO is configured for Split Mode, the upper 256 bytes (0x0540 to 0x063F) are used by Endpoint3 IN and the lower 256 bytes (0x0440 to 0x053F) are used by Endpoint3 OUT.

If an endpoint FIFO is not configured for Split Mode, that endpoint IN/OUT pair's FIFOs are combined to form a single IN *or* OUT FIFO. In this case only one direction of the endpoint IN/OUT pair may be used at a time. The endpoint direction (IN/OUT) is determined by the DIRSEL bit in the corresponding endpoint's EINCSRH register (see SFR Definition 16.20).



#### 16.5.2. FIFO Double Buffering

FIFO slots for Endpoints1-3 can be configured for double-buffered mode. In this mode, the maximum packet size is halved and the FIFO may contain two packets at a time. This mode is available for Endpoints1-3. When an endpoint is configured for Split Mode, double buffering may be enabled for the IN Endpoint and/or the OUT endpoint. When Split Mode is not enabled, double-buffering may be enabled for the entire endpoint FIFO. See Table 16.3 for a list of maximum packet sizes for each FIFO configuration.

Endpoint Number	Split Mode Enabled?	Maximum IN Packet Size (Dou- ble Buffer Disabled / Enabled)	Maximum OUT Packet Size (Double Buffer Disabled / Enabled)				
0	N/A	64	4				
1	N	128	/ 64				
1	Y	64 / 32	64 / 32				
2	N	256 / 128					
2	Y	128 / 64	128 / 64				
3	N	512 /	256				
3	Y	256 / 128	256 / 128				

#### Table 16.3. FIFO Configurations

#### 16.5.1. FIFO Access

Each endpoint FIFO is accessed through a corresponding FIFOn register. A read of an endpoint FIFOn register unloads one byte from the FIFO; a write of an endpoint FIFOn register loads one byte into the endpoint FIFO. When an endpoint FIFO is configured for Split Mode, a read of the endpoint FIFOn register unloads one byte from the OUT endpoint FIFO; a write of the endpoint FIFOn register loads one byte into the IN endpoint FIFO.

### USB Register Definition 16.6. FIFOn: USB0 Endpoint FIFO Access

	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value			
			FIFO	DATA				0000000			
Bit7	Bit6	Bit5	Bit4	it4 Bit3	Bit2	Bit1	Bit0	USB Addres			
								0x20 - 0x2			
USB Addresses 0x20–0x23 provide access to the 4 pairs of endpoint FIFOs:											
_											
	IN/OUT En	ndpoint FIFO	USE	3 Address							
		0		0x20							
Γ		1		0x21							
		2	0x22								
-		3		0x23							



#### 16.6. Function Addressing

The FADDR register holds the current USB0 function address. Software should write the host-assigned 7-bit function address to the FADDR register when received as part of a SET\_ADDRESS command. A new address written to FADDR will not take effect (USB0 will not respond to the new address) until the end of the current transfer (typically following the status phase of the SET\_ADDRESS command transfer). The UPDATE bit (FADDR.7) is set to '1' by hardware when software writes a new address to the FADDR register. Hardware clears the UPDATE bit when the new address takes effect as described above.

#### USB Register Definition 16.7. FADDR: USB0 Function Address

_	-	5.44	-	<b>5</b>	-	-	<b>5</b> * • •	5
R	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Update	<b>;</b>		Fui	nction Addr	ess			0000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address:
								0x00
Bit7: Bits6–0:	Update: Fun Set to '1' who address take 0: The last a 1: The last a Function Add Holds the 7-h the SET_AD takes effect of	en software es effect. ddress writ ddress writ dress bit function DRESS sta	e writes the l ten to FADE ten to FADE address for andard device	DR is in effe DR is not ye USB0. This ce request	ect. It in effect. s address sł s received o	nould be wr	itten by sc	ftware when

### 16.7. Function Configuration and Control

The USB register POWER (SFR Definition 16.8) is used to configure and control USB0 at the device level (enable/disable, Reset/Suspend/Resume handling, etc.).

**USB Reset:** The USBRST bit (POWER.3) is set to '1' by hardware when Reset signaling is detected on the bus. Upon this detection, the following occur:

- 1. The USB0 Address is reset (FADDR = 0x00).
- 2. Endpoint FIFOs are flushed.
- 3. Control/status registers are reset to 0x00 (E0CSR, EINCSRL, EINCSRH, EOUTCSRL, EOUTCSRH).
- 4. USB register INDEX is reset to 0x00.
- 5. All USB interrupts (excluding the Suspend interrupt) are enabled and their corresponding flags cleared.
- 6. A USB Reset interrupt is generated if enabled.

Writing a '1' to the USBRST bit will generate an asynchronous USB0 reset. All USB registers are reset to their default values following this asynchronous reset.

**Suspend Mode:** With Suspend Detection enabled (SUSEN = '1'), USB0 will enter Suspend Mode when Suspend signaling is detected on the bus. An interrupt will be generated if enabled (SUSINTE = '1'). The Suspend Interrupt Service Routine (ISR) should perform application-specific configuration tasks such as disabling appropriate peripherals and/or configuring clock sources for low power modes. See **Section "14. Oscillators" on page 131** for more details on internal oscillator configuration, including the Suspend



mode feature of the internal oscillator.

USB0 exits Suspend mode when any of the following occur: (1) Resume signaling is detected or generated, (2) Reset signaling is detected, or (3) a device or USB reset occurs. If suspended, the internal oscillator will exit Suspend mode upon any of the above listed events.

**Resume Signaling:** USB0 will exit Suspend mode if Resume signaling is detected on the bus. A Resume interrupt will be generated upon detection if enabled (RESINTE = '1'). Software may force a Remote Wakeup by writing '1' to the RESUME bit (POWER.2). When forcing a Remote Wakeup, software should write RESUME = '0' to end Resume signaling 10-15 ms after the Remote Wakeup is initiated (RESUME = '1').

**ISO Update:** When software writes '1' to the ISOUP bit (POWER.7), the ISO Update function is enabled. With ISO Update enabled, new packets written to an ISO IN endpoint will not be transmitted until a new Start-Of-Frame (SOF) is received. If the ISO IN endpoint receives an IN token before a SOF, USB0 will transmit a zero-length packet. When ISOUP = '1', ISO Update is enabled for all ISO endpoints.

**USB Enable:** USB0 is disabled following a Power-On-Reset (POR). USB0 is enabled by clearing the USBINH bit (POWER.4). Once written to '0', the USBINH can only be set to '1' by one of the following: (1) a Power-On-Reset (POR), or (2) an asynchronous USB0 reset generated by writing '1' to the USBRST bit (POWER.3).

Software should perform all USB0 configuration before enabling USB0. The configuration sequence should be performed as follows:

- Step 1. Select and enable the USB clock source.
- Step 2. Reset USB0 by writing USBRST= '1'.
- Step 3. Configure and enable the USB Transceiver.
- Step 4. Perform any USB0 function configuration (interrupts, Suspend detect).
- Step 5. Enable USB0 by writing USBINH =  $0^{\circ}$ .



#### USB Register Definition 16.8. POWER: USB0 Power

R/W	R/W	R/W	R/W	R/W	R/W	R	R/W	Reset Value					
ISOUD	-	-	USBINH	USBRST	RESUME	SUSMD	SUSEN	00010000					
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address:					
								0x01					
Bit7:	ISOUD: ISO	Update											
	This bit affects all IN Isochronous endpoints.												
	0: When software writes INPRDY = '1', USB0 will send the packet when the next IN token is												
	received.												
	1: When software writes INPRDY = '1', USB0 will wait for a SOF token before sending the packet. If an IN token is received before a SOF token, USB0 will send a zero-length data												
	packet.					DO WIN SCH		igin data					
Bits6-5:													
Bit4:	USBINH: US												
		This bit is set to '1' following a power-on reset (POR) or an asynchronous USB0 reset (see Bit3: RESET). Software should clear this bit after all USB0 and transceiver initialization is											
	complete. Sc					and transc		1241101115					
	0: USB0 ena												
	1: USB0 inhi		SB traffic is	ignored.									
Bit3:	USBRST: R€			ah rangua I I		) a a din a thic	hit na roviala						
	Writing '1' to status inform		es an asyno	chronous U	SBU reset. F	keading this	s bit provide	es dus reset					
	Read:												
	0: Reset sigr												
Bit2:	1: Reset sigr RESUME: F			bus.									
DILZ.	Software car			na on the bu	is to wake U	ISB0 from s	suspend mo	ode Writing					
	a '1' to this b												
	naling on the												
	10 ms to15 r SUSMD, who					s generated	d, and hard	ware clears					
Bit1:	SUSMD: Sus			50WE = 0.									
BRIT	Set to '1' by			enters susp	pend mode.	Cleared by	hardware	when soft-					
	ware writes I				wakeup) o	r reads the	CMINT reg	ister after					
	detection of			he bus.									
	0: USB0 not 1: USB0 in s												
Bit0:	SUSEN: Sus	•		е									
	0: Suspend of												
	1: Suspend of on the bus.	detection er	abled. USE	30 will enter	suspend m	ode if it det	ects susper	nd signaling					



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R	R	R	R	R	R	R	R	Reset Value	
Frame Number Low									
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address	
								0x0C	
Bits7-0: Frame Number Low									

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# USB Register Definition 16.10. FRAMEH: USB0 Frame Number High

R	R	R	R	R	R	R	R	Reset Value
-	-	-	-	-	Fram	ne Number	High	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address:
								0x0D
Bits2-0:	Unused. Rea Frame Numb This register	per High By	te	_	ved frame r	number.		

#### 16.8. Interrupts

The read-only USB0 interrupt flags are located in the USB registers shown in USB Register Definition 16.11 through USB Register Definition 16.13. The associated interrupt enable bits are located in the USB registers shown in USB Register Definition 16.14 through USB Register Definition 16.16. A USB0 interrupt is generated when any of the USB interrupt flags is set to '1'. The USB0 interrupt is enabled via the EIE1 SFR (see Section "9.3. Interrupt Handler" on page 88).

Important Note: Reading a USB interrupt flag register resets all flags in that register to '0'.



<b>USB</b> Register	r Definition 16.11	. IN1INT: USB0 IN	Endpoint Interrupt
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	R	R	R	R	R	R	R	R	Reset Value
ΙΓ	-	-	-	-	IN3	IN2	IN1	EP0	00000000
	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address:
									0x02
E	Bit2: Bit1: Bit0:	Unused. Rea IN3: IN Endp This bit is cle 0: IN Endpoi 1: IN Endpoi IN2: IN Endpoi 1: IN Endpoi 1: IN Endpoi 1: IN Endpoi 1: IN Endpoi EP0: Endpoi This bit is cle 0: Endpoint ( 1: Endpoint (	point 3 Inter pared when nt 3 interrup nt 3 interrup ooint 2 Inter point 2 Interrup nt 2 interrup ooint 1 Interrup nt 1 interrup nt 1 interrup nt 0 Interrup pared when D interrupt in	rupt-pendin software re ot inactive. rupt-pendin software re ot inactive. ot active. rupt-pendin software re ot inactive. ot active. ot active. ot active. ot active. ot active. ot active. ot active.	ig Flag eads the IN? g Flag eads the IN? ig Flag eads the IN? Flag	IINT registe	er. er.		UXU2

# USB Register Definition 16.12. OUT1INT: USB0 Out Endpoint Interrupt

R	R	R	R	R	R	R	R	Reset Value			
-	-	-	-	OUT3	OUT2	OUT1	-	00000000			
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address:			
								0x04			
Dito7 4	Unused. Rea	ad 0000b	\//rita_da	n't ooro							
Bit3:											
DIIJ.	OUT3: OUT	•				otor					
	This bit is cle				or man regi	ster.					
	0: OUT Endp		•								
Dire	1: OUT Endp										
Bit2:	OUT2: OUT										
	This bit is cle				JIIINI regi	ster.					
	0: OUT Endp		•								
	1: OUT Endpoint 2 interrupt active.										
Bit1:	OUT1: OUT	•									
	This bit is cle				JT1INT regi	ster.					
	0: OUT Endpoint 1 interrupt inactive.										
	1: OUT Endp		•								
Bit0:	Unused. Rea	ad = 0; Writ	e = don't ca	are.							



### USB Register Definition 16.13. CMINT: USB0 Common Interrupt

R	R	R	R	R	R	R	R	Reset Value
-	-	-	-	SOF	RSTINT	RSUINT	SUSINT	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address:
								0x06
Bit7 Bits7–4: Bit3: Bit2: Bit1: Bit1:	Bit6 Unused. Rea SOF: Start o Set by hardw ware: an inter the actual SO This bit is cle 0: SOF inter 1: SOF inter RSTINT: Res Set by hardw This bit is cle 0: Reset inter 1: Reset inter RSUINT: Re Set by hardw mode. This bit is cle 0: Resume in 1: Resume in SUSINT: Sus When Susper ware when S	ad = 0000b; f Frame Inte vare when a errupt will be DF signal is eared when rupt inactive set Interrup vare when f eared when rrupt active sume Interr vare when f eared when herrupt active sume Interr vare when f eared when herrupt inactive source Interre- vare when f eared when herrupt inactive source Interre- vare when f eared when herrupt inactive source Interre-	Write = do errupt a SOF token generated missed or software re- t-pending F Reset signal software re- ve. upt-pending Resume signal software re- ctive. ve. rupt-pendin is enable	n't care. n is received when hard corrupted. eads the CM lag ling is detect eads the CM g Flag maling is detect eads the CM g Flag d (bit SUSE	d. This inter ware expect 1INT registe tected on the 1INT registe tected on the 1INT registe	errupt event i ets to receive er. bus. er. he bus while er. er POWER)	s synthesiz e a SOF ev e USB0 is in , this bit is s	0x06 red by hard- rent, even if n suspend set by hard-
	reads the CM 0: Suspend i	nterrupt ina	ctive.					
	1: Suspend i	nterrupt act	ive.					



#### USB Register Definition 16.14. IN1IE: USB0 IN Endpoint Interrupt Enable

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value	
-	-	-	-	IN3E	IN2E	IN1E	EP0E	00001111	
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address:	
								0x07	
Bits7–4:	Unused. Rea	ad = 0000b.	. Write = do	n't care.					
Bit3:	IN3E: IN End	point 3 Inte	errupt Enab	le					
	0: IN Endpoi	nt 3 interru	ot disabled.						
	1: IN Endpoir	nt 3 interru	ot enabled.						
Bit2:	IN2E: IN End	point 2 Inte	errupt Enab	le					
	0: IN Endpoi	nt 2 interru	ot disabled.						
	1: IN Endpoi	nt 2 interru	ot enabled.						
Bit1:	IN1E: IN End	dpoint 1 Inte	errupt Enab	le					
	0: IN Endpoint 1 interrupt disabled. 1: IN Endpoint 1 interrupt enabled.								
Bit0:	EP0E: Endpo	oint 0 Interr	upt Enable						
	0: Endpoint (	) interrupt c	lisabled.						
	1: Endpoint (		امعا ما م						

# USB Register Definition 16.15. OUT1IE: USB0 Out Endpoint Interrupt Enable

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value				
-	-	-	-	OUT3E	OUT2E	OUT1E	-	00001110				
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address:				
								0x09				
Bits7–4:	Unused. Rea	ad = 0000b.	Write = do	n't care.								
Bit3:	OUT3E: OUT	Γ Endpoint	3 Interrupt	Enable								
	0: OUT Endp	oint 3 inter	rupt disable	ed.								
	1: OUT End	1: OUT Endpoint 3 interrupt enabled.										
Bit2:	OUT2E: OU		•									
	0: OUT Endp	oint 2 inter	rupt disable	ed.								
	1: OUT Endp		•									
Bit1:	OUT1E: OUT Endpoint 1 Interrupt Enable											
	0: OUT Endpoint 1 interrupt disabled.											
	1: OUT Endpoint 1 interrupt enabled.											



#### USB Register Definition 16.16. CMIE: USB0 Common Interrupt Enable

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value				
-	-	-	-	SOFE	RSTINTE	RSUINTE	SUSINTE	00000110				
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address				
								0x0B				
Bits7–4:	Unused. Rea	ad = 0000b	; Write = do	n't care.								
Bit3:	SOFE: Start	of Frame Ir	nterrupt Ena	able								
	0: SOF inter	rupt disable	ed.									
	1: SOF interrupt enabled.											
Bit2:	RSTINTE: R	leset Interru	ipt Enable									
	0: Reset interrupt disabled.											
	1: Reset interrupt enabled.											
Bit1:	RSUINTE: F	Resume Inte	errupt Enabl	е								
	0: Resume interrupt disabled.											
	1: Resume interrupt enabled.											
Bit0:	SUSINTE: Suspend Interrupt Enable											
	0: Suspend	0: Suspend interrupt disabled.										
	1: Suspend	interrunt on	abled									

### 16.9. The Serial Interface Engine

The Serial Interface Engine (SIE) performs all low level USB protocol tasks, interrupting the processor when data has successfully been transmitted or received. When receiving data, the SIE will interrupt the processor when a complete data packet has been received; appropriate handshaking signals are automatically generated by the SIE. When transmitting data, the SIE will interrupt the processor when a complete data packet has been received; appropriate handshaking signals are automatically generated by the SIE. When transmitting data, the SIE will interrupt the processor when a complete data packet has been transmitted and the appropriate handshake signal has been received.

The SIE will not interrupt the processor when corrupted/erroneous packets are received.

#### 16.10. Endpoint0

Endpoint0 is managed through the USB register E0CSR (USB Register Definition 16.17). The INDEX register must be loaded with 0x00 to access the E0CSR register.

An Endpoint0 interrupt is generated when:

- 1. A data packet (OUT or SETUP) has been received and loaded into the Endpoint0 FIFO. The OPRDY bit (E0CSR.0) is set to '1' by hardware.
- 2. An IN data packet has successfully been unloaded from the Endpoint0 FIFO and transmitted to the host; INPRDY is reset to '0' by hardware.
- 3. An IN transaction is completed (this interrupt generated during the status stage of the transaction).
- 4. Hardware sets the STSTL bit (E0CSR.2) after a control transaction ended due to a protocol violation.
- 5. Hardware sets the SUEND bit (E0CSR.4) because a control transfer ended before firmware sets the DATAEND bit (E0CSR.3).



The E0CNT register (USB Register Definition 16.18) holds the number of received data bytes in the Endpoint0 FIFO.

Hardware will automatically detect protocol errors and send a STALL condition in response. Firmware may force a STALL condition to abort the current transfer. When a STALL condition is generated, the STSTL bit will be set to '1' and an interrupt generated. The following conditions will cause hardware to generate a STALL condition:

- 1. The host sends an OUT token during a OUT data phase after the DATAEND bit has been set to '1'.
- 2. The host sends an IN token during an IN data phase after the DATAEND bit has been set to '1'.
- 3. The host sends a packet that exceeds the maximum packet size for Endpoint0.
- 4. The host sends a non-zero length DATA1 packet during the status phase of an IN transaction.
- 5. Firmware sets the SDSTL bit (E0CSR.5) to '1'.

#### 16.10.1.Endpoint0 SETUP Transactions

All control transfers must begin with a SETUP packet. SETUP packets are similar to OUT packets, containing an 8-byte data field sent by the host. Any SETUP packet containing a command field of anything other than 8 bytes will be automatically rejected by USB0. An Endpoint0 interrupt is generated when the data from a SETUP packet is loaded into the Endpoint0 FIFO. Software should unload the command from the Endpoint0 FIFO, decode the command, perform any necessary tasks, and set the SOPRDY bit to indicate that it has serviced the OUT packet.

#### 16.10.2.Endpoint0 IN Transactions

When a SETUP request is received that requires USB0 to transmit data to the host, one or more IN requests will be sent by the host. For the first IN transaction, firmware should load an IN packet into the Endpoint0 FIFO, and set the INPRDY bit (E0CSR.1). An interrupt will be generated when an IN packet is transmitted successfully. Note that no interrupt will be generated if an IN request is received before firmware has loaded a packet into the Endpoint0 FIFO. If the requested data exceeds the maximum packet size for Endpoint0 (as reported to the host), the data should be split into multiple packets; each packet should be of the maximum packet size excluding the last (residual) packet. If the requested data is an integer multiple of the maximum packet size for Endpoint0, the last data packet should be a zero-length packet signaling the end of the transfer. Firmware should set the DATAEND bit to '1' after loading into the Endpoint0 FIFO the last data packet for a transfer.

Upon reception of the first IN token for a particular control transfer, Endpoint0 is said to be in Transmit Mode. In this mode, only IN tokens should be sent by the host to Endpoint0. The SUEND bit (E0CSR.4) is set to '1' if a SETUP or OUT token is received while Endpoint0 is in Transmit Mode.

Endpoint0 will remain in Transmit Mode until any of the following occur:

- 1. USB0 receives an Endpoint0 SETUP or OUT token.
- 2. Firmware sends a packet less than the maximum Endpoint0 packet size.
- 3. Firmware sends a zero-length packet.

Firmware should set the DATAEND bit (E0CSR.3) to '1' when performing (2) and (3) above.

The SIE will transmit a NAK in response to an IN token if there is no packet ready in the IN FIFO (INPRDY = '0').



#### 16.10.3.Endpoint0 OUT Transactions

When a SETUP request is received that requires the host to transmit data to USB0, one or more OUT requests will be sent by the host. When an OUT packet is successfully received by USB0, hardware will set the OPRDY bit (E0CSR.0) to '1' and generate an Endpoint0 interrupt. Following this interrupt, firmware should unload the OUT packet from the Endpoint0 FIFO and set the SOPRDY bit (E0CSR.6) to '1'.

If the amount of data required for the transfer exceeds the maximum packet size for Endpoint0, the data will be split into multiple packets. If the requested data is an integer multiple of the maximum packet size for Endpoint0 (as reported to the host), the host will send a zero-length data packet signaling the end of the transfer.

Upon reception of the first OUT token for a particular control transfer, Endpoint0 is said to be in Receive Mode. In this mode, only OUT tokens should be sent by the host to Endpoint0. The SUEND bit (E0CSR.4) is set to '1' if a SETUP or IN token is received while Endpoint0 is in Receive Mode.

Endpoint0 will remain in Receive mode until:

- 1. The SIE receives a SETUP or IN token.
- 2. The host sends a packet less than the maximum Endpoint0 packet size.
- 3. The host sends a zero-length packet.

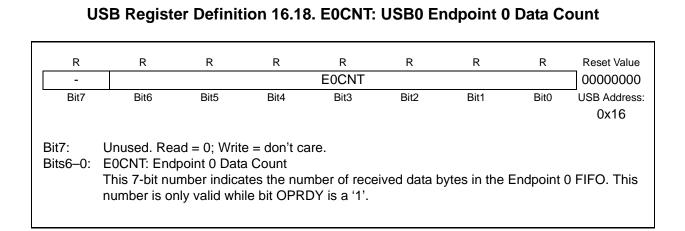
Firmware should set the DATAEND bit (E0CSR.3) to '1' when the expected amount of data has been received. The SIE will transmit a STALL condition if the host sends an OUT packet after the DATAEND bit has been set by firmware. An interrupt will be generated with the STSTL bit (E0CSR.2) set to '1' after the STALL is transmitted.



#### USB Register Definition 16.17. E0CSR: USB0 Endpoint0 Control

R/W	R/W	R/W	R	R/W	R/W	R/W	R	Reset Value			
SSUEN	O SOPRDY	SDSTL	SUEND	DATAEND	STSTL	INPRDY	OPRDY	00000000			
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address:			
								0x11			
Bit7:	SSUEND: S		•								
				to '1' after se				event.			
				hen software	writes '1'	to SSUEND					
Bit6:	Read: This & SOPRDY: S										
DILO.				this bit after	servicina a	received F	ndpoint0 p	acket The			
				ite of '1' to S							
	Read: This b										
Bit5:	SDSTL: Sen	d Stall									
				terminate the		•					
			luest, etc.).	Hardware w	ill clear thi	s bit to '0' w	hen the ST	ALL hand-			
Bit4:	shake is tran SUEND: Set										
DIL4.			l-only hit to	'1' when a c	ontrol tran	saction end	s hefore so	ftware has			
						trol transaction ends before software has nis bit when software writes '1' to SSU-					
	END.										
Bit3:	DATAEND: [	Data End									
	Software sho					_					
				the last outg							
				a zero-lengtl ter servicing			nacket				
	This bit is au	•		•		Johning uata	packet.				
Bit2:	STSTL: Sen		cical ca by	na ana oi							
	Hardware se	ets this bit to	o '1' after tr	ansmitting a	STALL ha	ndshake sig	nal. This fla	ag must be			
	cleared by s										
Bit1:	INPRDY: IN			6 I I			<b>–</b> 1 1 1				
				after loading							
	transmit. Hardware clears this bit and generates an interrupt under either of the following conditions:										
	1. The packe	et is transm	itted.								
				incoming SE	TUP pack	et.					
	3. The packe	et is overwri	itten by an	incoming OL	JT packet.						
Bit0:	OPRDY: OU										
				nd generates				has been			
	received. Th	IS DIT IS Clea	ared only w	hen software	e writes '1'	to the SOPI	KUY DIt.				





# 16.11. Configuring Endpoints1-3

Endpoints1-3 are configured and controlled through their own sets of the following control/status registers: IN registers EINCSRL and EINCSRH, and OUT registers EOUTCSRL and EOUTCSRH. Only one set of endpoint control/status registers is mapped into the USB register address space at a time, defined by the contents of the INDEX register (USB Register Definition 16.4).

Endpoints1-3 can be configured as IN, OUT, or both IN/OUT (Split Mode) as described in **Section 16.5.1**. The endpoint mode (Split/Normal) is selected via the SPLIT bit in register EINCSRH.

When SPLIT = '1', the corresponding endpoint FIFO is split, and both IN and OUT pipes are available.

When SPLIT = '0', the corresponding endpoint functions as either IN or OUT; the endpoint direction is selected by the DIRSEL bit in register EINCSRH.

#### 16.12. Controlling Endpoints1-3 IN

Endpoints1-3 IN are managed via USB registers EINCSRL and EINCSRH. All IN endpoints can be used for Interrupt, Bulk, or Isochronous transfers. Isochronous (ISO) mode is enabled by writing '1' to the ISO bit in register EINCSRH. Bulk and Interrupt transfers are handled identically by hardware.

An Endpoint1-3 IN interrupt is generated by any of the following conditions:

- 1. An IN packet is successfully transferred to the host.
- 2. Software writes '1' to the FLUSH bit (EINCSRL.3) when the target FIFO is not empty.
- 3. Hardware generates a STALL condition.

#### 16.12.1.Endpoints1-3 IN Interrupt or Bulk Mode

When the ISO bit (EINCSRH.6) = '0' the target endpoint operates in Bulk or Interrupt Mode. Once an endpoint has been configured to operate in Bulk/Interrupt IN mode (typically following an Endpoint0 SET\_IN-TERFACE command), firmware should load an IN packet into the endpoint IN FIFO and set the INPRDY bit (EINCSRL.0). Upon reception of an IN token, hardware will transmit the data, clear the INPRDY bit, and generate an interrupt.



Writing '1' to INPRDY without writing any data to the endpoint FIFO will cause a zero-length packet to be transmitted upon reception of the next IN token.

A Bulk or Interrupt pipe can be shut down (or Halted) by writing '1' to the SDSTL bit (EINCSRL.4). While SDSTL = '1', hardware will respond to all IN requests with a STALL condition. Each time hardware generates a STALL condition, an interrupt will be generated and the STSTL bit (EINCSRL.5) set to '1'. The STSTL bit must be reset to '0' by firmware.

Hardware will automatically reset INPRDY to '0' when a packet slot is open in the endpoint FIFO. Note that if double buffering is enabled for the target endpoint, it is possible for firmware to load two packets into the IN FIFO at a time. In this case, hardware will reset INPRDY to '0' immediately after firmware loads the first packet into the FIFO and sets INPRDY to '1'. An interrupt will not be generated in this case; an interrupt will only be generated when a data packet is transmitted.

When firmware writes '1' to the FCDT bit (EINCSRH.3), the data toggle for each IN packet will be toggled continuously, regardless of the handshake received from the host. This feature is typically used by Interrupt endpoints functioning as rate feedback communication for Isochronous endpoints. When FCDT = '0', the data toggle bit will only be toggled when an ACK is sent from the host in response to an IN packet.

#### 16.12.2.Endpoints1-3 IN Isochronous Mode

When the ISO bit (EINCSRH.6) is set to '1', the target endpoint operates in Isochronous (ISO) mode. Once an endpoint has been configured for ISO IN mode, the host will send one IN token (data request) per frame; the location of data within each frame may vary. Because of this, it is recommended that double buffering be enabled for ISO IN endpoints.

Hardware will automatically reset INPRDY (EINCSRL.0) to '0' when a packet slot is open in the endpoint FIFO. Note that if double buffering is enabled for the target endpoint, it is possible for firmware to load two packets into the IN FIFO at a time. In this case, hardware will reset INPRDY to '0' immediately after firmware loads the first packet into the FIFO and sets INPRDY to '1'. An interrupt will not be generated in this case; an interrupt will only be generated when a data packet is transmitted.

If there is not a data packet ready in the endpoint FIFO when USB0 receives an IN token from the host, USB0 will transmit a zero-length data packet and set the UNDRUN bit (EINCSRL.2) to '1'.

The ISO Update feature (see Section 16.7) can be useful in starting a double buffered ISO IN endpoint. If the host has already set up the ISO IN pipe (has begun transmitting IN tokens) when firmware writes the first data packet to the endpoint FIFO, the next IN token may arrive and the first data packet sent before firmware has written the second (double buffered) data packet to the FIFO. The ISO Update feature ensures that any data packet written to the endpoint FIFO will not be transmitted during the current frame; the packet will only be sent after a SOF signal has been received.



# USB Register Definition 16.19. EINCSRL: USB0 IN Endpoint Control Low Byte

R	W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value	
-	CLRDT	STSTL	SDSTL	FLUSH	UNDRUN	FIFONE	INPRDY	00000000	
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address: 0x11	
Bit7:	Unused. Rea	ad = 0; Writ	e = don't ca	are.					
Bit6:	CLRDT: Clea	ar Data Tog	gle.						
	Write: Softw			this bit to re	eset the IN E	indpoint da	ta toggle to	ʻ0'.	
	Read: This b		eads '0'.						
Bit5:	STSTL: Sen		. (1)						
	Hardware se							FIFO IS	
Bit4:	flushed, and SDSTL: Sen		r bit cleare	a. This hag	i must be cie	eared by so	ntware.		
DII4.	Software sho		' to this bit	to generate	a STALL h	andshake ii	n response	to an IN	
	token. Softw								
	effect in ISO								
Bit3:	FLUSH: FIF	O Flush.							
	Writing a '1'								
	The FIFO po								
	ets, software			SH for each	packet. Har	dware rese	ets the FLU	SH bit to '0'	
D:+0.	when the FI		•						
Bit2:	UNDRUN: D The function			the IN End	noint mode:				
	Isochronous		•		•		is received	while hit	
	INPRDY = '(		a zere leng				10 10001/00	while bit	
	Interrupt/Bul	-	s not used i	n these mo	des and will	always rea	ad a '0'.		
	This bit mus					,			
Bit1:	FIFONE: FIF								
	0: The IN Er	•							
Dire	1. The IN Er			one or more	e packets.				
Bit0:	INPRDY: In			- <b>f</b> t			NI En de si		
		Software should write '1' to this bit after loading a data packet into the IN Endpoint FIFO.							
		lardware clears INPRDY due to any of the following: . A data packet is transmitted.							
		2. Double buffering is enabled (DBIEN = '1') and there is an open FIFO packet slot.							
		B. If the endpoint is in Isochronous Mode (ISO = '1') and ISOUD = '1', INPRDY will read '0'							
	until the nex	ntil the next SOF is received.							
				generated v	when hardw	are clears	INPRDY a	s a result	
	of a packet	n interrupt (if enabled) will be generated when hardware clears INPRDY as a result f a packet being transmitted.							



### USB Register Definition 16.20. EINCSRH: USB0 IN Endpoint Control High Byte

5.44	5444	5.44	-	<b>5</b>	5.44	-	_	<b>D</b>
R/W	R/W	R/W	R	R/W	R/W	R	R	Reset Value
DBIEN	ISO	DIRSEL	-	FCDT	SPLIT	-	-	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address:
								0x12
Bit7:	DBIEN: IN E	indpoint Do	uble-buffer	Enable.				
	0: Double-bu	•						
	1: Double-bu	uffering ena	bled for the	selected IN	l endpoint.			
Bit6:	ISO: Isochro	nous Trans	fer Enable.					
	This bit enab					rent endpoi	int.	
	0: Endpoint of	•		•				
	1: Endpoint of	•			rs.			
Bit5:	DIRSEL: En	•						
	This bit is va				not split (S	SPLIT = '0').		
	0: Endpoint of							
	1: Endpoint of							
Bit4:	Unused. Rea			are.				
Bit3:	FCDT: Force							
	0: Endpoint of		switches or	nly when an	ACK is rec	eived follov	ving a data	a packet
	transmission							
	1: Endpoint of		forced to sv	witch after e	very data p	acket is trar	nsmitted, r	egardless of
		ACK reception.						
Bit2:		SPLIT: FIFO Split Enable.						
	When SPLIT = '1', the selected endpoint FIFO is split. The upper half of the selected FIFO is							
	used by the IN endpoint; the lower half of the selected FIFO is used by the OUT endpoint.						endpoint.	
Bits1–0:	ts1–0: Unused. Read = 00b; Write = don't care.							

# 16.13. Controlling Endpoints1-3 OUT

Endpoints1-3 OUT are managed via USB registers EOUTCSRL and EOUTCSRH. All OUT endpoints can be used for Interrupt, Bulk, or Isochronous transfers. Isochronous (ISO) mode is enabled by writing '1' to the ISO bit in register EOUTCSRH. Bulk and Interrupt transfers are handled identically by hardware.

An Endpoint1-3 OUT interrupt may be generated by the following:

- 1. Hardware sets the OPRDY bit (EINCSRL.0) to '1'.
- 2. Hardware generates a STALL condition.

#### 16.13.1.Endpoints1-3 OUT Interrupt or Bulk Mode

When the ISO bit (EOUTCSRH.6) = '0' the target endpoint operates in Bulk or Interrupt mode. Once an endpoint has been configured to operate in Bulk/Interrupt OUT mode (typically following an Endpoint0 SET\_INTERFACE command), hardware will set the OPRDY bit (EOUTCSRL.0) to '1' and generate an interrupt upon reception of an OUT token and data packet. The number of bytes in the current OUT data packet (the packet ready to be unloaded from the FIFO) is given in the EOUTCNTH and EOUTCNTL registers. In response to this interrupt, firmware should unload the data packet from the OUT FIFO and reset the OPRDY bit to '0'.



# C8051F340/1/2/3/4/5/6/7/8/9/A/B/C/D

A Bulk or Interrupt pipe can be shut down (or Halted) by writing '1' to the SDSTL bit (EOUTCSRL.5). While SDSTL = '1', hardware will respond to all OUT requests with a STALL condition. Each time hardware generates a STALL condition, an interrupt will be generated and the STSTL bit (EOUTCSRL.6) set to '1'. The STSTL bit must be reset to '0' by firmware.

Hardware will automatically set OPRDY when a packet is ready in the OUT FIFO. Note that if double buffering is enabled for the target endpoint, it is possible for two packets to be ready in the OUT FIFO at a time. In this case, hardware will set OPRDY to '1' immediately after firmware unloads the first packet and resets OPRDY to '0'. A second interrupt will be generated in this case.

#### 16.13.2.Endpoints1-3 OUT Isochronous Mode

When the ISO bit (EOUTCSRH.6) is set to '1', the target endpoint operates in Isochronous (ISO) mode. Once an endpoint has been configured for ISO OUT mode, the host will send exactly one data per USB frame; the location of the data packet within each frame may vary, however. Because of this, it is recommended that double buffering be enabled for ISO OUT endpoints.

Each time a data packet is received, hardware will load the received data packet into the endpoint FIFO, set the OPRDY bit (EOUTCSRL.0) to '1', and generate an interrupt (if enabled). Firmware would typically use this interrupt to unload the data packet from the endpoint FIFO and reset the OPRDY bit to '0'.

If a data packet is received when there is no room in the endpoint FIFO, an interrupt will be generated and the OVRUN bit (EOUTCSRL.2) set to '1'. If USB0 receives an ISO data packet with a CRC error, the data packet will be loaded into the endpoint FIFO, OPRDY will be set to '1', an interrupt (if enabled) will be generated, and the DATAERR bit (EOUTCSRL.3) will be set to '1'. Software should check the DATAERR bit each time a data packet is unloaded from an ISO OUT endpoint FIFO.



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# USB Register Definition 16.21. EOUTCSRL: USB0 OUT Endpoint Control Low Byte

W	R/W	R/W	R/W	R	R/W	R	R/W	Reset Value	
CLRDT	STSTL	SDSTL	FLUSH	DATERR	OVRUN	FIFOFUL	OPRDY	00000000	
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address: 0x14	
Bit7:	CLRDT: Clear Data Toggle Write: Software should write '1' to this bit to reset the OUT endpoint data toggle to '0'. Read: This bit always reads '0'.								
Bit6:	STSTL: Sen Hardware se be cleared b	ets this bit to	oʻ1' when a	a STALL har	ndshake sig	inal is trans	mitted. This	s flag must	
Bit5:		ould write '1 to terminate						nould write	
Bit4:	<ul> <li>'0' to this bit to terminate the STALL signal. This bit has no effect in ISO mode.</li> <li>FLUSH: FIFO Flush</li> <li>Writing a '1' to this bit flushes the next packet to be read from the OUT endpoint FIFO. The</li> <li>FIFO pointer is reset and the OPRDY bit is cleared. If the FIFO contains multiple packets, software must write '1' to FLUSH for each packet. Hardware resets the FLUSH bit to '0' when the FIFO flush is complete.</li> <li>Note: If data for the current packet has already been read from the FIFO, the FLUSH bit should not be used to flush the packet. Instead, the entire data packet should be read from the</li> </ul>								
Bit3:	DATERR: Da In ISO mode It is cleared	e, this bit is a when softwa			•			uffing error.	
Bit2:	This bit is se endpoint FIF 0: No data o	OVRUN: Data Overrun This bit is set by hardware when an incoming data packet cannot be loaded into the OUT endpoint FIFO. This bit is only valid in ISO mode, and must be cleared by software. 0: No data overrun.							
Bit1:	<ul> <li>1: A data packet was lost because of a full FIFO since this flag was last cleared.</li> <li>FIFOFUL: OUT FIFO Full</li> <li>This bit indicates the contents of the OUT FIFO. If double buffering is enabled for the endpoint (DBIEN = '1'), the FIFO is full when the FIFO contains two packets. If DBIEN = '0', the FIFO is full when the FIFO contains one packet.</li> <li>0: OUT endpoint FIFO is not full.</li> <li>1: OUT endpoint FIFO is full.</li> </ul>								
Bit0:	OPRDY: OU Hardware se ware should	ets this bit to	o '1' and ge		•				

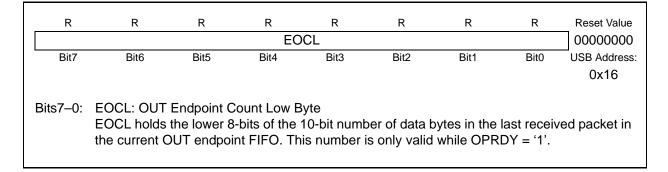


# C8051F340/1/2/3/4/5/6/7/8/9/A/B/C/D

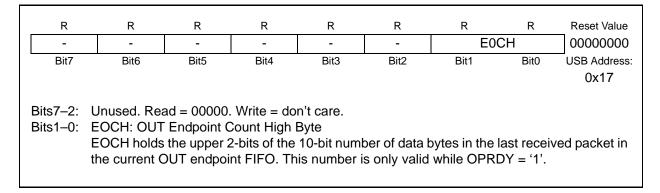
# USB Register Definition 16.22. EOUTCSRH: USB0 OUT Endpoint Control High Byte

R/W DBOEN	R/W	R/W	R/W	R -	R -	R -	R -	Reset Value		
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	USB Address:		
								0x15		
Bit7:	Bit7: DBOEN: Double-buffer Enable 0: Double-buffering disabled for the selected OUT endpoint.									
DitC	1: Double-bu	-		selected O	UT endpoir	nt.				
Bit6:	ISO: Isochro This bit enab			ous transfers	s on the cur	rent endooi	nt			
	This bit enables/disables isochronous transfers on the current endpoint. 0: Endpoint configured for bulk/interrupt transfers.									
	1: Endpoint configured for isochronous transfers.									
Bits5–0:	: Unused. Read = 000000b; Write = don't care.									

# USB Register Definition 16.23. EOUTCNTL: USB0 OUT Endpoint Count Low



# USB Register Definition 16.24. EOUTCNTH: USB0 OUT Endpoint Count High





# Table 16.4. USB Transceiver Electrical Characteristics

#### $V_{DD}$ = 3.0 to 3.6 V, -40 to +85 °C unless otherwise specified

Parameters	Symbol	Conditions	Min	Тур	Max	Units
Transmitter						
Output High Voltage	V <sub>OH</sub>		2.8			V
Output Low Voltage	V <sub>OL</sub>				0.8	V
Output Crossover Point	V <sub>CRS</sub>		1.3		2.0	V
	7	Driving High		38		0
Output Impedance	Z <sub>DRV</sub>	Driving Low		38		Ω
Dull un Desistense	<b>D</b>	Full Speed (D+ Pull-up)	1 405	1 5	1 575	1-0
Pull-up Resistance	R <sub>PU</sub>	Low Speed (D- Pull-up)	1.425	1.5	1.575	kΩ
Output Diag Time	т	Low Speed	75		300	
Output Rise Time	Τ <sub>R</sub>	Full Speed	4		20	ns
	т	Low Speed	75		300	
Output Fall Time	Τ <sub>F</sub>	Full Speed	4		20	ns
Receiver						
Differential Input	V <sub>DI</sub>		0.2			V
Sensitivity	V DI	(D+) – (D–)	0.2			v
Differential Input Common	V <sub>CM</sub>		0.8		2.5	V
Mode Range	* CM		0.0		2.5	v
Input Leakage Current	١L	Pullups Disabled		<1.0		μA

**Note:** Refer to the USB Specification for timing diagrams and symbol definitions.



# 17. SMBus

The SMBus I/O interface is a two-wire, bi-directional serial bus. The SMBus is compliant with the System Management Bus Specification, version 1.1, and compatible with the I2C serial bus. Reads and writes to the interface by the system controller are byte oriented with the SMBus interface autonomously controlling the serial transfer of the data. Data can be transferred at up to 1/20th of the system clock as a master or slave (this can be faster than allowed by the SMBus specification, depending on the system clock used). A method of extending the clock-low duration is available to accommodate devices with different speed capabilities on the same bus.

The SMBus interface may operate as a master and/or slave, and may function on a bus with multiple masters. The SMBus provides control of SDA (serial data), SCL (serial clock) generation and synchronization, arbitration logic, and START/STOP control and generation. Three SFRs are associated with the SMBus: SMB0CF configures the SMBus; SMB0CN controls the status of the SMBus; and SMB0DAT is the data register, used for both transmitting and receiving SMBus data and slave addresses.

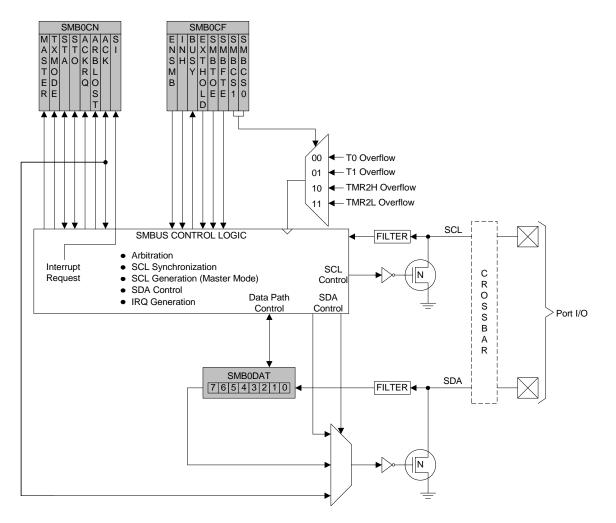


Figure 17.1. SMBus Block Diagram



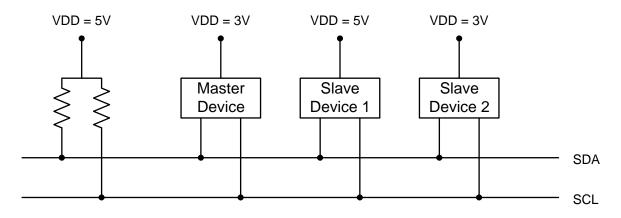
# 17.1. Supporting Documents

It is assumed the reader is familiar with or has access to the following supporting documents:

- 1. The I2C-Bus and How to Use It (including specifications), Philips Semiconductor.
- 2. The I2C-Bus Specification -- Version 2.0, Philips Semiconductor.
- 3. System Management Bus Specification -- Version 1.1, SBS Implementers Forum.

# 17.2. SMBus Configuration

Figure 17.2 shows a typical SMBus configuration. The SMBus specification allows any recessive voltage between 3.0 V and 5.0 V; different devices on the bus may operate at different voltage levels. The bi-directional SCL (serial clock) and SDA (serial data) lines must be connected to a positive power supply voltage through a pull-up resistor or similar circuit. Every device connected to the bus must have an open-drain or open-collector output for both the SCL and SDA lines, so that both are pulled high (recessive state) when the bus is free. The maximum number of devices on the bus is limited only by the requirement that the rise and fall times on the bus not exceed 300 ns and 1000 ns, respectively.





# 17.3. SMBus Operation

Two types of data transfers are possible: data transfers from a master transmitter to an addressed slave receiver (WRITE), and data transfers from an addressed slave transmitter to a master receiver (READ). The master device initiates both types of data transfers and provides the serial clock pulses on SCL. The SMBus interface may operate as a master or a slave, and multiple master devices on the same bus are supported. If two or more masters attempt to initiate a data transfer simultaneously, an arbitration scheme is employed with a single master always winning the arbitration. Note that it is not necessary to specify one device as the Master in a system; any device who transmits a START and a slave address becomes the master for the duration of that transfer.

A typical SMBus transaction consists of a START condition followed by an address byte (Bits7-1: 7-bit slave address; Bit0: R/W direction bit), one or more bytes of data, and a STOP condition. Each byte that is received (by a master or slave) must be acknowledged (ACK) with a low SDA during a high SCL (see Figure 17.3). If the receiving device does not ACK, the transmitting device will read a NACK (not acknowledge), which is a high SDA during a high SCL.



The direction bit (R/W) occupies the least-significant bit position of the address byte. The direction bit is set to logic 1 to indicate a "READ" operation and cleared to logic 0 to indicate a "WRITE" operation.

All transactions are initiated by a master, with one or more addressed slave devices as the target. The master generates the START condition and then transmits the slave address and direction bit. If the transaction is a WRITE operation from the master to the slave, the master transmits the data a byte at a time waiting for an ACK from the slave at the end of each byte. For READ operations, the slave transmits the data waiting for an ACK from the master at the end of each byte. At the end of the data transfer, the master generates a STOP condition to terminate the transaction and free the bus. Figure 17.3 illustrates a typical SMBus transaction.

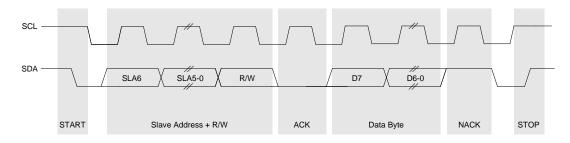


Figure 17.3. SMBus Transaction

#### 17.3.1. Arbitration

A master may start a transfer only if the bus is free. The bus is free after a STOP condition or after the SCL and SDA lines remain high for a specified time (see Section "17.3.4. SCL High (SMBus Free) Timeout" on page 191). In the event that two or more devices attempt to begin a transfer at the same time, an arbitration scheme is employed to force one master to give up the bus. The master devices continue transmitting until one attempts a HIGH while the other transmits a LOW. Since the bus is open-drain, the bus will be pulled LOW. The master attempting the HIGH will detect a LOW SDA and lose the arbitration. The winning master continues its transmission without interruption; the losing master becomes a slave and receives the rest of the transfer if addressed. This arbitration scheme is non-destructive: one device always wins, and no data is lost.



#### 17.3.2. Clock Low Extension

SMBus provides a clock synchronization mechanism, similar to I2C, which allows devices with different speed capabilities to coexist on the bus. A clock-low extension is used during a transfer in order to allow slower slave devices to communicate with faster masters. The slave may temporarily hold the SCL line LOW to extend the clock low period, effectively decreasing the serial clock frequency.

#### 17.3.3. SCL Low Timeout

If the SCL line is held low by a slave device on the bus, no further communication is possible. Furthermore, the master cannot force the SCL line high to correct the error condition. To solve this problem, the SMBus protocol specifies that devices participating in a transfer must detect any clock cycle held low longer than 25 ms as a "timeout" condition. Devices that have detected the timeout condition must reset the communication no later than 10 ms after detecting the timeout condition.

When the SMBTOE bit in SMB0CF is set, Timer 3 is used to detect SCL low timeouts. Timer 3 is forced to reload when SCL is high, and allowed to count when SCL is low. With Timer 3 enabled and configured to overflow after 25 ms (and SMBTOE set), the Timer 3 interrupt service routine can be used to reset (disable and re-enable) the SMBus in the event of an SCL low timeout.

#### 17.3.4. SCL High (SMBus Free) Timeout

The SMBus specification stipulates that if the SCL and SDA lines remain high for more that 50 µs, the bus is designated as free. When the SMBFTE bit in SMB0CF is set, the bus will be considered free if SCL and SDA remain high for more than 10 SMBus clock source periods. If the SMBus is waiting to generate a Master START, the START will be generated following this timeout. Note that a clock source is required for free timeout detection, even in a slave-only implementation.

#### 17.4. Using the SMBus

The SMBus can operate in both Master and Slave modes. The interface provides timing and shifting control for serial transfers; higher level protocol is determined by user software. The SMBus interface provides the following application-independent features:

- Byte-wise serial data transfers
- Clock signal generation on SCL (Master Mode only) and SDA data synchronization
- Timeout/bus error recognition, as defined by the SMB0CF configuration register
- START/STOP timing, detection, and generation
- Bus arbitration
- Interrupt generation
- Status information

SMBus interrupts are generated for each data byte or slave address that is transferred. When transmitting, this interrupt is generated after the ACK cycle so that software may read the received ACK value; when receiving data, this interrupt is generated before the ACK cycle so that software may define the outgoing ACK value. See **Section "17.5. SMBus Transfer Modes" on page 198** for more details on transmission sequences.

Interrupts are also generated to indicate the beginning of a transfer when a master (START generated), or the end of a transfer when a slave (STOP detected). Software should read the SMB0CN (SMBus Control register) to find the cause of the SMBus interrupt. The SMB0CN register is described in Section "17.4.2. SMB0CN Control Register" on page 195; Table 17.4 provides a quick SMB0CN decoding reference.



SMBus configuration options include:

- Timeout detection (SCL Low Timeout and/or Bus Free Timeout)
- SDA setup and hold time extensions
- Slave event enable/disable
- Clock source selection

These options are selected in the SMB0CF register, as described in **Section "17.4.1. SMBus Configura**tion Register" on page 192.

#### 17.4.1. SMBus Configuration Register

The SMBus Configuration register (SMB0CF) is used to enable the SMBus Master and/or Slave modes, select the SMBus clock source, and select the SMBus timing and timeout options. When the ENSMB bit is set, the SMBus is enabled for all master and slave events. Slave events may be disabled by setting the INH bit. With slave events inhibited, the SMBus interface will still monitor the SCL and SDA pins; however, the interface will NACK all received addresses and will not generate any slave interrupts. When the INH bit is set, all slave events will be inhibited following the next START (interrupts will continue for the duration of the current transfer).

SMBCS1	SMBCS0	SMBus Clock Source
0	0	Timer 0 Overflow
0	1	Timer 1 Overflow
1	0	Timer 2 High Byte Overflow
1	1	Timer 2 Low Byte Overflow

#### Table 17.1. SMBus Clock Source Selection

The SMBCS1-0 bits select the SMBus clock source, which is used only when operating as a master or when the Free Timeout detection is enabled. When operating as a master, overflows from the selected source determine the absolute minimum SCL low and high times as defined in Equation 17.1. Note that the selected clock source may be shared by other peripherals so long as the timer is left running at all times. For example, Timer 1 overflows may generate the SMBus and UART baud rates simultaneously. Timer configuration is covered in Section "21. Timers" on page 235.

$$T_{HighMin} = T_{LowMin} = \frac{1}{f_{ClockSourceOverflow}}$$

#### Equation 17.1. Minimum SCL High and Low Times

The selected clock source should be configured to establish the minimum SCL High and Low times as per Equation 17.1. When the interface is operating as a master (and SCL is not driven or extended by any other devices on the bus), the typical SMBus bit rate is approximated by Equation 17.2.

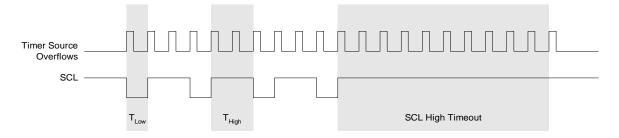
$$BitRate = \frac{f_{ClockSourceOverflow}}{3}$$

#### Equation 17.2. Typical SMBus Bit Rate



# C8051F340/1/2/3/4/5/6/7/8/9/A/B/C/D

Figure 17.4 shows the typical SCL generation described by Equation 17.2. Notice that  $T_{HIGH}$  is typically twice as large as  $T_{LOW}$ . The actual SCL output may vary due to other devices on the bus (SCL may be extended low by slower slave devices, or driven low by contending master devices). The bit rate when operating as a master will never exceed the limits defined by equation Equation 17.1.





Setting the EXTHOLD bit extends the minimum setup and hold times for the SDA line. The minimum SDA setup time defines the absolute minimum time that SDA is stable before SCL transitions from low-to-high. The minimum SDA hold time defines the absolute minimum time that the current SDA value remains stable after SCL transitions from high-to-low. EXTHOLD should be set so that the minimum setup and hold times meet the SMBus Specification requirements of 250 ns and 300 ns, respectively. Table 17.2 shows the minimum setup and hold times for the two EXTHOLD settings. Setup and hold time extensions are typically necessary when SYSCLK is above 10 MHz.

EXTHOLD	Minimum SDA Setup Time	Minimum SDA Hold Time
	T <sub>low</sub> - 4 system clocks	
0	OR	3 system clocks
	1 system clock + s/w delay*	
1	11 system clocks	12 system clocks

Table 17.2. Minimum SDA Setup and Hold Times

\*Note: Setup Time for ACK bit transmissions and the MSB of all data transfers. The s/w delay occurs between the time SMB0DAT or ACK is written and when SI is cleared. Note that if SI is cleared in the same write that defines the outgoing ACK value, s/w delay is zero.

With the SMBTOE bit set, Timer 3 should be configured to overflow after 25 ms in order to detect SCL low timeouts (see Section "17.3.3. SCL Low Timeout" on page 191). The SMBus interface will force Timer 3 to reload while SCL is high, and allow Timer 3 to count when SCL is low. The Timer 3 interrupt service routine should be used to reset SMBus communication by disabling and re-enabling the SMBus.

SMBus Free Timeout detection can be enabled by setting the SMBFTE bit. When this bit is set, the bus will be considered free if SDA and SCL remain high for more than 10 SMBus clock source periods (see Figure 17.4). When a Free Timeout is detected, the interface will respond as if a STOP was detected (an interrupt will be generated, and STO will be set).



R/W	R/W	R	R/W	R/W	R/W	R/W	R/W	Reset Value
ENSMB	INH	BUSY	EXTHOLD	SMBTOE	SMBFTE	SMBCS1	SMBCS0	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	-
							SFR Address	: 0xC1
Bit7:	ENSMB: SM						_	
	This bit enab			is interface.	When enal	bled, the int	erface cons	stantly mon-
	itors the SD 0: SMBus in		•					
	1: SMBus in							
Bit6:	INH: SMBus							
Bito.	When this bi			MBus does	not genera	ate an interr	upt when s	ave events
	occur. This e							
	not affected.	•						·
	0: SMBus S	lave Mode e	enabled.					
	1: SMBus S							
Bit5:	BUSY: SMB							
	This bit is se when a STC				anster is in	progress. It	is cleared	to logic U
Bit4:	EXTHOLD:				nsion Engh	ام		
DII4.	This bit cont							
	0: SDA Exte							
	1: SDA Exte							
Bit3:	SMBTOE: S							
	This bit enab							
	reload while							
	programmed				and the Tim	er 3 interru	ot service ro	outine
Bit2:	should reset				blo			
DILZ.	When this bi					e if SCL ar	d SDA rem	ain high for
	more than 1							annightor
Bits1–0:	SMBCS1-SM			•	ection.			
	These two b	its select th	e SMBus cl	ock source	which is u	sed to gene	rate the SN	/IBus bit
	rate. The se	lected devic	ce should be	e configured	according	to Equation	ו 17.1.	
	SMBCS1	SMBCS0	SM	Bus Clock	Sourco			
		<b>ЗМВСЗО</b>		Fimer 0 Ove				
	0	1	-	Timer 1 Ove				
	1	0		2 High Byte		—		
	1	1		2 Low Byte				
	·	· ·				]		



#### 17.4.2. SMB0CN Control Register

SMB0CN is used to control the interface and to provide status information (see SFR Definition 17.2). The higher four bits of SMB0CN (MASTER, TXMODE, STA, and STO) form a status vector that can be used to jump to service routines. MASTER and TXMODE indicate the master/slave state and transmit/receive modes, respectively.

STA and STO indicate that a START and/or STOP has been detected or generated since the last SMBus interrupt. STA and STO are also used to generate START and STOP conditions when operating as a master. Writing a '1' to STA will cause the SMBus interface to enter Master Mode and generate a START when the bus becomes free (STA is not cleared by hardware after the START is generated). Writing a '1' to STO while in Master Mode will cause the interface to generate a STOP and end the current transfer after the next ACK cycle. If STO and STA are both set (while in Master Mode), a STOP followed by a START will be generated.

As a receiver, writing the ACK bit defines the outgoing ACK value; as a transmitter, reading the ACK bit indicates the value received on the last ACK cycle. ACKRQ is set each time a byte is received, indicating that an outgoing ACK value is needed. When ACKRQ is set, software should write the desired outgoing value to the ACK bit before clearing SI. A NACK will be generated if software does not write the ACK bit before clearing SI. SDA will reflect the defined ACK value immediately following a write to the ACK bit; however SCL will remain low until SI is cleared. If a received slave address is not acknowledged, further slave events will be ignored until the next START is detected.

The ARBLOST bit indicates that the interface has lost an arbitration. This may occur anytime the interface is transmitting (master or slave). A lost arbitration while operating as a slave indicates a bus error condition. ARBLOST is cleared by hardware each time SI is cleared.

The SI bit (SMBus Interrupt Flag) is set at the beginning and end of each transfer, after each byte frame, or when an arbitration is lost; see Table 17.3 for more details.

**Important Note About the SI Bit:** The SMBus interface is stalled while SI is set; thus SCL is held low, and the bus is stalled until software clears SI.

Table 17.3 lists all sources for hardware changes to the SMB0CN bits. Refer to Table 17.4 for SMBus status decoding using the SMB0CN register.



SFR Definition 17.2	. SMB0CN: SMBus	Control
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R	R	R/W	R/W	R	R	R/W	R/W	Reset Value	
MASTE	R TXMODE	STA	STO	ACKRQ	ARBLOST	ACK	SI	0000000	
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	Bit Addressable	
							SFR Addres		
Bit7:	MASTER: SM	Bus Mast	er/Slave In	dicator.					
	This read-only	y bit indica	tes when t	he SMBus i	s operating a	s a maste	r.		
	0: SMBus ope								
	1: SMBus ope								
Bit6:	TXMODE: SN						•		
	This read-only	•		he SMBus i	s operating a	s a transm	nitter.		
	0: SMBus in F								
Bit5:	1: SMBus in T								
טווט.	STA: SMBus Write:	olari Flay.							
	0: No Start ge	enerated.							
	1: When oper		master, a S	START cond	lition is transr	nitted if the	e bus is fre	e (If the bus	
	is not free, the								
	STA is set by	software a	as an active	e Master, a	repeated STA	RT will be	e generate	d after the	
	next ACK cyc	le.							
	Read:								
	0: No Start or								
	1: Start or rep								
Bit4:	STO: SMBus	Stop Flag.	i da se						
	Write:	andition in		ام					
	0: No STOP of 1: Setting ST				dition to be tre	nomittod	oftor the n		
	cycle. When t	-							
	and STO are			-			-		
	Read:					59 4 0174			
	0: No Stop co	ndition det	tected.						
	1: Stop condit			ave Mode) o	or pending (if i	in Master	Mode).		
Bit3:	ACKRQ: SME		•	,			,		
	This read-only	y bit is set	to logic 1 v	when the SN	/Bus has rec	eived a by	te and nee	eds the ACK	
	bit to be writte				e value.				
Bit2:	ARBLOST: SI							_	
	This read-only	•	•					ting as a	
D:44	transmitter. A			a slave indi	cates a bus e	error condi	tion.		
Bit1:	ACK: SMBus		0 0	lovel and r	ooordo incom		avolo It ob	ould be writ	
	This bit define					-			
	ten each time a byte is received (when ACKRQ=1), or read after each byte is transmitted. 0: A "not acknowledge" has been received (if in Transmitter Mode) OR will be transmitted (if								
	in Receiver M	-			in mansmiller	mode) O		anonniteu (i	
	1: An "acknow	,	s been rec	eived (if in <sup>-</sup>	Fransmitter M	ode) OR \	vill be tran	smitted (if in	
	Receiver Mod	-	2 2 3 6 1 1 0 0			, <b>-</b>			
Bit0:	SI: SMBus Int	,	g.						
	This bit is set	•	-	ne conditior	ns listed in Tal	ble 17.3. S	SI must be	cleared by	
	software. Whi	•						,	



Bit	Set by Hardware When:	Cleared by Hardware When:
MASTER	A START is generated.	A STOP is generated.
MASILIN		<ul> <li>Arbitration is lost.</li> </ul>
	START is generated.	A START is detected.
TXMODE	<ul> <li>SMB0DAT is written before the start of an</li> </ul>	<ul> <li>Arbitration is lost.</li> </ul>
IXWODE	SMBus frame.	<ul> <li>SMB0DAT is not written before the</li> </ul>
		start of an SMBus frame.
STA	<ul> <li>A START followed by an address byte is received.</li> </ul>	<ul> <li>Must be cleared by software.</li> </ul>
	• A STOP is detected while addressed as a	• A pending STOP is generated.
STO	slave.	
	<ul> <li>Arbitration is lost due to a detected STOP.</li> </ul>	
ACKRQ	<ul> <li>A byte has been received and an ACK</li> </ul>	After each ACK cycle.
ACINIQ	response value is needed.	
	<ul> <li>A repeated START is detected as a MASTER</li> </ul>	<ul> <li>Each time SI is cleared.</li> </ul>
	when STA is low (unwanted repeated START).	
ARBLOST	• SCL is sensed low while attempting to gener-	
	ate a STOP or repeated START condition.	
	• SDA is sensed low while transmitting a '1'	
	(excluding ACK bits).	
ACK	<ul> <li>The incoming ACK value is low (ACKNOWL- EDGE).</li> </ul>	• The incoming ACK value is high (NOT ACKNOWLEDGE).
	<ul> <li>A START has been generated.</li> </ul>	<ul> <li>Must be cleared by software.</li> </ul>
	<ul> <li>Lost arbitration.</li> </ul>	
	<ul> <li>A byte has been transmitted and an ACK/</li> </ul>	
SI	NACK received.	
	• A byte has been received.	
	• A START or repeated START followed by a	
	slave address + R/W has been received.	
	<ul> <li>A STOP has been received.</li> </ul>	

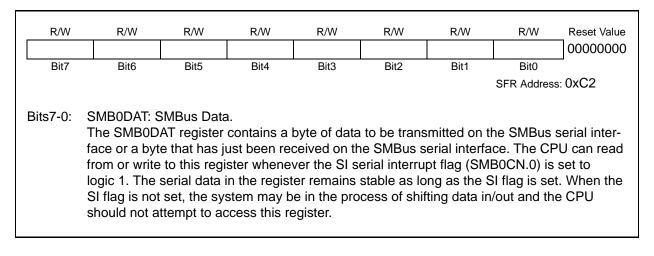
Table 17.3. Sources for Hardware Changes to SMB0CN



#### 17.4.3. Data Register

The SMBus Data register SMB0DAT holds a byte of serial data to be transmitted or one that has just been received. Software may safely read or write to the data register when the SI flag is set. Software should not attempt to access the SMB0DAT register when the SMBus is enabled and the SI flag is cleared to logic 0, as the interface may be in the process of shifting a byte of data into or out of the register.

Data in SMB0DAT is always shifted out MSB first. After a byte has been received, the first bit of received data is located at the MSB of SMB0DAT. While data is being shifted out, data on the bus is simultaneously being shifted in. SMB0DAT always contains the last data byte present on the bus. In the event of lost arbitration, the transition from master transmitter to slave receiver is made with the correct data or address in SMB0DAT.



# SFR Definition 17.3. SMB0DAT: SMBus Data

# 17.5. SMBus Transfer Modes

The SMBus interface may be configured to operate as master and/or slave. At any particular time, it will be operating in one of the following four modes: Master Transmitter, Master Receiver, Slave Transmitter, or Slave Receiver. The SMBus interface enters Master Mode any time a START is generated, and remains in Master Mode until it loses an arbitration or generates a STOP. An SMBus interrupt is generated at the end of all SMBus byte frames; however, note that the interrupt is generated before the ACK cycle when operating as a receiver, and after the ACK cycle when operating as a transmitter.

#### 17.5.1. Master Transmitter Mode

Serial data is transmitted on SDA while the serial clock is output on SCL. The SMBus interface generates the START condition and transmits the first byte containing the address of the target slave and the data direction bit. In this case the data direction bit (R/W) will be logic 0 (WRITE). The master then transmits one or more bytes of serial data. After each byte is transmitted, an acknowledge bit is generated by the slave. The transfer is ended when the STO bit is set and a STOP is generated. Note that the interface will switch to Master Receiver Mode if SMB0DAT is not written following a Master Transmitter interrupt. Figure 17.5 shows a typical Master Transmitter sequence. Two transmit data bytes are shown, though any number of bytes may be transmitted. Notice that the 'data byte transferred' interrupts occur **after** the ACK cycle in this mode.



# C8051F340/1/2/3/4/5/6/7/8/9/A/B/C/D

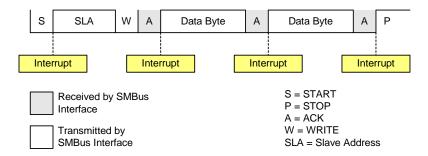


Figure 17.5. Typical Master Transmitter Sequence



#### 17.5.2. Master Receiver Mode

Serial data is received on SDA while the serial clock is output on SCL. The SMBus interface generates the START condition and transmits the first byte containing the address of the target slave and the data direction bit. In this case the data direction bit (R/W) will be logic 1 (READ). Serial data is then received from the slave on SDA while the SMBus outputs the serial clock. The slave transmits one or more bytes of serial data. After each byte is received, ACKRQ is set to '1' and an interrupt is generated. Software must write the ACK bit (SMB0CN.1) to define the outgoing acknowledge value (Note: writing a '1' to the ACK bit generates an ACK; writing a '0' generates a NACK). Software should write a '0' to the ACK bit after the last byte is received, to transmit a NACK. The interface exits Master Receiver Mode after the STO bit is set and a STOP is generated. Note that the interface will switch to Master Transmitter Mode if SMB0DAT is written while an active Master Receiver. Figure 17.6 shows a typical Master Receiver sequence. Two received data bytes are shown, though any number of bytes may be received. Notice that the 'data byte transferred' interrupts occur **before** the ACK cycle in this mode.

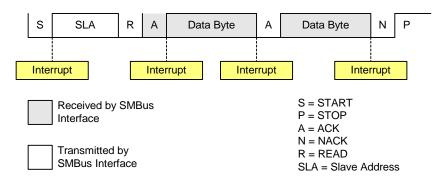


Figure 17.6. Typical Master Receiver Sequence



#### 17.5.3. Slave Receiver Mode

Serial data is received on SDA and the clock is received on SCL. When slave events are enabled (INH = 0), the interface enters Slave Receiver Mode when a START followed by a slave address and direction bit (WRITE in this case) is received. Upon entering Slave Receiver Mode, an interrupt is generated and the ACKRQ bit is set. Software responds to the received slave address with an ACK, or ignores the received slave address with a NACK. If the received slave address is ignored, slave interrupts will be inhibited until the next START is detected. If the received slave address is acknowledged, zero or more data bytes are received. Software must write the ACK bit after each received byte to ACK or NACK the received byte. The interface exits Slave Receiver Mode after receiving a STOP. Note that the interface will switch to Slave Transmitter Mode if SMB0DAT is written while an active Slave Receiver. Figure 17.7 shows a typical Slave Receiver sequence. Two received data bytes are shown, though any number of bytes may be received. Notice that the 'data byte transferred' interrupts occur **before** the ACK cycle in this mode.

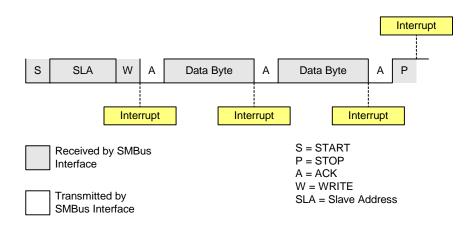
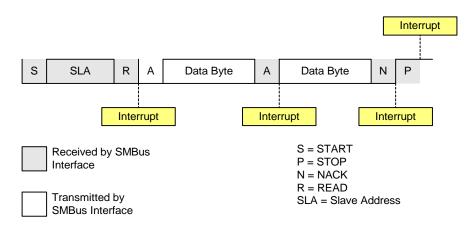


Figure 17.7. Typical Slave Receiver Sequence



#### 17.5.4. Slave Transmitter Mode

Serial data is transmitted on SDA and the clock is received on SCL. When slave events are enabled (INH = 0), the interface enters Slave Receiver Mode (to receive the slave address) when a START followed by a slave address and direction bit (READ in this case) is received. Upon entering Slave Transmitter Mode, an interrupt is generated and the ACKRQ bit is set. Software responds to the received slave address with an ACK, or ignores the received slave address with a NACK. If the received slave address is ignored, slave interrupts will be inhibited until a START is detected. If the received slave address is acknowledged, data should be written to SMB0DAT to be transmitted. The interface enters Slave Transmitter Mode, and transmits one or more bytes of data. After each byte is transmitted, the master sends an acknowledge bit; if the acknowledge bit is an ACK, SMB0DAT should be written with the next data byte. If the acknowledge bit is a NACK, SMB0DAT should not be written to before SI is cleared (Note: an error condition may be generated if SMB0DAT is written following a received NACK while in Slave Transmitter Mode). The interface exits Slave Transmitter Mode after receiving a STOP. Note that the interface will switch to Slave Receiver Mode if SMB0DAT is not written following a Slave Transmitter interrupt. Figure 17.8 shows a typical Slave Transmitter sequence. Two transmitted data bytes are shown, though any number of bytes may be transmitted. Notice that the 'data byte transferred' interrupts occur **after** the ACK cycle in this mode.



# Figure 17.8. Typical Slave Transmitter Sequence

# 17.6. SMBus Status Decoding

The current SMBus status can be easily decoded using the SMB0CN register. In the table below, STATUS VECTOR refers to the four upper bits of SMB0CN: MASTER, TXMODE, STA, and STO. Note that the shown response options are only the typical responses; application-specific procedures are allowed as long as they conform to the SMBus specification. Highlighted responses are allowed but do not conform to the SMBus specification.



# C8051F340/1/2/3/4/5/6/7/8/9/A/B/C/D

	Values F		ues Read				/alue Vritte			
Mode	Status Vector	ACKRQ	ARBLOST	ACK	Current SMbus State	Typical Response Options	STA	STo	ACK	
	1110	0	0	х	A master START was generated.	Load slave address + R/W into SMB0DAT.	0	0	х	
		0	0	0	A master data or address byte	Set STA to restart transfer.	1	0	Х	
er			Ŭ	Ŭ	was transmitted; NACK received.	Abort transfer.	0	1	Х	
nsmitt						Load next data byte into SMB0DAT.	0	0	х	
Tra						End transfer with STOP.	0	1	Х	
Master Transmitter	1100	0	0	1	10 master data or address byte	End transfer with STOP and start another transfer.	1	1	х	
2							Send repeated START.	1	0	Х
						Switch to Master Receiver Mode (clear SI without writ- ing new data to SMB0DAT).	0	0	x	
					-	Acknowledge received byte; Read SMB0DAT.	0	0	1	
						Send NACK to indicate last byte, and send STOP.	0	1	0	
						Send NACK to indicate last byte, and send STOP fol- lowed by START.	1	1	0	
ceiver						Send ACK followed by repeated START.	1	0	1	
Master Receiver	1000	1	0	х	A master data byte was received; ACK requested.	Send NACK to indicate last byte, and send repeated START.	1	0	0	
Z						Send ACK and switch to Master Transmitter Mode (write to SMB0DAT before clearing SI).	0	0	1	
						Send NACK and switch to Master Transmitter Mode (write to SMB0DAT before clearing SI).	0	0	0	

Table 17.4. SMBus Status Decoding



	Values Read			ł			Values Written				
Mode	Status Vector ACKRQ ARBLOST ACK		ACK	Current SMbus State	Typical Response Options	STA	STo	ACK			
er e		0	0	0	A slave byte was transmitted; NACK received.	No action required (expect- ing STOP condition).	0	0	х		
smitte	0100	0	0	1	A slave byte was transmitted; ACK received.	Load SMB0DAT with next data byte to transmit.	0	0	х		
Slave Transmitter		0	1	х	A Slave byte was transmitted; error detected.	No action required (expect- ing Master to end transfer).	0	0	х		
Slav	0101	0	x	x	An illegal STOP or bus error was detected while a Slave Transmission was in progress.	Clear STO.	0	0	x		
		1	0	x	A slave address was received; ac	Acknowledge received address.	0	0	1		
	0010					^	ACK requested.	Do not acknowledge received address.	0	0	0
					i	Acknowledge received address.	0	0	1		
		1	1	х	Lost arbitration as master; slave address received; ACK	Do not acknowledge received address.	0	0	0		
					requested.	Reschedule failed transfer; do not acknowledge received address.	1	0	0		
eiver	0010	0	1	х		Abort failed transfer.	0	0	Х		
kece					repeated START.	Reschedule failed transfer.	1	0	Х		
Slave Receiver		1	1	Х	Lost arbitration while attempting a STOP.	No action required (transfer complete/aborted).	0	0	0		
S	0001	0	0	х	A STOP was detected while addressed as a Slave Transmitter or Slave Receiver.	Clear STO.	0	0	х		
		0	1	х	Lost arbitration due to a detected	Abort transfer.	0	0	Х		
						STOP.	Reschedule failed transfer.	1	0	Х	
			1	0	x	A slave byte was received; ACK	Acknowledge received byte; Read SMB0DAT.	0	0	1	
	0000				requested.	Do not acknowledge received byte.	0	0	0		
		1	1	х	Lost arbitration while transmitting	Abort failed transfer.	0	0	0		
					a data byte as master.	Reschedule failed transfer.	1	0	0		

Table 17.4. SMBus Status Decoding (Continued)

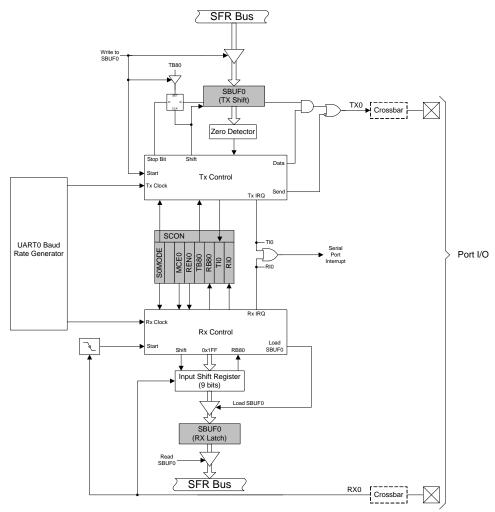


# 18. UART0

UART0 is an asynchronous, full duplex serial port offering modes 1 and 3 of the standard 8051 UART. Enhanced baud rate support allows a wide range of clock sources to generate standard baud rates (details in **Section "18.1. Enhanced Baud Rate Generation" on page 206**). Received data buffering allows UART0 to start reception of a second incoming data byte before software has finished reading the previous data byte.

UART0 has two associated SFRs: Serial Control Register 0 (SCON0) and Serial Data Buffer 0 (SBUF0). The single SBUF0 location provides access to both transmit and receive registers. Writes to SBUF0 always access the Transmit register. Reads of SBUF0 always access the buffered Receive register; it is not possible to read data from the Transmit register.

With UART0 interrupts enabled, an interrupt is generated each time a transmit is completed (TI0 is set in SCON0), or a data byte has been received (RI0 is set in SCON0). The UART0 interrupt flags are not cleared by hardware when the CPU vectors to the interrupt service routine. They must be cleared manually by software, allowing software to determine the cause of the UART0 interrupt (transmit complete or receive complete).







### 18.1. Enhanced Baud Rate Generation

The UART0 baud rate is generated by Timer 1 in 8-bit auto-reload mode. The TX clock is generated by TL1; the RX clock is generated by a copy of TL1 (shown as RX Timer in Figure 18.2), which is not user-accessible. Both TX and RX Timer overflows are divided by two to generate the TX and RX baud rates. The RX Timer runs when Timer 1 is enabled, and uses the same reload value (TH1). However, an RX Timer reload is forced when a START condition is detected on the RX pin. This allows a receive to begin any time a START is detected, independent of the TX Timer state.

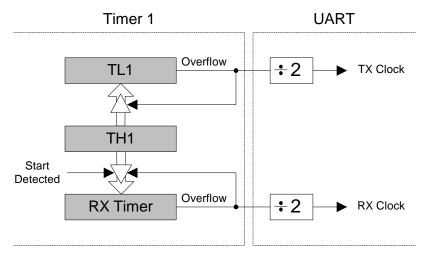


Figure 18.2. UART0 Baud Rate Logic

Timer 1 should be configured for Mode 2, 8-bit auto-reload (see Section "21.1.3. Mode 2: 8-bit Counter/ Timer with Auto-Reload" on page 237). The Timer 1 reload value should be set so that overflows will occur at two times the desired UART baud rate frequency. Note that Timer 1 may be clocked by one of six sources: SYSCLK, SYSCLK / 4, SYSCLK / 12, SYSCLK / 48, the external oscillator clock / 8, or an external input T1. For any given Timer 1 clock source, the UART0 baud rate is determined by Equation 18.1.

$$UartBaudRate = \frac{T1_{CLK}}{(256 - T1H)} \times \frac{1}{2}$$

# Equation 18.1. UART0 Baud Rate

Where  $T1_{CLK}$  is the frequency of the clock supplied to Timer 1, and T1H is the high byte of Timer 1 (reload value). Timer 1 clock frequency is selected as described in **Section "21. Timers" on page 235.** A quick reference for typical baud rates using the internal oscillator is given in Table 18.1. Note that the internal oscillator may still generate the system clock if an external oscillator is driving Timer 1.

#### 18.2. Operational Modes

UART0 provides standard asynchronous, full duplex communication. The UART mode (8-bit or 9-bit) is selected by the S0MODE bit (SCON0.7). Typical UART connection options are shown below.



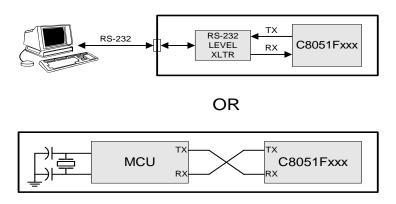


Figure 18.3. UART Interconnect Diagram

#### 18.2.1. 8-Bit UART

8-Bit UART mode uses a total of 10 bits per data byte: one start bit, eight data bits (LSB first), and one stop bit. Data are transmitted LSB first from the TX0 pin and received at the RX0 pin. On receive, the eight data bits are stored in SBUF0 and the stop bit goes into RB80 (SCON0.2).

Data transmission begins when software writes a data byte to the SBUF0 register. The TI0 Transmit Interrupt Flag (SCON0.1) is set at the end of the transmission (the beginning of the stop-bit time). Data reception can begin any time after the REN0 Receive Enable bit (SCON0.4) is set to logic 1. After the stop bit is received, the data byte will be loaded into the SBUF0 receive register if the following conditions are met: RI0 must be logic 0, and if MCE0 is logic 1, the stop bit must be logic 1. In the event of a receive data overrun, the first received 8 bits are latched into the SBUF0 receive register and the following overrun data bits are lost.

If these conditions are met, the eight bits of data is stored in SBUF0, the stop bit is stored in RB80 and the RI0 flag is set. If these conditions are not met, SBUF0 and RB80 will not be loaded and the RI0 flag will not be set. An interrupt will occur if enabled when either TI0 or RI0 is set.

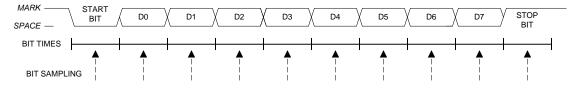


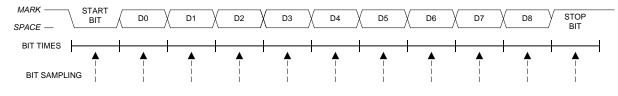
Figure 18.4. 8-Bit UART Timing Diagram



#### 18.2.2. 9-Bit UART

9-bit UART mode uses a total of eleven bits per data byte: a start bit, 8 data bits (LSB first), a programmable ninth data bit, and a stop bit. The state of the ninth transmit data bit is determined by the value in TB80 (SCON0.3), which is assigned by user software. It can be assigned the value of the parity flag (bit P in register PSW) for error detection, or used in multiprocessor communications. On receive, the ninth data bit goes into RB80 (SCON0.2) and the stop bit is ignored.

Data transmission begins when an instruction writes a data byte to the SBUF0 register. The TI0 Transmit Interrupt Flag (SCON0.1) is set at the end of the transmission (the beginning of the stop-bit time). Data reception can begin any time after the REN0 Receive Enable bit (SCON0.4) is set to '1'. After the stop bit is received, the data byte will be loaded into the SBUF0 receive register if the following conditions are met: (1) RI0 must be logic 0, and (2) if MCE0 is logic 1, the 9th bit must be logic 1 (when MCE0 is logic 0, the state of the ninth data bit is unimportant). If these conditions are met, the eight bits of data are stored in SBUF0, the ninth bit is stored in RB80, and the RI0 flag is set to '1'. A UART0 interrupt will occur if enabled when either TI0 or RI0 is set to '1'.





#### 18.3. Multiprocessor Communications

9-Bit UART mode supports multiprocessor communication between a master processor and one or more slave processors by special use of the ninth data bit. When a master processor wants to transmit to one or more slaves, it first sends an address byte to select the target(s). An address byte differs from a data byte in that its ninth bit is logic 1; in a data byte, the ninth bit is always set to logic 0.

Setting the MCE0 bit (SCON0.5) of a slave processor configures its UART such that when a stop bit is received, the UART will generate an interrupt only if the ninth bit is logic 1 (RB80 = 1) signifying an address byte has been received. In the UART interrupt handler, software will compare the received address with the slave's own assigned 8-bit address. If the addresses match, the slave will clear its MCE0 bit to enable interrupts on the reception of the following data byte(s). Slaves that weren't addressed leave their MCE0 bits set and do not generate interrupts on the reception of the following data byte(s) addressed slave resets its MCE0 bit to ignore all transmissions until it receives the next address byte.

Multiple addresses can be assigned to a single slave and/or a single address can be assigned to multiple slaves, thereby enabling "broadcast" transmissions to more than one slave simultaneously. The master processor can be configured to receive all transmissions or a protocol can be implemented such that the master/slave role is temporarily reversed to enable half-duplex transmission between the original master and slave(s).



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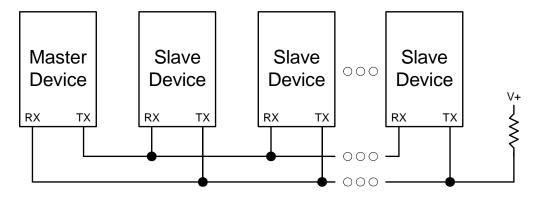


Figure 18.6. UART Multi-Processor Mode Interconnect Diagram



# C8051F340/1/2/3/4/5/6/7/8/9/A/B/C/D

R/W	R	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value					
SOMOD	-	MCE0	REN0	TB80	RB80	TI0	RI0	0100000					
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	Bit Addressable					
							SFR Addres						
Bit7:		arial Dart O	Oneration	Mada									
BIT :	S0MODE: Serial Port 0 Operation Mode. This bit selects the UART0 Operation Mode.												
	0: 8-bit UAR		•										
	1: 9-bit UAR												
Bit6:	UNUSED. R												
Bit5:													
	MCE0: Multiprocessor Communication Enable. The function of this bit is dependent on the Serial Port 0 Operation Mode.												
	S0MODE =					•							
	0: Logic level of stop bit is ignored. 1: RI0 will only be activated if stop bit is logic level 1.												
	SOMODE = 1: Multiprocessor Communications Enable.												
	0: Logic level of ninth bit is ignored.												
	1: R	10 is set and	d an interru	pt is genera	ited only who	en the nint	h bit is log	ic 1.					
Bit4:	REN0: Receive Enable.												
	This bit enables/disables the UART receiver.												
	0: UART0 re												
	1: UART0 re												
Bit3:	TB80: Ninth												
	The logic lev			•				ARI Mode. I					
D:40.	is not used in 8-bit UART Mode. Set or cleared by software as required.												
Bit2:	RB80: Ninth Receive Bit.												
	RB80 is assigned the value of the STOP bit in Mode 0; it is assigned the value of the 9th data bit in Mode 1.												
Bit1:			log										
DILT.	TIO: Transmit Interrupt Flag.												
	Set by hardware when a byte of data has been transmitted by UART0 (after the 8th bit in 8 bit UART Mode) when the												
	8-bit UART Mode, or at the beginning of the STOP bit in 9-bit UART Mode). When the UART0 interrupt is enabled, setting this bit causes the CPU to vector to the UART0 interrupt												
	service routine. This bit must be cleared manually by software.												
Bit0:	RI0: Receive												
	Set to '1' by	•	•	of data has	been receive	ed by UAR	T0 (set at t	he STOP bi					
	sampling tim												
	to vector to	,		•		•							
	ware.												

# SFR Definition 18.1. SCON0: Serial Port 0 Control



# SFR Definition 18.2. SBUF0: Serial (UART0) Port Data Buffer

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
							SFR Address	s: 0x99
Bits7–0:	SBUF0[7:0]: This SFR acc data is writte sion. Writing tents of the r	cesses two n to SBUF a byte to S	registers; a 0, it goes to BUF0 initia	transmit sh the transm	hift register a it shift regis	ter and is h	eld for seria	al transmis-



	Target	Actual	Baud	Oscillator	Timer Clock	SCA1-SCA0	T1M*	Timer 1
	Baud	Baud	Rate Error	Divide	Source	(pre-scale		Reload
	Rate (bps)	Rate (bps)		Factor		select*		Value (hex)
	230400	230769	0.16%	52	SYSCLK	XX	1	0xE6
MHz	115200	115385	0.16%	104	SYSCLK	XX	1	0xCC
	57600	57692	0.16%	208	SYSCLK	XX	1	0x98
12	28800	28846	0.16%	416	SYSCLK	XX	1	0x30
Ш Х	14400	14423	0.16%	832	SYSCLK/4	01	0	0x98
SYSCLK	9600	9615	0.16%	1248	SYSCLK/4	01	0	0x64
ŝ	2400	2404	0.16%	4992	SYSCLK / 12	00	0	0x30
S	1200	1202	0.16%	9984	SYSCLK/48	10	0	0x98
	230400	230769	0.16%	104	SYSCLK	XX	1	0xCC
보	115200	115385	0.16%	208	SYSCLK	XX	1	0x98
MHz	57600	57692	0.16%	416	SYSCLK	XX	1	0x30
24	28800	28846	0.16%	832	SYSCLK/4	01	0	0x98
	14400	14423	0.16%	1664	SYSCLK/4	01	0	0x30
SCLK	9600	9615	0.16%	2496	SYSCLK / 12	00	0	0x98
ŝ	2400	2404	0.16%	9984	SYSCLK/48	10	0	0x98
SΥ	1200	1202	0.16%	19968	SYSCLK/48	10	0	0x30
μz	230400	230769	0.16%	208	SYSCLK	XX	1	0x98
MHz	115200	115385	0.16%	416	SYSCLK	XX	1	0x30
48	57600	57692	0.16%	832	SYSCLK/4	01	0	0x98
Ш	28800	28846	0.16%	1664	SYSCLK/4	01	0	0x30
SCLK	14400	14388	0.08%	3336	SYSCLK / 12	00	0	0x75
SC	9600	9615	0.16%	4992	SYSCLK / 12	00	0	0x30
SY	2400	2404	0.16%	19968	SYSCLK / 48	10	0	0x30

# Table 18.1. Timer Settings for Standard Baud Rates Using the Internal Oscillator

X = Don't care

\*Note: SCA1-SCA0 and T1M define the Timer Clock Source. Bit definitions for these values can be found in Section 21.1.



# 19. UART1 (C8051F340/1/4/5/8/A/B/C Only)

UART1 is an asynchronous, full duplex serial port offering a variety of data formatting options. A dedicated baud rate generator with a 16-bit timer and selectable prescaler is included, which can generate a wide range of baud rates (details in **Section "19.1. Baud Rate Generator" on page 214**). A received data FIFO allows UART1 to receive up to three data bytes before data is lost and an overflow occurs.

UART1 has six associated SFRs. Three are used for the Baud Rate Generator (SBCON1, SBRLH1, and SBRLL1), two are used for data formatting, control, and status functions (SCON1, SMOD1), and one is used to send and receive data (SBUF1). The single SBUF1 location provides access to both the transmit holding register and the receive FIFO. Writes to SBUF1 always access the Transmit Holding Register. Reads of SBUF1 always access the first byte of the Receive FIFO; it is not possible to read data from the Transmit Holding Register.

With UART1 interrupts enabled, an interrupt is generated each time a transmit is completed (TI1 is set in SCON1), or a data byte has been received (RI1 is set in SCON1). The UART1 interrupt flags are not cleared by hardware when the CPU vectors to the interrupt service routine. They must be cleared manually by software, allowing software to determine the cause of the UART1 interrupt (transmit complete or receive complete). Note that if additional bytes are available in the Receive FIFO, the RI1 bit cannot be cleared by software.

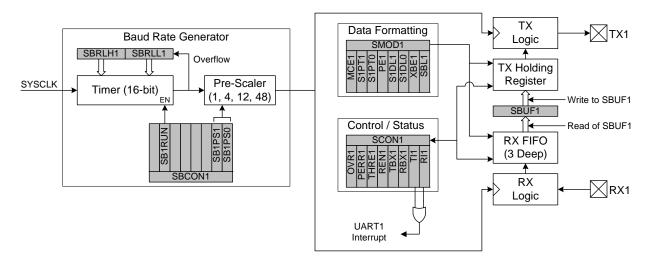


Figure 19.1. UART1 Block Diagram



#### 19.1. Baud Rate Generator

The UART1 baud rate is generated by a dedicated 16-bit timer which runs from the controller's core clock (SYSCLK), and has prescaler options of 1, 4, 12, or 48. The timer and prescaler options combined allow for a wide selection of baud rates over many SYSCLK frequencies.

The baud rate generator is configured using three registers: SBCON1, SBRLH1, and SBRLL1. The UART1 Baud Rate Generator Control Register (SBCON1, SFR Definition 19.4) enables or disables the baud rate generator, and selects the prescaler value for the timer. The baud rate generator must be enabled for UART1 to function. Registers SBRLH1 and SBRLL1 contain a 16-bit reload value for the dedicated 16-bit timer. The internal timer counts up from the reload value on every clock tick. On timer overflows (0xFFFF to 0x0000), the timer is reloaded. For reliable UART operation, it is recommended that the UART baud rate is not configured for baud rates faster than SYSCLK/16. The baud rate for UART1 is defined in Equation 19.1.

Baud Rate =  $\frac{\text{SYSCLK}}{(65536 - (\text{SBRLH1:SBRLL1}))} \times \frac{1}{2} \times \frac{1}{\text{Prescaler}}$ 

#### Equation 19.1. UART1 Baud Rate

A quick reference for typical baud rates and system clock frequencies is given in Table 19.1.

	Target Baud Rate (bps)	Actual Baud Rate (bps)	Baud Rate Error	Oscillator Divide Factor	SB1PS[1:0] (Prescaler Bits)	Reload Value in SBRLH1:SBRLL1
	230400	230769	0.16%	52	11	0xFFE6
Ν	115200	115385	0.16%	104	11	0xFFCC
MHz	57600	57692	0.16%	208	11	0xFF98
12	28800	28846	0.16%	416	11	0xFF30
Ш	14400	14388	0.08%	834	11	0xFE5F
SCLK	9600	9600	0.0%	1250	11	0xFD8F
SC	2400	2400	0.0%	5000	11	0xF63C
SY	1200	1200	0.0%	10000	11	0xEC78
	230400	230769	0.16%	104	11	0xFFCC
부	115200	115385	0.16%	208	11	0xFF98
MHz	57600	57692	0.16%	416	11	0xFF30
24	28800	28777	0.08%	834	11	0xFE5F
	14400	14406	0.04%	1666	11	0xFCBF
SYSCLK	9600	9600	0.0%	2500	11	0xFB1E
S S	2400	2400	0.0%	10000	11	0xEC78
S	1200	1200	0.0%	20000	11	0xD8F0
	230400	230769	0.16%	208	11	0xFF98
부	115200	115385	0.16%	416	11	0xFF30
MHz	57600	57554	0.08%	834	11	0xFE5F
48	28800	28812	0.04%	1666	11	0xFCBF
	14400	14397	0.02%	3334	11	0xF97D
SCLK	9600	9600	0.0%	5000	11	0xF63C
Š	2400	2400	0.0%	20000	11	0xD8F0
SY	1200	1200	0.0%	40000	11	0xB1E0

### Table 19.1. Baud Rate Generator Settings for Standard Baud Rates



### 19.2. Data Format

UART1 has a number of available options for data formatting. Data transfers begin with a start bit (logic low), followed by the data bits (sent LSB-first), a parity or extra bit (if selected), and end with one or two stop bits (logic high). The data length is variable between 5 and 8 bits. A parity bit can be appended to the data, and automatically generated and detected by hardware for even, odd, mark, or space parity. The stop bit length is selectable between short (1 bit time) and long (1.5 or 2 bit times), and a multi-processor communication mode is available for implementing networked UART buses. All of the data formatting options can be configured using the SMOD1 register, shown in SFR Definition 19.2. Figure 19.2 shows the timing for a UART1 transaction without parity or an extra bit enabled. Figure 19.3 shows the timing for a UART1 transaction with parity enabled (PE1 = 1). Figure 19.4 is an example of a UART1 transaction when the extra bit is enabled (XBE1 = 1). Note that the extra bit feature is not available when parity is enabled, and the second stop bit is only an option for data lengths of 6, 7, or 8 bits.

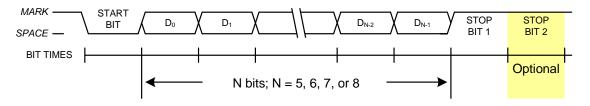


Figure 19.2. UART1 Timing Without Parity or Extra Bit

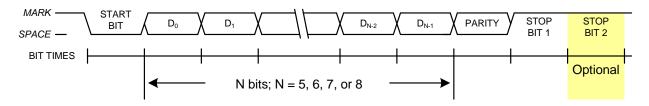


Figure 19.3. UART1 Timing With Parity

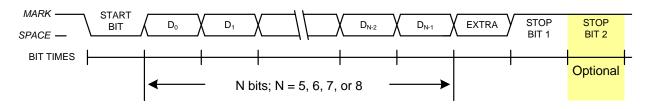


Figure 19.4. UART1 Timing With Extra Bit



# **19.3.** Configuration and Operation

UART1 provides standard asynchronous, full duplex communication. It can operate in a point-to-point serial communications application, or as a node on a multi-processor serial interface. To operate in a point-to-point application, where there are only two devices on the serial bus, the MCE1 bit in SMOD1 should be cleared to '0'. For operation as part of a multi-processor communications bus, the MCE1 and XBE1 bits should both be set to '1'. In both types of applications, data is transmitted from the microcontroller on the TX1 pin, and received on the RX1 pin. The TX1 and RX1 pins are configured using the crossbar and the Port I/O registers, as detailed in Section "15. Port Input/Output" on page 142.

In typical UART communications, The transmit (TX) output of one device is connected to the receive (RX) input of the other device, either directly or through a bus transceiver, as shown in Figure 19.5.

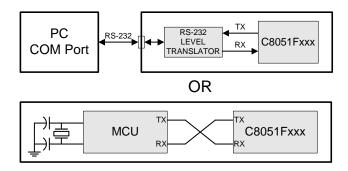


Figure 19.5. Typical UART Interconnect Diagram

#### 19.3.1. Data Transmission

Data transmission is double-buffered, and begins when software writes a data byte to the SBUF1 register. Writing to SBUF1 places data in the Transmit Holding Register, and the Transmit Holding Register Empty flag (THRE1) will be cleared to '0'. If the UARTs shift register is empty (i.e., no transmission is in progress) the data will be placed in the shift register, and the THRE1 bit will be set to '1'. If a transmission is in progress, the data will remain in the Transmit Holding Register until the current transmission is complete. The TI1 Transmit Interrupt Flag (SCON1.1) will be set at the end of any transmission (the beginning of the stop-bit time). If enabled, an interrupt will occur when TI1 is set.

If the extra bit function is enabled (XBE1 = '1') and the parity function is disabled (PE1 = '0'), the value of the TBX1 (SCON1.3) bit will be sent in the extra bit position. When the parity function is enabled (PE1 = '1'), hardware will generate the parity bit according to the selected parity type (selected with S1PT[1:0]), and append it to the data field. Note: when parity is enabled, the extra bit function is not available.

#### 19.3.2. Data Reception

Data reception can begin any time after the REN1 Receive Enable bit (SCON1.4) is set to logic 1. After the stop bit is received, the data byte will be stored in the receive FIFO if the following conditions are met: the receive FIFO (3 bytes deep) must not be full, and the stop bit(s) must be logic 1. In the event that the receive FIFO is full, the incoming byte will be lost, and a Receive FIFO Overrun Error will be generated (OVR1 in register SCON1 will be set to logic 1). If the stop bit(s) were logic 0, the incoming data will not be stored in the receive FIFO. If the reception conditions are met, the data is stored in the receive FIFO, and the RI1 flag will be set. Note: when MCE1 = '1', RI1 will only be set if the extra bit was equal to '1'. Data can be read from the receive FIFO by reading the SBUF1 register. The SBUF1 register represents the oldest byte in the FIFO. After SBUF1 is read, the next byte in the FIFO is immediately loaded into SBUF1, and



space is made available in the FIFO for another incoming byte. If enabled, an interrupt will occur when RI1 is set. RI1 can only be cleared to '0' by software when there is no more information in the FIFO. The recommended procedure to empty the FIFO contents is as follows:

- 1. Clear RI1 to '0'.
- 2. Read SBUF1.
- 3. Check RI1, and repeat at step 1 if RI1 is set to '1'.

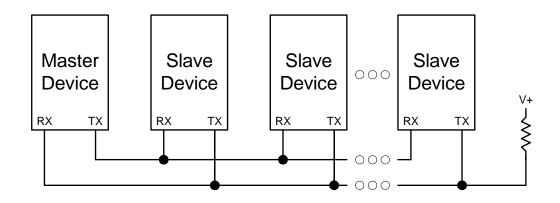
If the extra bit function is enabled (XBE1 = '1') and the parity function is disabled (PE1 = '0'), the extra bit for the oldest byte in the FIFO can be read from the RBX1 bit (SCON1.2). If the extra bit function is not enabled, the value of the stop bit for the oldest FIFO byte will be presented in RBX1. When the parity function is enabled (PE1 = '1'), hardware will check the received parity bit against the selected parity type (selected with S1PT[1:0]) when receiving data. If a byte with parity error is received, the PERR1 flag will be set to '1'. This flag must be cleared by software. Note: when parity is enabled, the extra bit function is not available.

#### 19.3.3. Multiprocessor Communications

UART1 supports multiprocessor communication between a master processor and one or more slave processors by special use of the extra data bit. When a master processor wants to transmit to one or more slaves, it first sends an address byte to select the target(s). An address byte differs from a data byte in that its extra bit is logic 1; in a data byte, the extra bit is always set to logic 0.

Setting the MCE1 bit (SMOD1.7) of a slave processor configures its UART such that when a stop bit is received, the UART will generate an interrupt only if the extra bit is logic 1 (RBX1 = 1) signifying an address byte has been received. In the UART interrupt handler, software will compare the received address with the slave's own assigned address. If the addresses match, the slave will clear its MCE1 bit to enable interrupts on the reception of the following data byte(s). Slaves that weren't addressed leave their MCE1 bits set and do not generate interrupts on the reception of the following data bytes, thereby ignoring the data. Once the entire message is received, the addressed slave resets its MCE1 bit to ignore all transmissions until it receives the next address byte.

Multiple addresses can be assigned to a single slave and/or a single address can be assigned to multiple slaves, thereby enabling "broadcast" transmissions to more than one slave simultaneously. The master processor can be configured to receive all transmissions or a protocol can be implemented such that the master/slave role is temporarily reversed to enable half-duplex transmission between the original master and slave(s).





## Figure 19.6. UART Multi-Processor Mode Interconnect Diagram

R/W	R/W	R	R/W	R/W	R/W	R/W	R/W	Reset Value
OVR1	PERR1	THRE1	REN1	TBX1	RBX1	TI1	RI1	00100000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
							SFR Addres	ss: 0xD2
Bit7:	OVR1: Rece							
	This bit is us				run conditio	on.		
	0: Receive F							
	1: Receive F	-IFO Overru	in has occu	rred (an inc	oming char	acter was d	iscarded o	tue to a full
	FIFO).	t ha alaaraa	to '0' by or	ftworo				
Bit6:	This bit must PERR1: Par			Jilware.				
Dito.	When parity	•	-	sed to indic	ate that a na	arity error h	as occurre	d It is set to
	'1' when the							
	0: Parity Erro		•					any type.
	1: Parity Erro							
	This bit must			oftware.				
Bit5:	THRE1: Trai	nsmit Holdir	ng Register	Empty Flag	<b>]</b> .			
	0: Transmit I	Holding Reg	jister not E	mpty - do no	ot write to S	BUF1.		
	1: Transmit I	Holding Reg	gister Empt	y - it is safe	to write to S	SBUF1.		
Bit4:	REN1: Rece							
	This bit enab		s the UART	receiver. W	/hen disable	ed, bytes ca	n still be re	ead from the
	receive FIFC							
	0: UART1 re	•						
D:+0.	1: UART1 re							
Bit3:	TBX1: Extra The logic lev			igned to the	ovtro tropo	mingion hit	when VD	
	'1'. This bit is			-		5111551011 DIL		
Bit2:	RBX1: Extra			is enabled.				
	RBX1 is ass			extra bit who	en XBE1 is :	set to '1'. If	XBE1 is cl	leared to '0'.
	RBX1 will be							
	enabled.	0	0					,
Bit1:	TI1: Transmi	it Interrupt F	lag.					
	Set to a '1' b	y hardware	after data	has been tra	ansmitted, a	at the begin	ning of the	STOP bit.
	When the U/		•	-				
	UART1 inter			nis bit must	be cleared	manually by	/ software.	
Bit0:	RI1: Receive	•	-					
	Set to '1' by I		•			•		
	sampling tim	•		•		-		
	to vector to t							
	ware. Note the last byte				•			FIFU. Alter
	ine last byle	nas been s			500F1, KI	i can be Cl	ealeu.	

## SFR Definition 19.1. SCON1: UART1 Control



## SFR Definition 19.2. SMOD1: UART1 Mode

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
MCE1	S1PT1	S1PT0	PE1	S1DL1	S1DL0	XBE1	SBL1	00001100
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
							SFR Addres	ss: 0xE5
Bit7:	MCE1: Multi							
	0: RI will be							
	1: RI will be	activated if	stop bit(s)	and extra bi	t are '1' (ex	tra bit must	be enable	d using
	XBE1).	mation is no	t ovollable			a anablad		
Ritce 5:	Note: This fu S1PT[1:0]: F		ot available	when hard	vare parity	is enabled.		
Dit30-3.	00: Odd	anty type.						
	01: Even							
	10: Mark							
	11: Space							
Bit4:	PE1: Parity	Enable.						
	This bit activ				nd checking	g. The parit	y type is se	elected by
	bits S1PT1-0		•	ed.				
	0: Hardware							
Dite 2 Di	1: Hardware							
Bits3-2:	S1DL[1:0]: E 00: 5-bit data	•	•					
	01: 6-bit data							
	10: 7-bit data							
	11: 8-bit data							
Bit1:	XBE1: Extra	Bit Enable						
	When enable	ed, the valu	e of TBX1	will be appe	nded to the	data field.		
	0: Extra Bit I							
Duc	1: Extra Bit I							
Bit0:	SBL1: Stop	-		L !4 4!				
	0: Short - Sto 1: Long - Sto				ta longth –	67 or 9 hi	te) or 1 5	hit timos
	(data length	•		ni iines (ua	ia ieriyiri =	0, <i>1</i> , 01 8 DI	15), 01 1.5	
	(data longth	- 0 010).						



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
							SFR Addres	s: 0xD3
	SBUF1[7:0]: This SFR is UART1 rece Write: Writing first goes to t transmit shift be written ag Read: Readi the receive F the FIFO. If t '1', even after	used to bo ive FIFO. g a byte to the Transm t register is gain. ing SBUF1 FIFO is retu there are a	th send data SBUF1 initi it Holding R available, c retrieves da urned, and re dditional byte	a from the L ates the tra egister, wh data is trans ata from the emoved fro tes availabl	JART and to nsmission. V ere it is held ferred into t e receive FIF m the FIFO.	When data for serial he shift re O. When Up to thre	a is written t transmissio gister, and read, the ol ee bytes ma	to SBUF1, it n. When the SBUF1 may dest byte in ay be held in

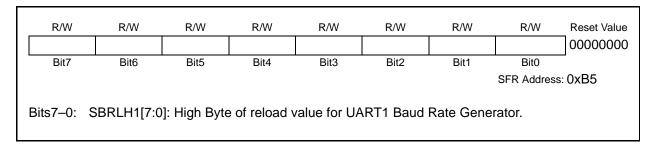
## SFR Definition 19.3. SBUF1: UART1 Data Buffer

## SFR Definition 19.4. SBCON1: UART1 Baud Rate Generator Control

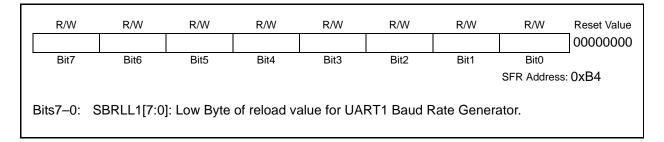
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Reserve	d SB1RUN	Reserved	Reserved	Reserved	Reserved	SB1PS1	SB1PS0	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	-
							SFR Address:	0xAC
Bit7: Bit6: Bits5–2: Bits1–0:	RESERVED SB1RUN: Ba 0: Baud Rate 1: Baud Rate RESERVED SB1PS[1:0]: 00: Prescale 01: Prescale 10: Prescale 11: Prescale	aud Rate G e Generator e Generator : Read = 00 Baud Rate r = 12 r = 4 r = 48	enerator En is disablec is enabled 000b; Must	able. I. UART1 w write 0000b		on.		



### SFR Definition 19.5. SBRLH1: UART1 Baud Rate Generator High Byte



## SFR Definition 19.6. SBRLL1: UART1 Baud Rate Generator Low Byte





## 20. Enhanced Serial Peripheral Interface (SPI0)

The Enhanced Serial Peripheral Interface (SPI0) provides access to a flexible, full-duplex synchronous serial bus. SPI0 can operate as a master or slave device in both 3-wire or 4-wire modes, and supports multiple masters and slaves on a single SPI bus. The slave-select (NSS) signal can be configured as an input to select SPI0 in slave mode, or to disable Master Mode operation in a multi-master environment, avoiding contention on the SPI bus when more than one master attempts simultaneous data transfers. NSS can also be configured as a chip-select output in master mode, or disabled for 3-wire operation. Additional general purpose port I/O pins can be used to select multiple slave devices in master mode.

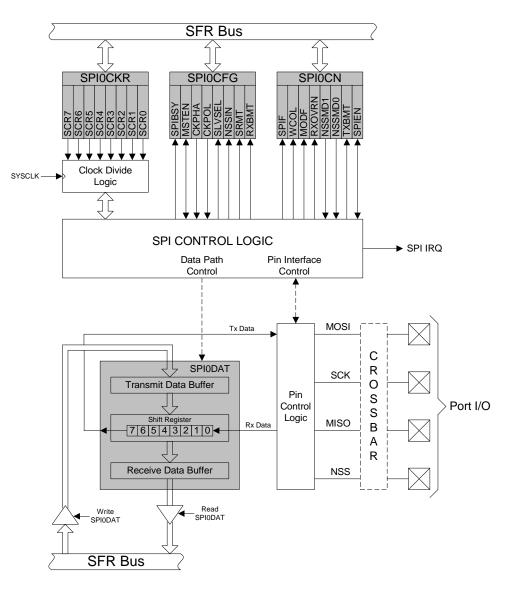


Figure 20.1. SPI Block Diagram



## 20.1. Signal Descriptions

The four signals used by SPI0 (MOSI, MISO, SCK, NSS) are described below.

### 20.1.1. Master Out, Slave In (MOSI)

The master-out, slave-in (MOSI) signal is an output from a master device and an input to slave devices. It is used to serially transfer data from the master to the slave. This signal is an output when SPI0 is operating as a master and an input when SPI0 is operating as a slave. Data is transferred most-significant bit first. When configured as a master, MOSI is driven by the MSB of the shift register in both 3- and 4-wire mode.

### 20.1.2. Master In, Slave Out (MISO)

The master-in, slave-out (MISO) signal is an output from a slave device and an input to the master device. It is used to serially transfer data from the slave to the master. This signal is an input when SPI0 is operating as a master and an output when SPI0 is operating as a slave. Data is transferred most-significant bit first. The MISO pin is placed in a high-impedance state when the SPI module is disabled and when the SPI operates in 4-wire mode as a slave that is not selected. When acting as a slave in 3-wire mode, MISO is always driven by the MSB of the shift register.

### 20.1.3. Serial Clock (SCK)

The serial clock (SCK) signal is an output from the master device and an input to slave devices. It is used to synchronize the transfer of data between the master and slave on the MOSI and MISO lines. SPI0 generates this signal when operating as a master. The SCK signal is ignored by a SPI slave when the slave is not selected (NSS = 1) in 4-wire slave mode.

#### 20.1.4. Slave Select (NSS)

The function of the slave-select (NSS) signal is dependent on the setting of the NSSMD1 and NSSMD0 bits in the SPI0CN register. There are three possible modes that can be selected with these bits:

- NSSMD[1:0] = 00: 3-Wire Master or 3-Wire Slave Mode: SPI0 operates in 3-wire mode, and NSS is disabled. When operating as a slave device, SPI0 is always selected in 3-wire mode. Since no select signal is present, SPI0 must be the only slave on the bus in 3-wire mode. This is intended for point-to-point communication between a master and one slave.
- NSSMD[1:0] = 01: 4-Wire Slave or Multi-Master Mode: SPI0 operates in 4-wire mode, and NSS is enabled as an input. When operating as a slave, NSS selects the SPI0 device. When operating as a master, a 1-to-0 transition of the NSS signal disables the master function of SPI0 so that multiple master devices can be used on the same SPI bus.
- NSSMD[1:0] = 1x: 4-Wire Master Mode: SPI0 operates in 4-wire mode, and NSS is enabled as an output. The setting of NSSMD0 determines what logic level the NSS pin will output. This configuration should only be used when operating SPI0 as a master device.

See Figure 20.2, Figure 20.3, and Figure 20.4 for typical connection diagrams of the various operational modes. Note that the setting of NSSMD bits affects the pinout of the device. When in 3-wire master or 3-wire slave mode, the NSS pin will not be mapped by the crossbar. In all other modes, the NSS signal will be mapped to a pin on the device. See Section "15. Port Input/Output" on page 142 for general purpose port I/O and crossbar information.



## 20.2. SPI0 Master Mode Operation

A SPI master device initiates all data transfers on a SPI bus. SPI0 is placed in master mode by setting the Master Enable flag (MSTEN, SPI0CFG.6). Writing a byte of data to the SPI0 data register (SPI0DAT) when in master mode writes to the transmit buffer. If the SPI shift register is empty, the byte in the transmit buffer is moved to the shift register, and a data transfer begins. The SPI0 master immediately shifts out the data serially on the MOSI line while providing the serial clock on SCK. The SPIF (SPI0CN.7) flag is set to logic 1 at the end of the transfer. If interrupts are enabled, an interrupt request is generated when the SPIF flag is set. While the SPI0 master transfers data to a slave on the MOSI line, the addressed SPI slave device simultaneously transfers the contents of its shift register to the SPI master on the MISO line in a full-duplex operation. Therefore, the SPIF flag serves as both a transmit-complete and receive-data-ready flag. The data byte received from the slave is transferred MSB-first into the master's shift register. When a byte is fully shifted into the register, it is moved to the receive buffer where it can be read by the processor by reading SPI0DAT.

When configured as a master, SPI0 can operate in one of three different modes: multi-master mode, 3-wire single-master mode, and 4-wire single-master mode. The default, multi-master mode is active when NSS-MD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 1. In this mode, NSS is an input to the device, and is used to disable the master SPI0 when another master is accessing the bus. When NSS is pulled low in this mode, MSTEN (SPI0CFG.6) and SPIEN (SPI0CN.0) are set to 0 to disable the SPI master device, and a Mode Fault is generated (MODF, SPI0CN.5 = 1). Mode Fault will generate an interrupt if enabled. SPI0 must be manually re-enabled in software under these circumstances. In multi-master systems, devices will typically default to being slave devices while they are not acting as the system master device. In multi-master mode, slave devices can be addressed individually (if needed) using general-purpose I/O pins. Figure 20.2 shows a connection diagram between two master devices in multiple-master mode.

3-wire single-master mode is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 0. In this mode, NSS is not used, and is not mapped to an external port pin through the crossbar. Any slave devices that must be addressed in this mode should be selected using general-purpose I/O pins. Figure 20.3 shows a connection diagram between a master device in 3-wire master mode and a slave device.

4-wire single-master mode is active when NSSMD1 (SPI0CN.3) = 1. In this mode, NSS is configured as an output pin, and can be used as a slave-select signal for a single SPI device. In this mode, the output value of NSS is controlled (in software) with the bit NSSMD0 (SPI0CN.2). Additional slave devices can be addressed using general-purpose I/O pins. Figure 20.4 shows a connection diagram for a master device in 4-wire master mode and two slave devices.



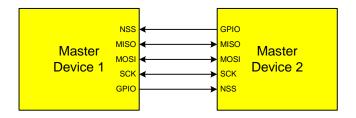


Figure 20.2. Multiple-Master Mode Connection Diagram

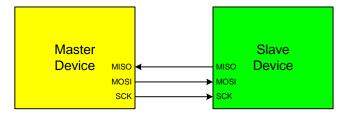


Figure 20.3. 3-Wire Single Master and Slave Mode Connection Diagram

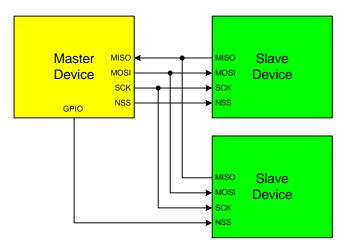


Figure 20.4. 4-Wire Single Master Mode and Slave Mode Connection Diagram



## 20.3. SPI0 Slave Mode Operation

When SPI0 is enabled and not configured as a master, it will operate as a SPI slave. As a slave, bytes are shifted in through the MOSI pin and out through the MISO pin by a master device controlling the SCK signal. A bit counter in the SPI0 logic counts SCK edges. When 8 bits have been shifted through the shift register, the SPIF flag is set to logic 1, and the byte is copied into the receive buffer. Data is read from the receive buffer by reading SPI0DAT. A slave device cannot initiate transfers. Data to be transferred to the master device is pre-loaded into the shift register by writing to SPI0DAT. Writes to SPI0DAT are double-buffered, and are placed in the transmit buffer first. If the shift register is empty, the contents of the transmit buffer will immediately be transferred into the shift register. When the shift register already contains data, the SPI will load the shift register with the transmit buffer's contents after the last SCK edge of the next (or current) SPI transfer.

When configured as a slave, SPI0 can be configured for 4-wire or 3-wire operation. The default, 4-wire slave mode, is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 1. In 4-wire mode, the NSS signal is routed to a port pin and configured as a digital input. SPI0 is enabled when NSS is logic 0, and disabled when NSS is logic 1. The bit counter is reset on a falling edge of NSS. Note that the NSS signal must be driven low at least 2 system clocks before the first active edge of SCK for each byte transfer. Figure 20.4 shows a connection diagram between two slave devices in 4-wire slave mode and a master device.

3-wire slave mode is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 0. NSS is not used in this mode, and is not mapped to an external port pin through the crossbar. Since there is no way of uniquely addressing the device in 3-wire slave mode, SPI0 must be the only slave device present on the bus. It is important to note that in 3-wire slave mode there is no external means of resetting the bit counter that determines when a full byte has been received. The bit counter can only be reset by disabling and re-enabling SPI0 with the SPIEN bit. Figure 20.3 shows a connection diagram between a slave device in 3-wire slave mode and a master device.

## 20.4. SPI0 Interrupt Sources

When SPI0 interrupts are enabled, the following four flags will generate an interrupt when they are set to logic 1:

Note that all of the following bits must be cleared by software.

- 1. The SPI Interrupt Flag, SPIF (SPI0CN.7) is set to logic 1 at the end of each byte transfer. This flag can occur in all SPI0 modes.
- 2. The Write Collision Flag, WCOL (SPI0CN.6) is set to logic 1 if a write to SPI0DAT is attempted when the transmit buffer has not been emptied to the SPI shift register. When this occurs, the write to SPI0DAT will be ignored, and the transmit buffer will not be written. This flag can occur in all SPI0 modes.
- 3. The Mode Fault Flag MODF (SPI0CN.5) is set to logic 1 when SPI0 is configured as a master, and for multi-master mode and the NSS pin is pulled low. When a Mode Fault occurs, the MSTEN and SPIEN bits in SPI0CN are set to logic 0 to disable SPI0 and allow another master device to access the bus.
- 4. The Receive Overrun Flag RXOVRN (SPI0CN.4) is set to logic 1 when configured as a slave, and a transfer is completed and the receive buffer still holds an unread byte from a previous transfer. The new byte is not transferred to the receive buffer, allowing the previously received data byte to be read. The data byte which caused the overrun is lost.



## 20.5. Serial Clock Timing

Four combinations of serial clock phase and polarity can be selected using the clock control bits in the SPI0 Configuration Register (SPI0CFG). The CKPHA bit (SPI0CFG.5) selects one of two clock phases (edge used to latch the data). The CKPOL bit (SPI0CFG.4) selects between an active-high or active-low clock. Both master and slave devices must be configured to use the same clock phase and polarity. SPI0 should be disabled (by clearing the SPIEN bit, SPI0CN.0) when changing the clock phase or polarity. The clock and data line relationships for master mode are shown in Figure 20.5. For slave mode, the clock and data relationships are shown in Figure 20.6 and Figure 20.7.

The SPI0 Clock Rate Register (SPI0CKR) as shown in SFR Definition 20.3 controls the master mode serial clock frequency. This register is ignored when operating in slave mode. When the SPI is configured as a master, the maximum data transfer rate (bits/sec) is one-half the system clock frequency or 12.5 MHz, whichever is slower. When the SPI is configured as a slave, the maximum data transfer rate (bits/sec) for full-duplex operation is 1/10 the system clock frequency, provided that the master issues SCK, NSS (in 4-wire slave mode), and the serial input data synchronously with the slave's system clock. If the master issues SCK, NSS, and the serial input data asynchronously, the maximum data transfer rate (bits/sec) must be less than 1/10 the system clock frequency. In the special case where the master only wants to transmit data to the slave and does not need to receive data from the slave (i.e. half-duplex operation), the SPI slave can receive data at a maximum data transfer rate (bits/sec) of 1/4 the system clock frequency. This is provided that the master issues SCK, NSS, and the serial input data transfer rate (bits/sec) of 1/4 the system clock frequency. This is provided that the master issues SCK, NSS, and the serial input data synchronously with the slave's system clock frequency.

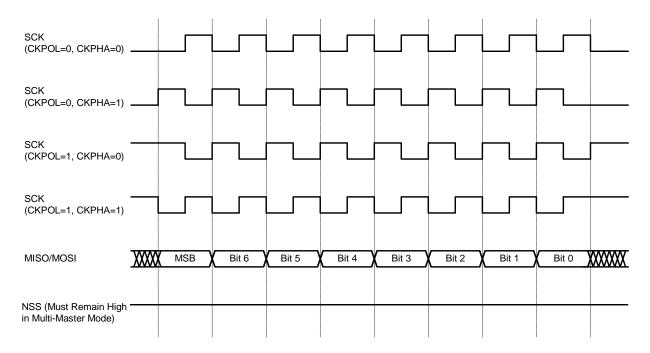
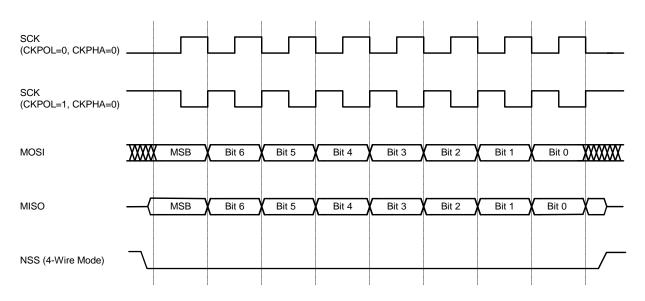
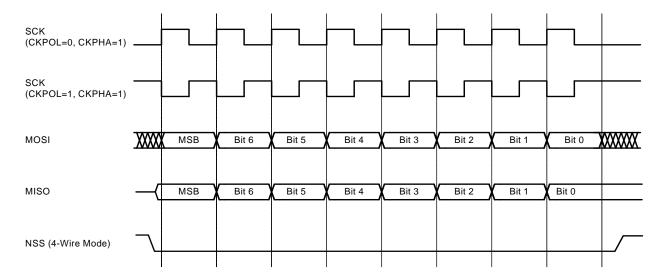


Figure 20.5. Master Mode Data/Clock Timing













## 20.6. SPI Special Function Registers

SPI0 is accessed and controlled through four special function registers in the system controller: SPI0CN Control Register, SPI0DAT Data Register, SPI0CFG Configuration Register, and SPI0CKR Clock Rate Register. The four special function registers related to the operation of the SPI0 Bus are described in the following figures.

R	R/W	R/W	R/W	R	R	R	R	Reset Value
SPIBSY		CKPHA	CKPOL	SLVSEL	NSSIN	SRMT	RXBMT	0000011
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
							SFR Address	: 0xA1
Bit 7:		I Duov (roo						
DIL 7.	SPIBSY: SP This bit is se			l transfor is	in prograss	(Mactor or	dava Mode	
Bit 6:	MSTEN: Ma				in progress	(master or	Slave Would	<del>,</del> ).
<i>n</i> t 0.	0: Disable m			n slave mod	e			
	1: Enable m				0.			
Bit 5:	CKPHA: SP		•	e a maeten				
	This bit cont			ase.				
	0: Data cent							
	1: Data cent				d.*			
Bit 4:	CKPOL: SPI							
	This bit cont	rols the SPI	0 clock pol	arity.				
	0: SCK line l	ow in idle s	tate.	-				
	1: SCK line l	high in idle :	state.					
Bit 3:	SLVSEL: Sla	ave Selecte	d Flag (read	d only).				
	This bit is se							
	is cleared to							
	instantaneou					ed version (	of the pin in	put.
Bit 2:	NSSIN: NSS							
	This bit mim					the NSS po	ort pin at the	e time that
	the register i							
Bit 1:	SRMT: Shift	•				• /		
	This bit will b		•					•
	and there is							
	receive buffe				byte is tran	isterred to t	ne snift reg	ister from
	the transmit							
Bit 0:	NOTE: SRM				Mada road	only)		
511 U.	RXBMT: Red This bit will b						nd containa	00 0014
	information.							
	this bit will re			lion availabi			nat nas not	Deenieau
	NOTE: RXB	•		r Mode				
ote: In sla	ive mode, data	on MOSI is	sampled in t	he center of	each data bit	. In master n	node, data or	n MISO is
	led one SYSC		-					

## SFR Definition 20.1. SPI0CFG: SPI0 Configuration

Note: In slave mode, data on MOSI is sampled in the center of each data bit. In master mode, data on MISO is sampled one SYSCLK before the end of each data bit, to provide maximum settling time for the slave device. See Table 20.1 for timing parameters.



## SFR Definition 20.2. SPI0CN: SPI0 Control

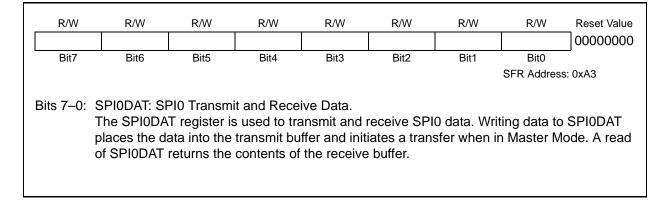
R/W SPIF	R/W WCOL	R/W MODF	R/W RXOVRN	R/W	R/W	R TXBMT	R/W SPIEN	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	Bit Addressable
							SFR Address	:: 0xF8
Bit 7:	SPIF: SPI0 I This bit is se setting this b automatically	t to logic 1 bit causes th	by hardwar he CPU to v	ector to the	SPI0 interr	upt service		
Bit 6:	WCOL: Write This bit is se been emptie ignored, and must be clea	e Collision It to logic 1 d to the SP I the transm	Flag. if a write to I shift regis nit buffer wil	SPI0DAT is ter. When th	attempted	when the tr	SPI0DAT w	rill be
Bit 5:	MODF: Mod This bit is se collision is de matically cle	e Fault Flag t to logic 1 etected (NS	g. by hardwar SS is low, M	STEN = 1,	and NSSME	D[1:0] = 01)		
Bit 4:	RXOVRN: R This bit is se fer still holds shifted into the be cleared b	t to logic 1 unread da he SPI0 sh	by hardwar ta from a pr ift register.	e (and gene evious tran	erates a SPI sfer and the	last bit of t	he current	transfer is
Bits 3–2:	NSSMD1–N Selects betw (See Section Slave Mode 00: 3-Wire S 01: 4-Wire S 1x: 4-Wire S assume the	veen the fol n "20.2. SF Operation ave or 3-w lave or Mu ingle-Maste	lowing NSS Plo Master I n" on page vire Master I lti-Master M er Mode. NS	operation ( Mode Operation) 226). Mode. NSS ode (Defau	ation" on pa signal is no lt). NSS is a	t routed to Ilways an ir	a port pin. put to the c	device.
Bit 1:	TXBMT: Tran This bit will b data in the tr indicating that	nsmit Buffe be set to log ansmit buff at it is safe	r Empty. gic 0 when r fer is transfe	erred to the	SPI shift reg	gister, this b		
Bit 0:	SPIEN: SPIC This bit enab 0: SPI disab 1: SPI enabl	oles/disable led.	es the SPI.					



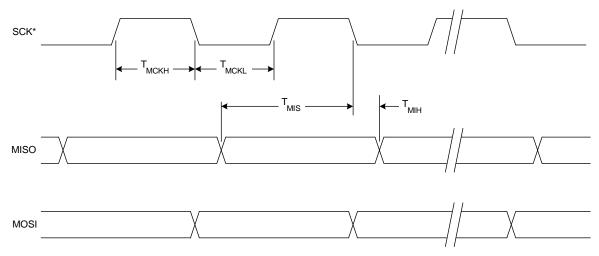
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
SCR7	SCR6	SCR5	SCR4	SCR3	SCR2	SCR1	SCR0	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
							SFR Address	: 0xA2
fo c a	SCR7–SCR( These bits d for master m lock, and is ind SPIOCK $S_{SCK} = \frac{1}{2 \times 2}$	etermine th ode operat given in th <i>R</i> is the 8-b	e frequency ion. The SC e following bit value hel	CK clock free equation, w	quency is a here SYSC	divided ver <i>LK</i> is the sy	sion of the	system
fo	or 0 <= SPI	0CKR <= 2	55					
Example: If	SYSCLK =	2 MHz and	SPIOCKR	= 0x04,				
$f_{SCK} =$ $f_{SCK} = 2$	$\frac{2000000}{2 \times (4+1)}$	<u>)</u> )						

## SFR Definition 20.3. SPI0CKR: SPI0 Clock Rate

## SFR Definition 20.4. SPI0DAT: SPI0 Data

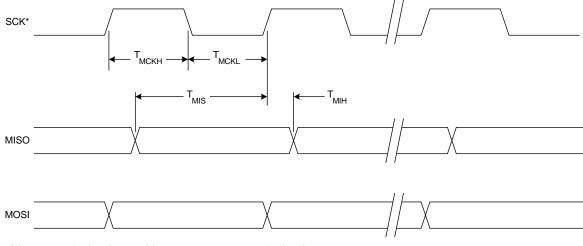






\* SCK is shown for CKPOL = 0. SCK is the opposite polarity for CKPOL = 1.

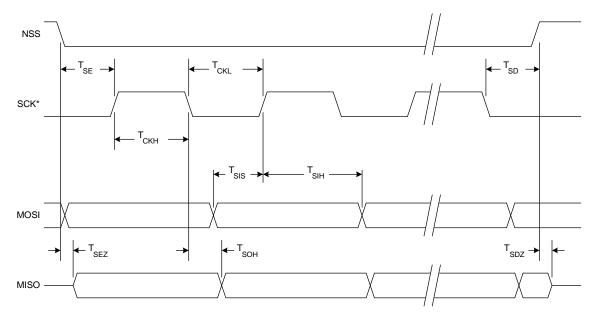




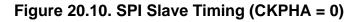
\* SCK is shown for CKPOL = 0. SCK is the opposite polarity for CKPOL = 1.

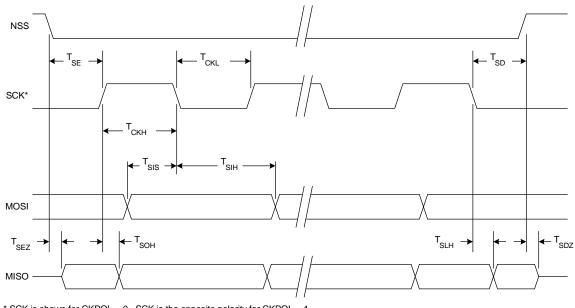
## Figure 20.9. SPI Master Timing (CKPHA = 1)





\* SCK is shown for CKPOL = 0. SCK is the opposite polarity for CKPOL = 1.





\* SCK is shown for CKPOL = 0. SCK is the opposite polarity for CKPOL = 1.

Figure 20.11. SPI Slave Timing (CKPHA = 1)



Parameter	Description	Min	Max	Units
	Master Mode Timing* (See Figure 20.8	and Figure 20.9)	I	
Т <sub>МСКН</sub>	SCK High Time	1 x T <sub>SYSCLK</sub>		ns
T <sub>MCKL</sub>	SCK Low Time	1 x T <sub>SYSCLK</sub>		ns
T <sub>MIS</sub>	MISO Valid to SCK Shift Edge	1 x T <sub>SYSCLK</sub> + 20		ns
т <sub>мін</sub>	SCK Shift Edge to MISO Change	0		ns
	Slave Mode Timing* (See Figure 20.10	and Figure 20.11)	I	
T <sub>SE</sub>	NSS Falling to First SCK Edge	2 x T <sub>SYSCLK</sub>		ns
T <sub>SD</sub>	Last SCK Edge to NSS Rising	2 x T <sub>SYSCLK</sub>		ns
T <sub>SEZ</sub>	NSS Falling to MISO Valid		4 x T <sub>SYSCLK</sub>	ns
T <sub>SDZ</sub>	NSS Rising to MISO High-Z		4 x T <sub>SYSCLK</sub>	ns
тскн	SCK High Time	5 x T <sub>SYSCLK</sub>		ns
Т <sub>СКL</sub>	SCK Low Time	5 x T <sub>SYSCLK</sub>		ns
T <sub>SIS</sub>	MOSI Valid to SCK Sample Edge	2 x T <sub>SYSCLK</sub>		ns
T <sub>SIH</sub>	SCK Sample Edge to MOSI Change	2 x T <sub>SYSCLK</sub>		ns
Т <sub>SOH</sub>	SCK Shift Edge to MISO Change		4 x T <sub>SYSCLK</sub>	ns
T <sub>SLH</sub>	Last SCK Edge to MISO Change (CKPHA = 1 ONLY)	6 x T <sub>SYSCLK</sub>	8 x T <sub>SYSCLK</sub>	ns

Table 20.1. SPI Slave Timing Parameters

\*Note:  $T_{SYSCLK}$  is equal to one period of the device system clock (SYSCLK).



## 21. Timers

Each MCU includes four counter/timers: two are 16-bit counter/timers compatible with those found in the standard 8051, and two are 16-bit auto-reload timer for use with the ADC, SMBus, USB (frame measurements), Low-Frequency Oscillator (period measurements), or for general purpose use. These timers can be used to measure time intervals, count external events and generate periodic interrupt requests. Timer 0 and Timer 1 are nearly identical and have four primary modes of operation. Timer 2 and Timer 3 offer 16-bit and split 8-bit timer functionality with auto-reload.

Timer 0 and Timer 1 Modes:	Timer 2 Modes:	Timer 3 Modes:		
13-bit counter/timer	16-bit timer with auto-reload	16-bit timer with auto-reload		
16-bit counter/timer				
8-bit counter/timer with auto-reload	Two 8-bit timers with	Two 8-bit timers with		
Two 8-bit counter/timers (Timer 0 only)	auto-reload	auto-reload		

Timers 0 and 1 may be clocked by one of five sources, determined by the Timer Mode Select bits (T1M-T0M) and the Clock Scale bits (SCA1-SCA0). The Clock Scale bits define a pre-scaled clock from which Timer 0 and/or Timer 1 may be clocked (See SFR Definition 21.3 for pre-scaled clock selection).

Timer 0/1 may then be configured to use this pre-scaled clock signal or the system clock. Timer 2 and Timer 3 may be clocked by the system clock, the system clock divided by 12, or the external oscillator clock source divided by 8.

Timer 0 and Timer 1 may also be operated as counters. When functioning as a counter, a counter/timer register is incremented on each high-to-low transition at the selected input pin (T0 or T1). Events with a frequency of up to one-fourth the system clock's frequency can be counted. The input signal need not be periodic, but it should be held at a given level for at least two full system clock cycles to ensure the level is properly sampled.

### 21.1. Timer 0 and Timer 1

Each timer is implemented as a 16-bit register accessed as two separate bytes: a low byte (TL0 or TL1) and a high byte (TH0 or TH1). The Counter/Timer Control register (TCON) is used to enable Timer 0 and Timer 1 as well as indicate status. Timer 0 interrupts can be enabled by setting the ET0 bit in the IE register (Section "9.3.5. Interrupt Register Descriptions" on page 90); Timer 1 interrupts can be enabled by setting the ET1 bit in the IE register (Section 9.3.5). Both counter/timers operate in one of four primary modes selected by setting the Mode Select bits T1M1-T0M0 in the Counter/Timer Mode register (TMOD). Each timer can be configured independently. Each operating mode is described below.

#### 21.1.1. Mode 0: 13-bit Counter/Timer

Timer 0 and Timer 1 operate as 13-bit counter/timers in Mode 0. The following describes the configuration and operation of Timer 0. However, both timers operate identically, and Timer 1 is configured in the same manner as described for Timer 0.

The TH0 register holds the eight MSBs of the 13-bit counter/timer. TL0 holds the five LSBs in bit positions TL0.4-TL0.0. The three upper bits of TL0 (TL0.7-TL0.5) are indeterminate and should be masked out or ignored when reading. As the 13-bit timer register increments and overflows from 0x1FFF (all ones) to 0x0000, the timer overflow flag TF0 (TCON.5) is set and an interrupt will occur if Timer 0 interrupts are enabled.



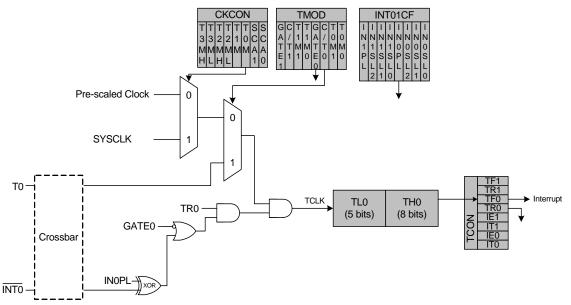
The C/T0 bit (TMOD.2) selects the counter/timer's clock source. When C/T0 is set to logic 1, high-to-low transitions at the selected Timer 0 input pin (T0) increment the timer register (Refer to Section "15.1. Priority Crossbar Decoder" on page 144 for information on selecting and configuring external I/O pins). Clearing C/T selects the clock defined by the T0M bit (CKCON.3). When T0M is set, Timer 0 is clocked by the system clock. When T0M is cleared, Timer 0 is clocked by the source selected by the Clock Scale bits in CKCON (see SFR Definition 21.3).

Setting the TR0 bit (TCON.4) enables the timer when either GATE0 (TMOD.3) is logic 0 or the input signal INT0 is active as defined by bit INOPL in register INT01CF (see SFR Definition 9.13). Setting GATE0 to '1' allows the timer to be controlled by the external input signal INT0 (see Section "9.3.5. Interrupt Register Descriptions" on page 90), facilitating pulse width measurements.

TR0	GATE0	INTO	Counter/Timer
0	Х	Х	Disabled
1	0 X		Enabled
1	1	0	Disabled
1	1	1	Enabled
X = Dc	on't Care		

Setting TR0 does not force the timer to reset. The timer registers should be loaded with the desired initial value before the timer is enabled.

TL1 and TH1 form the 13-bit register for Timer 1 in the same manner as described above for TL0 and TH0. Timer 1 is configured and controlled using the relevant TCON and TMOD bits just as with Timer 0. The input signal INT1 is used with Timer 1; the INT1 polarity is defined by bit IN1PL in register INT01CF (see SFR Definition 9.13).





### 21.1.2. Mode 1: 16-bit Counter/Timer

Mode 1 operation is the same as Mode 0, except that the counter/timer registers use all 16 bits. The counter/timers are enabled and configured in Mode 1 in the same manner as for Mode 0.



### 21.1.3. Mode 2: 8-bit Counter/Timer with Auto-Reload

Mode 2 configures Timer 0 and Timer 1 to operate as 8-bit counter/timers with automatic reload of the start value. TL0 holds the count and TH0 holds the reload value. When the counter in TL0 overflows from all ones to 0x00, the timer overflow flag TF0 (TCON.5) is set and the counter in TL0 is reloaded from TH0. If Timer 0 interrupts are enabled, an interrupt will occur when the TF0 flag is set. The reload value in TH0 is not changed. TL0 must be initialized to the desired value before enabling the timer for the first count to be correct. When in Mode 2, Timer 1 operates identically to Timer 0.

Both counter/timers are enabled and configured in Mode 2 in the same manner as Mode 0. Setting the TR0 bit (TCON.4) enables the timer when either GATE0 (TMOD.3) is logic 0 or when the input signal INT0 is active as defined by bit IN0PL in register INT01CF (see Section "9.3.2. External Interrupts" on page 88 for details on the external input signals INT0 and INT1).

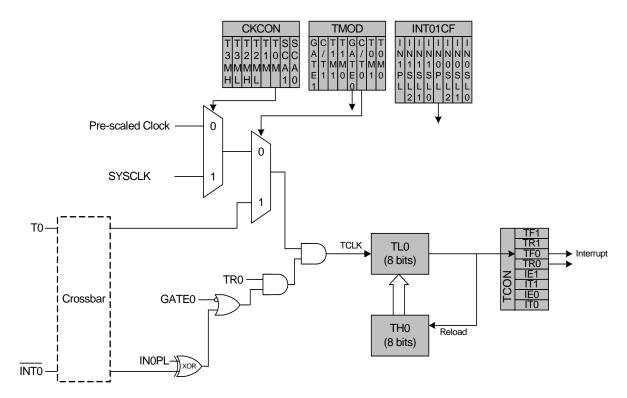


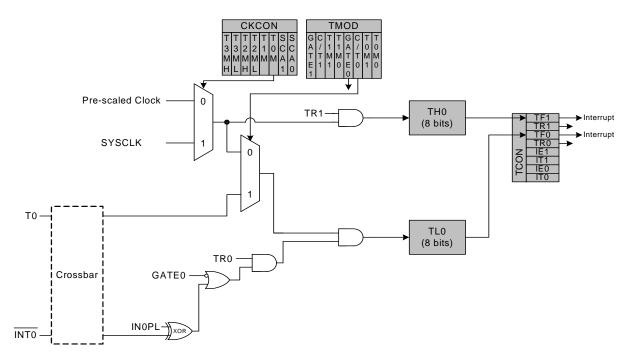
Figure 21.2. T0 Mode 2 Block Diagram



### 21.1.4. Mode 3: Two 8-bit Counter/Timers (Timer 0 Only)

In Mode 3, Timer 0 is configured as two separate 8-bit counter/timers held in TL0 and TH0. The counter/ timer in TL0 is controlled using the Timer 0 control/status bits in TCON and TMOD: TR0, C/T0, GATE0 and TF0. TL0 can use either the system clock or an external input signal as its timebase. The TH0 register is restricted to a timer function sourced by the system clock or prescaled clock. TH0 is enabled using the Timer 1 run control bit TR1. TH0 sets the Timer 1 overflow flag TF1 on overflow and thus controls the Timer 1 interrupt.

Timer 1 is inactive in Mode 3. When Timer 0 is operating in Mode 3, Timer 1 can be operated in Modes 0, 1 or 2, but cannot be clocked by external signals nor set the TF1 flag and generate an interrupt. However, the Timer 1 overflow can be used to generate baud rates for the SMBus and/or UART, and/or initiate ADC conversions. While Timer 0 is operating in Mode 3, Timer 1 run control is handled through its mode settings. To run Timer 1 while Timer 0 is in Mode 3, set the Timer 1 Mode as 0, 1, or 2. To disable Timer 1, configure it for Mode 3.







SFR Definition 21.1.	TCON: Time	r Control
----------------------	------------	-----------

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Valu
TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0	0000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Addres
						(bi	t addressable	e) 0x88
Bit7:	TF1: Timer 1	Overflow I	-lag.					
	Set by hardw	are when -	Timer 1 over	rflows. This	flag can be	e cleared by	/ software	but is auto-
	matically clea	ared when	the CPU ve	ctors to the	Timer 1 int	errupt serv	ice routine	
	0: No Timer 2							
	1: Timer 1 ha							
Bit6:	TR1: Timer 1		rol.					
	0: Timer 1 dis							
	1: Timer 1 en							
Bit5:	TF0: Timer 0		-	а. <b>т</b> .:	(I			
	Set by hardw							
	matically clea 0: No Timer (			ctors to the	Timer 0 Int	errupt serv	ice routine	•
	1: Timer 0 ha							
Bit4:	TR0: Timer 0							
JILT.	0: Timer 0 dis		01.					
	1: Timer 0 en							
Bit3:	IE1: External							
	This flag is se	•		n edge/leve	el of type de	fined by IT <sup>,</sup>	1 is detecte	ed. It can be
	cleared by so							
	rupt 1 service	e routine if	IT1 = 1. Wh	en IT1 = 0,	this flag is	set to '1' wl	nen INT1 i	s active as
	defined by bi	t IN1PL in	register INT	01CF (see	SFR Definit	tion 9.13).		
Bit2:	IT1: Interrupt							
	This bit selec							
	configured ad	ctive low or	high by the	IN1PL bit	in the IT010	CF register	(see SFR	Definition
	9.1 <u>3).</u>							
	0: <u>INT1</u> is lev							
2:44	1: INT1 is ed							
Bit1:	IE0: External			a adaa/layu		fined by IT(	) ia data at	d Itaan bu
	This flag is se							
	cleared by so rupt 0 service							
	defined by bi							s active as
BitO:	ITO: Interrupt			0101 (366	OF IX Definit			
5110.	This bit selec			red INT0 in	terrupt will b	ne edae or l	evel sensi	tive INT0 is
	configured ad							
	0: INTO is lev					- (		
	1: INT0 is ed							



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
GATE1	C/T1	T1M1	T1M0	GATE0	C/T0	T0M1	T0M0	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0x89
Bit7:	GATE1: Ti	mer 1 Gate	Control.					
			nen TR1 = 1 i	rrespective	of INT1 log	ic level.		
			ly when TR1				by bit IN1F	L in register
	INT01CF (	see SFR D	efinition 9.13)	).				
Bit6:		nter/Timer						
			ner 1 increme			•		
		Function:	Timer 1 increr	nented by h	igh-to-low	transitions o	on external	input pin
	(T1).	10 T						
Bits5–4:			Mode Select					
	mese bits	select the	Timer 1 opera	alion mode.				
	T1M1	T1M0		Mode				
	0	0		): 13-bit cou				
	0	1	Mode 1: 16-bit counter/timer					
	1	0	Mode 2:	8-bit counte		1		
	_	-		auto-reloa	-			
	1	1	Mode	3: Timer 1	inactive			
Bit3:		mer 0 Gate	Control					
Dito.			nen TR0 = 1 i	rrespective	of INTO log	ic level		
			ly when TR0				by bit IN0F	L in register
			efinition 9.13				-,	
Bit2:	C/T0: Cou	nter/Timer	Select.					
			ner 0 increme					
		Function:	Timer 0 increr	mented by h	igh-to-low	transitions o	on external	input pin
<b>D</b> 14 4 A	(T0).							
Bits1–0:			Mode Select					
	I nese bits	select the	Timer 0 opera	ation mode.				
	T0M1	T0M0		Mode				
	0	0	Mode 0	: 13-bit cou	nter/timer			
	0	1		: 16-bit cou				
	1	0	Mode 2:	8-bit counte		)		
		_		auto-reloa				
	1 1 Mode 3: Two 8-bit counter/timers							

## SFR Definition 21.2. TMOD: Timer Mode



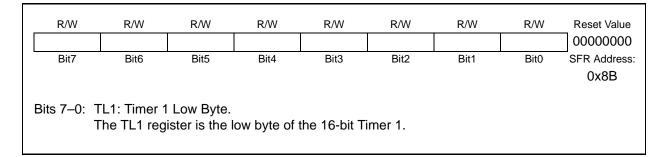
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value	
ТЗМН	T3ML	T2MH	T2ML	T1M	TOM	SCA1	SCA0	0000000	
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address 0x8E	
Bit7:	T3MH: Time	er 3 High By	yte Clock Se	elect.					
	This bit sele						is configur	ed in split	
	8-bit timer n		-		•				
	0: Timer 3 h				the T3XCL	.K bit in TM	R3CN.		
Bit6:	1: Timer 3 h T3ML: Time								
DILO.	This bit sele				lf Timer 3 is	configured	in split 8-h	hit timer	
	mode, this b					-	in spin o c		
	0: Timer 3 lo			•			R3CN.		
	1: Timer 3 lo	•		•					
Bit5:	T2MH: Time								
	This bit sele				• •		is configur	ed in split	
	8-bit timer n						DOON		
	0: Timer 2 h	• •			the I2XCL	.K bit in TM	R2CN.		
Bit4:	1: Timer 2 h	• •							
DIL4.	T2ML: Time This bit sele	•			lf Timer 2 is	configured	in split 8-h	vit timer	
	mode, this b						in spin o c		
	0: Timer 2 lo			•			2CN.		
	1: Timer 2 lo	•		•					
Bit3:	T1M: Timer	1 Clock Se	lect.						
						-	n C/T1 is s	et to logic 1.	
	0: Timer 1 u			by the prese	ale bits, SC	CA1-SCA0.			
Dire	1: Timer 1 u								
Bit2:	TOM: Timer			unaliad ta Ti		lia invariad			
	This bit sele logic 1.	cts the cloc	CK SOURCE SU	ipplied to TI	mer U. TUW	i is ignored	when C/T	) is set to	
	0: Counter/	Fimer () use	s the clock	defined by t	he prescale	hits SCA	I-SCA0		
	1: Counter/						00/10.		
Bits1–0:	SCA1-SCA								
	These bits o	control the c	division of th	e clock sup	plied to Tim	ner 0 and/or	Timer 1 if	configured	
	to use prescaled clock inputs.								
	SCA1	SCA0	Presc	aled Clock	(				
	0	0	•	ock divided	-				
	0	1	•	ock divided	•				
	1	0		ock divided					
	1 1 External clock divided by 8								
	Note: External clock divided by 8 is synchronized with the								
		mal clock div m clock.	ided by 8 is s	synchronized	with the				



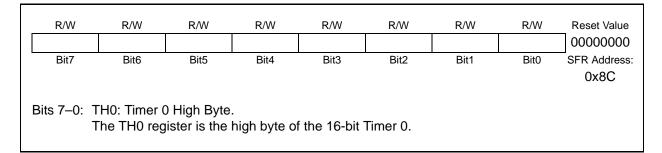
## SFR Definition 21.4. TL0: Timer 0 Low Byte

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
								0x8A
Bits 7–0: TL0: Timer 0 Low Byte. The TL0 register is the low byte of the 16-bit Timer 0.								

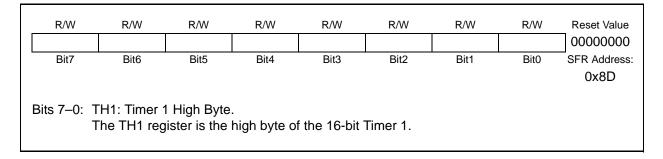
## SFR Definition 21.5. TL1: Timer 1 Low Byte



## SFR Definition 21.6. TH0: Timer 0 High Byte



## SFR Definition 21.7. TH1: Timer 1 High Byte





## 21.2. Timer 2

Timer 2 is a 16-bit timer formed by two 8-bit SFRs: TMR2L (low byte) and TMR2H (high byte). Timer 2 may operate in 16-bit auto-reload mode, (split) 8-bit auto-reload mode, USB Start-of-Frame (SOF) capture mode, or Low-Frequency Oscillator (LFO) Falling Edge capture mode. The Timer 2 operation mode is defined by the T2SPLIT (TMR2CN.3), T2CE (TMR2CN.4) bits, and T2CSS (TMR2CN.1) bits.

Timer 2 may be clocked by the system clock, the system clock divided by 12, or the external oscillator source divided by 8. The external clock mode is ideal for real-time clock (RTC) functionality, where the internal oscillator drives the system clock while Timer 2 (and/or the PCA) is clocked by an external precision oscillator. Note that the external oscillator source divided by 8 is synchronized with the system clock.

#### 21.2.1. 16-bit Timer with Auto-Reload

When T2SPLIT = '0' and T2CE = '0', Timer 2 operates as a 16-bit timer with auto-reload. Timer 2 can be clocked by SYSCLK, SYSCLK divided by 12, or the external oscillator clock source divided by 8. As the 16-bit timer register increments and overflows from 0xFFFF to 0x0000, the 16-bit value in the Timer 2 reload registers (TMR2RLH and TMR2RLL) is loaded into the Timer 2 register as shown in Figure 21.4, and the Timer 2 High Byte Overflow Flag (TMR2CN.7) is set. If Timer 2 interrupts are enabled, an interrupt will be generated on each Timer 2 overflow. Additionally, if Timer 2 interrupts are enabled and the TF2LEN bit is set (TMR2CN.5), an interrupt will be generated each time the lower 8 bits (TMR2L) overflow from 0xFF to 0x000.

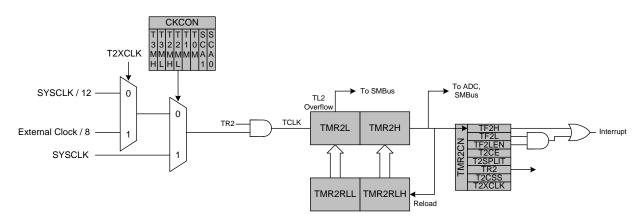


Figure 21.4. Timer 2 16-Bit Mode Block Diagram



### 21.2.2. 8-bit Timers with Auto-Reload

When T2SPLIT = '1' and T2CE = '0', Timer 2 operates as two 8-bit timers (TMR2H and TMR2L). Both 8-bit timers operate in auto-reload mode as shown in Figure 21.5. TMR2RLL holds the reload value for TMR2L; TMR2RLH holds the reload value for TMR2H. The TR2 bit in TMR2CN handles the run control for TMR2H. TMR2L is always running when configured for 8-bit Mode.

Each 8-bit timer may be configured to use SYSCLK, SYSCLK divided by 12, or the external oscillator clock source divided by 8. The Timer 2 Clock Select bits (T2MH and T2ML in CKCON) select either SYSCLK or the clock defined by the Timer 2 External Clock Select bit (T2XCLK in TMR2CN), as follows:

T2MH	T2XCLK	TMR2H Clock Source
0	0	SYSCLK / 12
0	1	External Clock / 8
1	Х	SYSCLK

T2ML	T2XCLK	TMR2L Clock Source
0	0	SYSCLK / 12
0	1	External Clock / 8
1	Х	SYSCLK

The TF2H bit is set when TMR2H overflows from 0xFF to 0x00; the TF2L bit is set when TMR2L overflows from 0xFF to 0x00. When Timer 2 interrupts are enabled, an interrupt is generated each time TMR2H overflows. If Timer 2 interrupts are enabled and TF2LEN (TMR2CN.5) is set, an interrupt is generated each time either TMR2L or TMR2H overflows. When TF2LEN is enabled, software must check the TF2H and TF2L flags to determine the source of the Timer 2 interrupt. The TF2H and TF2L interrupt flags are not cleared by hardware and must be manually cleared by software.

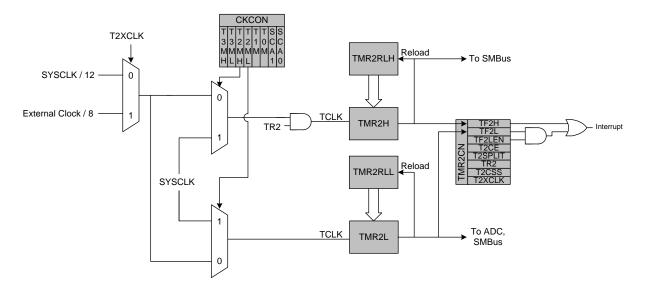


Figure 21.5. Timer 2 8-Bit Mode Block Diagram



### 21.2.3. Timer 2 Capture Modes: USB Start-of-Frame or LFO Falling Edge

When T2CE = '1', Timer 2 will operate in one of two special capture modes. The capture event can be selected between a USB Start-of-Frame (SOF) capture, and a Low-Frequency Oscillator (LFO) Falling Edge capture, using the T2CSS bit. The USB SOF capture mode can be used to calibrate the system clock or external oscillator against the known USB host SOF clock. The LFO falling-edge capture mode can be used to calibrate the internal Low-Frequency Oscillator against the internal High-Frequency Oscillator or an external clock source. When T2SPLIT = '0', Timer 2 counts up and overflows from 0xFFFF to 0x0000. Each time a capture event is received, the contents of the Timer 2 registers (TMR2H:TMR2L) are latched into the Timer 2 Reload registers (TMR2RLH:TMR2RLL). A Timer 2 interrupt is generated if enabled.

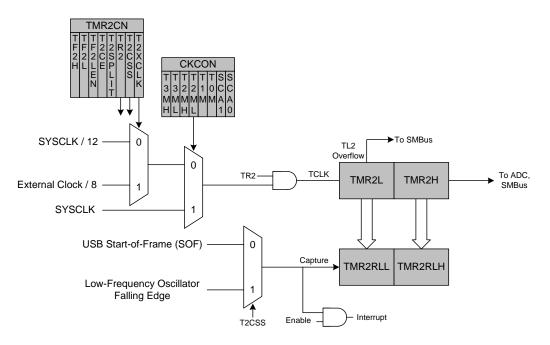


Figure 21.6. Timer 2 Capture Mode (T2SPLIT = '0')



When T2SPLIT = '1', the Timer 2 registers (TMR2H and TMR2L) act as two 8-bit counters. Each counter counts up independently and overflows from 0xFF to 0x00. Each time a capture event is received, the contents of the Timer 2 registers are latched into the Timer 2 Reload registers (TMR2RLH and TMR2RLL). A Timer 2 interrupt is generated if enabled.

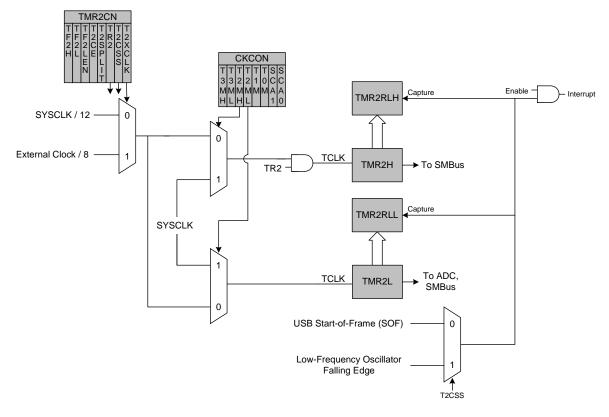


Figure 21.7. Timer 2 Capture Mode (T2SPLIT = '1')



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
TF2H	TF2L	TF2LEN	T2CE	T2SPLIT	TR2	T2CSS	T2XCLK	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
						(bit	t addressable)	) 0xC8
Bit7:		r 2 High Byte		-				
	•	ware when th						
		ur when Time						
		tting this bit of automaticall					•	
Bit6:		r 2 Low Byte			and must		by soliware	•
Dito:		ware when the		•	erflows fror	m 0xFF to 0	x00. When	this bit is
		rupt will be g						
	will set whe	n the low byt	e overflow	s regardless	of the Tim	er 2 mode.	This bit is r	not automat-
		d by hardwa						
Bit5:		mer 2 Low B		•				0.1.4
		bles/disables		•	•			
		abled, an int .ow Byte inte				e iow byte of		vernows.
		ow Byte inte						
Bit4:		r 2 Capture I	•					
		unction disal						
		unction enab						
		S. Each time						
	•	d TMR2L) ar					MR2RLH a	ind
Bit3:	,	, and a Timeı mer 2 Split M		-	ed (il enabl	ea).		
Dito.		it is set, Time			-bit timers v	with auto-rel	load.	
		perates in 16	•					
		perates as tv						
Bit2:		2 Run Contre						
		bles/disables			e, this bit e	nables/disal	bles TMR2	H only;
		lways enable	ed in this m	node.				
	0: Timer 2 d 1: Timer 2 e							
Bit1:		ner 2 Capture	Source S	Select				
Bitti		ects the source			hen bit T2	CE is set to	'1'.	
		source is US						
	•	source is falli			ency Oscill	ator.		
Bit0:		mer 2 Extern						
		ects the exter						
		external osci T2MH and T				•		
		ck and the sy				ามอ นออน เป		
		external clock				ided by 12.		
		xternal clock		•		•	Note that th	ne external
	oscillator so	urce divided	by 8 is sy	nchronized v	with the sys	stem clock.		

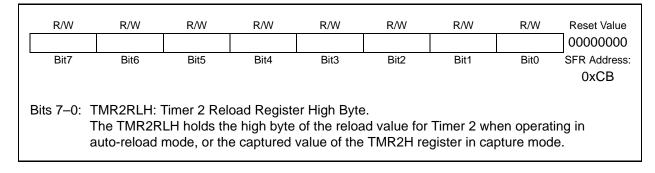
## SFR Definition 21.8. TMR2CN: Timer 2 Control



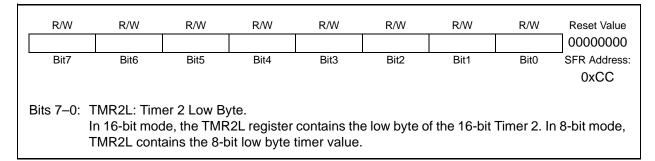
## SFR Definition 21.9. TMR2RLL: Timer 2 Reload Register Low Byte

	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
									0000000
	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
									0xCA
E	Bits 7–0: TMR2RLL: Timer 2 Reload Register Low Byte. TMR2RLL holds the low byte of the reload value for Timer 2 when operating in auto-reload mode, or the captured value of the TMR2L register in capture mode.								

## SFR Definition 21.10. TMR2RLH: Timer 2 Reload Register High Byte



## SFR Definition 21.11. TMR2L: Timer 2 Low Byte



## SFR Definition 21.12. TMR2H Timer 2 High Byte

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xCD
Bits 7–0: TMR2H: Timer 2 High Byte. In 16-bit mode, the TMR2H register contains the high byte of the 16-bit Timer 2. In 8-bit mode, TMR2H contains the 8-bit high byte timer value.								



## 21.3. Timer 3

Timer 3 is a 16-bit timer formed by two 8-bit SFRs: TMR3L (low byte) and TMR3H (high byte). Timer 3 may operate in 16-bit auto-reload mode, (split) 8-bit auto-reload mode, USB Start-of-Frame (SOF) capture mode, or Low-Frequency Oscillator (LFO) Rising Edge capture mode. The Timer 3 operation mode is defined by the T3SPLIT (TMR3CN.3), T3CE (TMR3CN.4) bits, and T3CSS (TMR3CN.1) bits.

Timer 3 may be clocked by the system clock, the system clock divided by 12, or the external oscillator source divided by 8. The external clock mode is ideal for real-time clock (RTC) functionality, where the internal oscillator drives the system clock while Timer 3 (and/or the PCA) is clocked by an external precision oscillator. Note that the external oscillator source divided by 8 is synchronized with the system clock.

#### 21.3.1. 16-bit Timer with Auto-Reload

When T3SPLIT (TMR3CN.3) is '0' and T3CE = '0', Timer 3 operates as a 16-bit timer with auto-reload. Timer 3 can be clocked by SYSCLK, SYSCLK divided by 12, or the external oscillator clock source divided by 8. As the 16-bit timer register increments and overflows from 0xFFFF to 0x0000, the 16-bit value in the Timer 3 reload registers (TMR3RLH and TM3RLL) is loaded into the Timer 3 register as shown in Figure 21.4, and the Timer 3 High Byte Overflow Flag (TMR3CN.7) is set. If Timer 3 interrupts are enabled, an interrupt will be generated on each Timer 3 overflow. Additionally, if Timer 3 interrupts are enabled and the TF3LEN bit is set (TMR3CN.5), an interrupt will be generated each time the lower 8 bits (TMR3L) overflow from 0xFF to 0x00.

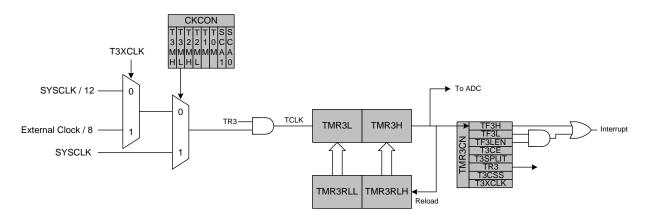


Figure 21.8. Timer 3 16-Bit Mode Block Diagram



### 21.3.2. 8-bit Timers with Auto-Reload

When T3SPLIT is '1' and T3CE = '0', Timer 3 operates as two 8-bit timers (TMR3H and TMR3L). Both 8-bit timers operate in auto-reload mode as shown in Figure 21.5. TMR3RLL holds the reload value for TMR3L; TMR3RLH holds the reload value for TMR3H. The TR3 bit in TMR3CN handles the run control for TMR3H. TMR3L is always running when configured for 8-bit Mode.

Each 8-bit timer may be configured to use SYSCLK, SYSCLK divided by 12, or the external oscillator clock source divided by 8. The Timer 3 Clock Select bits (T3MH and T3ML in CKCON) select either SYSCLK or the clock defined by the Timer 3 External Clock Select bit (T3XCLK in TMR3CN), as follows:

ТЗМН	T3XCLK	TMR3H Clock Source
0	0	SYSCLK / 12
0	1	External Clock / 8
1	Х	SYSCLK

T3ML	T3XCLK	TMR3L Clock Source
0	0	SYSCLK / 12
0	1	External Clock / 8
1	Х	SYSCLK

The TF3H bit is set when TMR3H overflows from 0xFF to 0x00; the TF3L bit is set when TMR3L overflows from 0xFF to 0x00. When Timer 3 interrupts are enabled, an interrupt is generated each time TMR3H overflows. If Timer 3 interrupts are enabled and TF3LEN (TMR3CN.5) is set, an interrupt is generated each time either TMR3L or TMR3H overflows. When TF3LEN is enabled, software must check the TF3H and TF3L flags to determine the source of the Timer 3 interrupt. The TF3H and TF3L interrupt flags are not cleared by hardware and must be manually cleared by software.

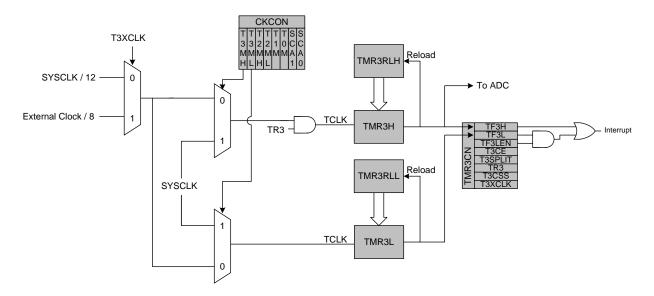


Figure 21.9. Timer 3 8-Bit Mode Block Diagram



### 21.3.3. USB Start-of-Frame Capture

When T3CE = '1', Timer 3 will operate in one of two special capture modes. The capture event can be selected between a USB Start-of-Frame (SOF) capture, and a Low-Frequency Oscillator (LFO) Rising Edge capture, using the T3CSS bit. The USB SOF capture mode can be used to calibrate the system clock or external oscillator against the known USB host SOF clock. The LFO rising-edge capture mode can be used to calibrate the internal Low-Frequency Oscillator against the internal High-Frequency Oscillator or an external clock source. When T3SPLIT = '0', Timer 3 counts up and overflows from 0xFFFF to 0x0000. Each time a capture event is received, the contents of the Timer 3 registers (TMR3H:TMR3L) are latched into the Timer 3 Reload registers (TMR3RLH:TMR3RLL). A Timer 3 interrupt is generated if enabled.

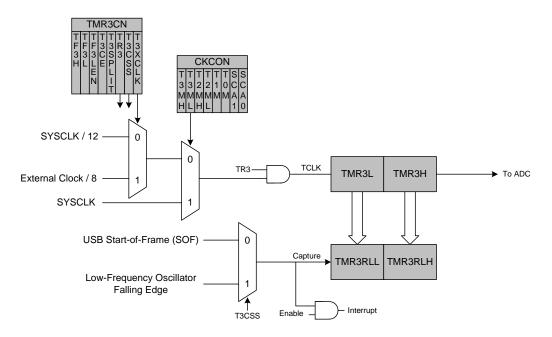


Figure 21.10. Timer 3 Capture Mode (T3SPLIT = '0')



When T3SPLIT = '1', the Timer 3 registers (TMR3H and TMR3L) act as two 8-bit counters. Each counter counts up independently and overflows from 0xFF to 0x00. Each time a capture event is received, the contents of the Timer 3 registers are latched into the Timer 3 Reload registers (TMR3RLH and TMR3RLL). A Timer 3 interrupt is generated if enabled.

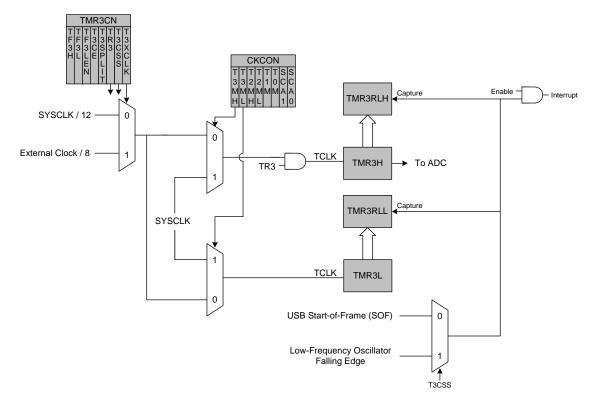


Figure 21.11. Timer 3 Capture Mode (T3SPLIT = '1')



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
TF3H	TF3L	TF3LEN	T3CE	T3SPLIT	TR3	T3CSS	T3XCLK	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
								0x91
Bit7:	TF3H: Time			-				
	Set by hard							
	this will occu							
	enabled, set	•					•	
Dite	TF3H is not				and must	be cleared i	by software	
Bit6:	TF3L: Timer Set by hard			•	orflowe from			thic hit ic
	set, an interi							
	will set wher							
	ically cleared	•		3 regardless		er o mode.		
Bit5:	TF3LEN: Tir			ot Enable.				
Bitor	This bit enal				errupts. If T	F3LEN is s	et and Time	er 3 inter-
	rupts are en							
	This bit shou							
	0: Timer 3 L			-				
	1: Timer 3 L	ow Byte inte	errupts ena	bled.				
Bit4:	T3CE: Time	r 3 Capture	Enable					
	0: Capture f							
	1: Capture f							
	by bit T3CS							
	(TMR3H and	,					MR3RLH a	nd
Dire	TMR3RLH),		•	-	ed (if enabl	ed).		
Bit3:	T3SPLIT: Ti	•					اممط	
	When this b		•			with auto-re	load.	
	0: Timer 3 o 1: Timer 3 o	•						
Bit2:	TR3: Timer 3	•			1013.			
DILZ.	This bit enal			In 8-bit mode	e this bit e	nables/disal	bles TMR3I	H only:
	TMR3L is al				, the bit of			r only,
	0: Timer 3 d							
	1: Timer 3 e							
Bit1:	T3CSS: Tim	er 3 Captur	e Source S	select.				
	This bit sele	cts the sour	ce of a cap	oture event w	hen bit T3/	CE is set to	'1'.	
	0: Capture s	ource is US	B SOF eve	ent.				
	1: Capture s				ency Oscilla	ator.		
Bit0:	T3XCLK: Tir							
	This bit sele							
	selects the e							
	Select bits (					i be used to	select betw	veen the
	external clos		•			ided by 10		
	0: Timer 3 e 1: Timer 3 e			•		•	Note that th	o ovtornol
	oscillator so					•		ie exterrial
	USCIIIAIUI SU		i by 0 15 Syl		viai ule sys			

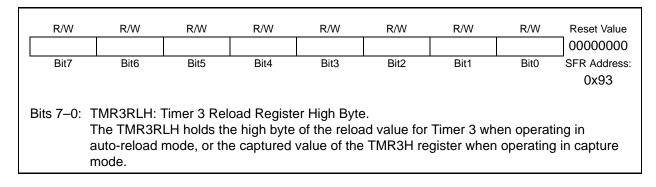
#### SFR Definition 21.13. TMR3CN: Timer 3 Control



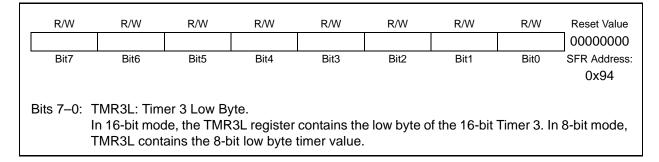
#### SFR Definition 21.14. TMR3RLL: Timer 3 Reload Register Low Byte

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value				
								00000000				
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0x92				
Bits 7–0:	Bits 7–0: TMR3RLL: Timer 3 Reload Register Low Byte. TMR3RLL holds the low byte of the reload value for Timer 3 when operating in auto-reload mode, or the captured value of the TMR3L register when operating in capture mode.											

#### SFR Definition 21.15. TMR3RLH: Timer 3 Reload Register High Byte



#### SFR Definition 21.16. TMR3L: Timer 3 Low Byte



#### SFR Definition 21.17. TMR3H Timer 3 High Byte

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0x95
	TMR3H: Tim In 16-bit moo mode, TMR3	de, the TMF	R3H register		• •	of the 16-b	bit Timer 3	3. In 8-bit



### 22. Programmable Counter Array (PCA0)

The Programmable Counter Array (PCA0) provides enhanced timer functionality while requiring less CPU intervention than the standard 8051 counter/timers. The PCA consists of a dedicated 16-bit counter/timer and five 16-bit capture/compare modules. Each capture/compare module has its own associated I/O line (CEXn) which is routed through the Crossbar to Port I/O when enabled (See Section "15.1. Priority Crossbar Decoder" on page 144 for details on configuring the Crossbar). The counter/timer is driven by a programmable timebase that can select between six sources: system clock, system clock divided by four, system clock signal on the ECI input pin. Each capture/compare module may be configured to operate independently in one of six modes: Edge-Triggered Capture, Software Timer, High-Speed Output, Frequency Output, 8-Bit PWM, or 16-Bit PWM (each mode is described in Section "22.2. Capture/Compare Modules" on page 257). The external oscillator clock option is ideal for real-time clock (RTC) functionality, allowing the PCA to be clocked by a precision external oscillator while the internal oscillator drives the system clock. The PCA is configured and controlled through the system controller's Special Function Registers. The PCA block diagram is shown in Figure 22.1

**Important Note:** The PCA Module 4 may be used as a watchdog timer (WDT), and is enabled in this mode following a system reset. Access to certain PCA registers is restricted while WDT mode is enabled. See **Section 22.3** for details.

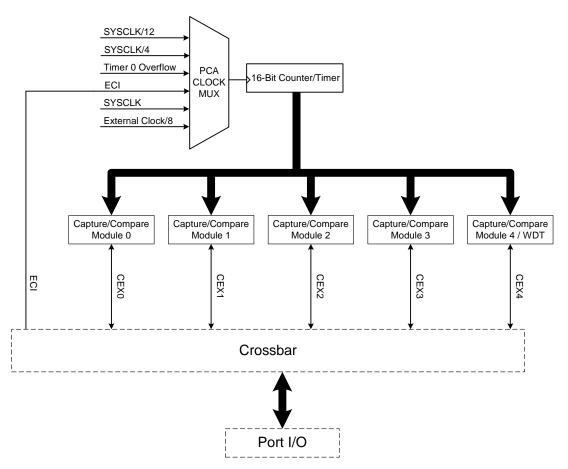


Figure 22.1. PCA Block Diagram



#### 22.1. PCA Counter/Timer

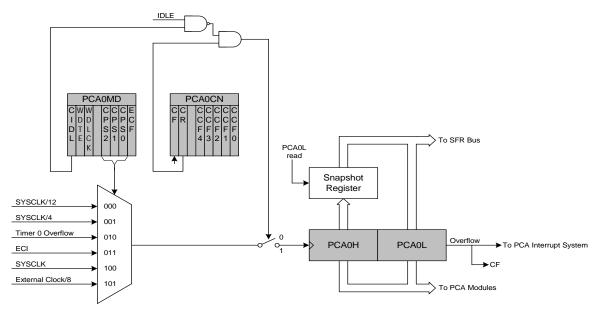
The 16-bit PCA counter/timer consists of two 8-bit SFRs: PCA0L and PCA0H. PCA0H is the high byte (MSB) of the 16-bit counter/timer and PCA0L is the low byte (LSB). Reading PCA0L automatically latches the value of PCA0H into a "snapshot" register; the following PCA0H read accesses this "snapshot" register. **Reading the PCA0L Register first guarantees an accurate reading of the entire 16-bit PCA0 counter.** Reading PCA0H or PCA0L does not disturb the counter operation. The CPS2-CPS0 bits in the PCA0MD register select the timebase for the counter/timer as shown in Table 22.1.

When the counter/timer overflows from 0xFFFF to 0x0000, the Counter Overflow Flag (CF) in PCA0MD is set to logic 1 and an interrupt request is generated if CF interrupts are enabled. Setting the ECF bit in PCA0MD to logic 1 enables the CF flag to generate an interrupt request. The CF bit is not automatically cleared by hardware when the CPU vectors to the interrupt service routine, and must be cleared by software (Note: PCA0 interrupts must be globally enabled before CF interrupts are recognized. PCA0 interrupts are globally enabled by setting the EA bit (IE.7) and the EPCA0 bit in EIE1 to logic 1). Clearing the CIDL bit in the PCA0MD register allows the PCA to continue normal operation while the CPU is in Idle mode.

CPS2	CPS1	CPS0	Timebase
0	0	0	System clock divided by 12
0	0	1	System clock divided by 4
0	1	0	Timer 0 overflow
0	1	1	High-to-low transitions on ECI (max rate = system clock divided by 4)
1	0	0	System clock
1	0	1	External oscillator source divided by 8*

Table 22.1. PCA Timebase Input Options	Table 22.1.	PCA	Timebase	Input	Options
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\*Note: External oscillator source divided by 8 is synchronized with the system clock.





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#### 22.2. Capture/Compare Modules

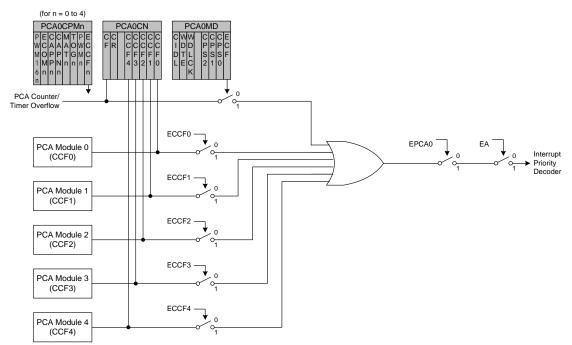
Each module can be configured to operate independently in one of six operation modes: Edge-triggered Capture, Software Timer, High Speed Output, Frequency Output, 8-Bit Pulse Width Modulator, or 16-Bit Pulse Width Modulator. Each module has Special Function Registers (SFRs) associated with it in the CIP-51 system controller. These registers are used to exchange data with a module and configure the module's mode of operation.

Table 22.2 summarizes the bit settings in the PCA0CPMn registers used to select the PCA capture/compare module's operating modes. Setting the ECCFn bit in a PCA0CPMn register enables the module's CCFn interrupt. Note: PCA0 interrupts must be globally enabled before individual CCFn interrupts are recognized. PCA0 interrupts are globally enabled by setting the EA bit and the EPCA0 bit to logic 1. See Figure 22.3 for details on the PCA interrupt configuration.

PWM16	ECOM	CAPP	CAPN	MAT	TOG	PWM	ECCF	Operation Mode
Х	Х	1	0	0	0	0	Х	Capture triggered by positive edge on CEXn
Х	Х	0	1	0	0	0	Х	Capture triggered by negative edge on CEXn
Х	Х	1	1	0	0	0	Х	Capture triggered by transition on CEXn
Х	1	0	0	1	0	0	Х	Software Timer
Х	1	0	0	1	1	0	Х	High Speed Output
Х	1	0	0	Х	1	1	Х	Frequency Output
0	1	0	0	Х	0	1	Х	8-Bit Pulse Width Modulator
1	1	0	0	Х	0	1	Х	16-Bit Pulse Width Modulator

#### Table 22.2. PCA0CPM Register Settings for PCA Capture/Compare Modules









#### 22.2.1. Edge-triggered Capture Mode

In this mode, a valid transition on the CEXn pin causes the PCA to capture the value of the PCA counter/ timer and load it into the corresponding module's 16-bit capture/compare register (PCA0CPLn and PCA0CPHn). The CAPPn and CAPNn bits in the PCA0CPMn register are used to select the type of transition that triggers the capture: low-to-high transition (positive edge), high-to-low transition (negative edge), or either transition (positive or negative edge). When a capture occurs, the Capture/Compare Flag (CCFn) in PCA0CN is set to logic 1 and an interrupt request is generated if CCF interrupts are enabled. The CCFn bit is not automatically cleared by hardware when the CPU vectors to the interrupt service routine, and must be cleared by software. If both CAPPn and CAPNn bits are set to logic 1, then the state of the Port pin associated with CEXn can be read directly to determine whether a rising-edge or falling-edge caused the capture.

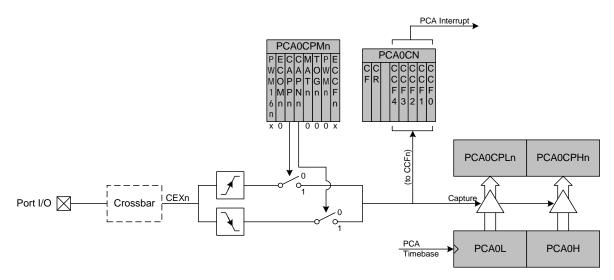


Figure 22.4. PCA Capture Mode Diagram

Note: The CEXn input signal must remain high or low for at least 2 system clock cycles to be recognized by the hardware.



#### 22.2.2. Software Timer (Compare) Mode

In Software Timer mode, the PCA counter/timer value is compared to the module's 16-bit capture/compare register (PCA0CPHn and PCA0CPLn). When a match occurs, the Capture/Compare Flag (CCFn) in PCA0CN is set to logic 1 and an interrupt request is generated if CCF interrupts are enabled. The CCFn bit is not automatically cleared by hardware when the CPU vectors to the interrupt service routine, and must be cleared by software. Setting the ECOMn and MATn bits in the PCA0CPMn register enables Software Timer mode.

**Important Note About Capture/Compare Registers**: When writing a 16-bit value to the PCA0 Capture/ Compare registers, the low byte should always be written first. Writing to PCA0CPLn clears the ECOMn bit to '0'; writing to PCA0CPHn sets ECOMn to '1'.

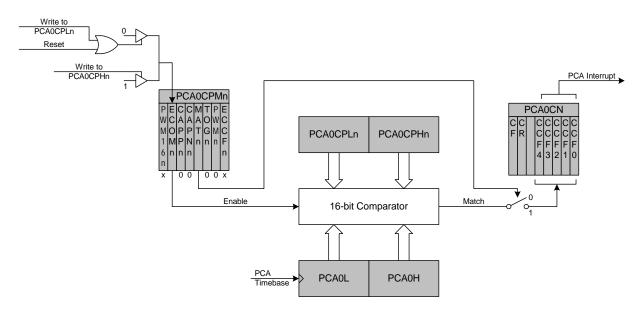


Figure 22.5. PCA Software Timer Mode Diagram



#### 22.2.3. High Speed Output Mode

In High Speed Output mode, a module's associated CEXn pin is toggled each time a match occurs between the PCA Counter and the module's 16-bit capture/compare register (PCA0CPHn and PCA0CPLn) Setting the TOGn, MATn, and ECOMn bits in the PCA0CPMn register enables the High-Speed Output mode.

**Important Note About Capture/Compare Registers**: When writing a 16-bit value to the PCA0 Capture/ Compare registers, the low byte should always be written first. Writing to PCA0CPLn clears the ECOMn bit to '0'; writing to PCA0CPHn sets ECOMn to '1'.

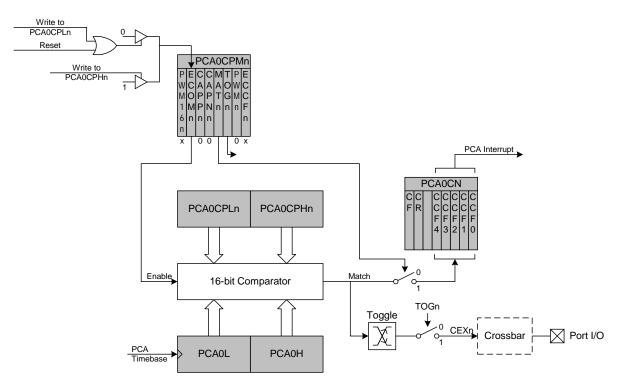


Figure 22.6. PCA High Speed Output Mode Diagram



#### 22.2.4. Frequency Output Mode

Frequency Output Mode produces a programmable-frequency square wave on the module's associated CEXn pin. The capture/compare module high byte holds the number of PCA clocks to count before the output is toggled. The frequency of the square wave is then defined by Equation 22.1.

$$F_{CEXn} = \frac{F_{PCA}}{2 \times PCA0CPHn}$$

Note: A value of 0x00 in the PCA0CPHn register is equal to 256 for this equation.

#### Equation 22.1. Square Wave Frequency Output

Where  $F_{PCA}$  is the frequency of the clock selected by the CPS2-0 bits in the PCA mode register, PCA0MD. The lower byte of the capture/compare module is compared to the PCA counter low byte; on a match, CEXn is toggled and the offset held in the high byte is added to the matched value in PCA0CPLn. Frequency Output Mode is enabled by setting the ECOMn, TOGn, and PWMn bits in the PCA0CPMn register.

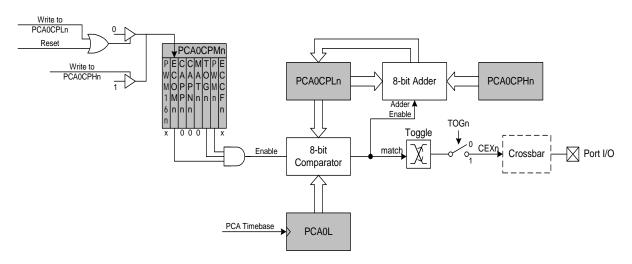


Figure 22.7. PCA Frequency Output Mode



#### 22.2.5. 8-Bit Pulse Width Modulator Mode

Each module can be used independently to generate a pulse width modulated (PWM) output on its associated CEXn pin. The frequency of the output is dependent on the timebase for the PCA counter/timer. The duty cycle of the PWM output signal is varied using the module's PCA0CPLn capture/compare register. When the value in the low byte of the PCA counter/timer (PCA0L) is equal to the value in PCA0CPLn, the output on the CEXn pin will be set. When the count value in PCA0L overflows, the CEXn output will be reset (see Figure 22.8). Also, when the counter/timer low byte (PCA0L) overflows from 0xFF to 0x00, PCA0CPLn is reloaded automatically with the value stored in the module's capture/compare high byte (PCA0CPHn) without software intervention. Setting the ECOMn and PWMn bits in the PCA0CPMn register enables 8-Bit Pulse Width Modulator mode. The duty cycle for 8-Bit PWM Mode is given by Equation 22.2.

**Important Note About Capture/Compare Registers**: When writing a 16-bit value to the PCA0 Capture/ Compare registers, the low byte should always be written first. Writing to PCA0CPLn clears the ECOMn bit to '0'; writing to PCA0CPHn sets ECOMn to '1'.

$$DutyCycle = \frac{(256 - PCA0CPHn)}{256}$$

#### Equation 22.2. 8-Bit PWM Duty Cycle

Using Equation 22.2, the largest duty cycle is 100% (PCA0CPHn = 0), and the smallest duty cycle is 0.39% (PCA0CPHn = 0xFF). A 0% duty cycle may be generated by clearing the ECOMn bit to '0'.

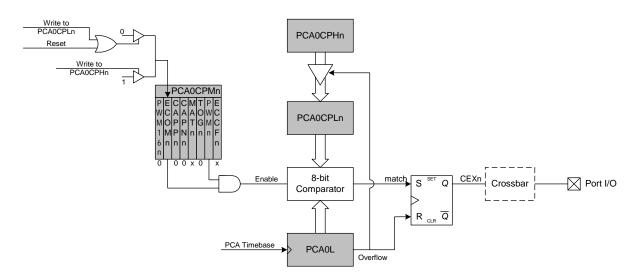


Figure 22.8. PCA 8-Bit PWM Mode Diagram



#### 22.2.6. 16-Bit Pulse Width Modulator Mode

A PCA module may also be operated in 16-Bit PWM mode. In this mode, the 16-bit capture/compare module defines the number of PCA clocks for the low time of the PWM signal. When the PCA counter matches the module contents, the output on CEXn is asserted high; when the counter overflows, CEXn is asserted low. To output a varying duty cycle, new value writes should be synchronized with PCA CCFn match interrupts. 16-Bit PWM Mode is enabled by setting the ECOMn, PWMn, and PWM16n bits in the PCA0CPMn register. For a varying duty cycle, match interrupts should be enabled (ECCFn = 1 AND MATn = 1) to help synchronize the capture/compare register writes. The duty cycle for 16-Bit PWM Mode is given by Equation 22.3.

**Important Note About Capture/Compare Registers**: When writing a 16-bit value to the PCA0 Capture/ Compare registers, the low byte should always be written first. Writing to PCA0CPLn clears the ECOMn bit to '0'; writing to PCA0CPHn sets ECOMn to '1'.

$$DutyCycle = \frac{(65536 - PCA0CPn)}{65536}$$

#### Equation 22.3. 16-Bit PWM Duty Cycle

Using Equation 22.3, the largest duty cycle is 100% (PCA0CPn = 0), and the smallest duty cycle is 0.0015% (PCA0CPn = 0xFFFF). A 0% duty cycle may be generated by clearing the ECOMn bit to '0'.

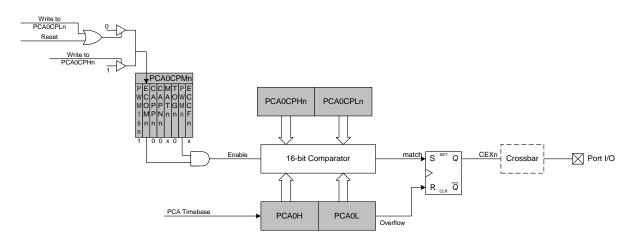


Figure 22.9. PCA 16-Bit PWM Mode



#### 22.3. Watchdog Timer Mode

A programmable watchdog timer (WDT) function is available through the PCA Module 4. The WDT is used to generate a reset if the time between writes to the WDT update register (PCA0CPH4) exceed a specified limit. The WDT can be configured and enabled/disabled as needed by software.

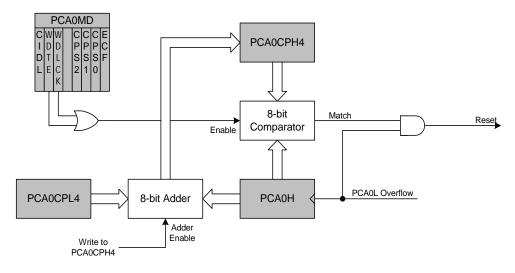
With the WDTE and/or WDLCK bits set to '1' in the PCA0MD register, Module 4 operates as a watchdog timer (WDT). The Module 4 high byte is compared to the PCA counter high byte; the Module 4 low byte holds the offset to be used when WDT updates are performed. **The Watchdog Timer is enabled on reset. Writes to some PCA registers are restricted while the Watchdog Timer is enabled.** 

#### 22.3.1. Watchdog Timer Operation

While the WDT is enabled:

- PCA counter is forced on.
- Writes to PCA0L and PCA0H are not allowed.
- PCA clock source bits (CPS2-CPS0) are frozen.
- PCA Idle control bit (CIDL) is frozen.
- Module 4 is forced into Watchdog Timer mode.
- Writes to the Module 4 mode register (PCA0CPM4) are disabled.

While the WDT is enabled, writes to the CR bit will not change the PCA counter state; the counter will run until the WDT is disabled. The PCA counter run control (CR) will read zero if the WDT is enabled but user software has not enabled the PCA counter. If a match occurs between PCA0CPH4 and PCA0H while the WDT is enabled, a reset will be generated. To prevent a WDT reset, the WDT may be updated with a write of any value to PCA0CPH4. Upon a PCA0CPH4 write, PCA0H plus the offset held in PCA0CPL4 is loaded into PCA0CPH4 (See Figure 22.10).



#### Figure 22.10. PCA Module 4 with Watchdog Timer Enabled

Note that the 8-bit offset held in PCA0CPH4 is compared to the upper byte of the 16-bit PCA counter. This offset value is the number of PCA0L overflows before a reset. Up to 256 PCA clocks may pass before the first PCA0L overflow occurs, depending on the value of the PCA0L when the update is performed. The total offset is then given (in PCA clocks) by Equation 22.4, where PCA0L is the value of the PCA0L register at the time of the update.



#### $Offset = (256 \times PCA0CPL4) + (256 - PCA0L)$

#### Equation 22.4. Watchdog Timer Offset in PCA Clocks

The WDT reset is generated when PCA0L overflows while there is a match between PCA0CPH4 and PCA0H. Software may force a WDT reset by writing a '1' to the CCF4 flag (PCA0CN.4) while the WDT is enabled.

#### 22.3.2. Watchdog Timer Usage

To configure the WDT, perform the following tasks:

- 1. Disable the WDT by writing a '0' to the WDTE bit.
- 2. Select the desired PCA clock source (with the CPS2-CPS0 bits).
- 3. Load PCA0CPL4 with the desired WDT update offset value.
- 4. Configure the PCA Idle mode (set CIDL if the WDT should be suspended while the CPU is in Idle mode).
- 5. Enable the WDT by setting the WDTE bit to '1'.
- 6. (optional) Lock the WDT (prevent WDT disable until the next system reset) by setting the WDLCK bit to '1'.
- 7. Write a value to PCA0CPH4 to reload the WDT.

The PCA clock source and Idle mode select cannot be changed while the WDT is enabled. The watchdog timer is enabled by setting the WDTE or WDLCK bits in the PCA0MD register. When WDLCK is set, the WDT cannot be disabled until the next system reset. If WDLCK is not set, the WDT is disabled by clearing the WDTE bit.

The WDT is enabled following any reset. The PCA0 counter clock defaults to the system clock divided by 12, PCA0L defaults to 0x00, and PCA0CPL4 defaults to 0x00. Using Equation 22.4, this results in a WDT timeout interval of 256 PCA clocks. Table 22.3 lists some example timeout intervals for typical system clocks.

System Clock (Hz)	PCA0CPL4	Timeout Interval (ms)
12,000,000	255	65.5
12,000,000	128	33.0
12,000,000	32	8.4
24,000,000	255	32.8
24,000,000	128	16.5
24,000,000	32	4.2
1,500,000 <sup>2</sup>	255	524.3
1,500,000 <sup>2</sup>	128	264.2
1,500,000 <sup>2</sup>	32	67.6
32,768	255	24,000
32,768	128	12,093.75
32,768	32	3,093.75

#### Table 22.3. Watchdog Timer Timeout Intervals<sup>1</sup>

Notes:

- 1. Assumes SYSCLK / 12 as the PCA clock source, and a PCA0L
- value of 0x00 at the update time.
- **2.** System Clock reset frequency.



#### 22.4. Register Descriptions for PCA

Following are detailed descriptions of the special function registers related to the operation of the PCA.

#### SFR Definition 22.1. PCA0CN: PCA Control

~ -	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
CF	CR	-	CCF4	CCF3	CCF2	CCF1	CCF0	0000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Addres
						(bit	addressable	) 0xD8
Bit7:	CF: PCA Co	unter/Time	r Overflow F	lag.				
	Set by hardv	vare when	he PCA Cou	unter/Timer	overflows f	rom 0xFFFI	to 0x000	0. When the
	Counter/Tim	er Overflov	(CF) interr	upt is enabl	ed, setting	this bit caus	ses the CP	U to vector
	to the PCA in	nterrupt se	vice routine	. This bit is	not automa	tically clear	ed by harc	lware and
	must be clea	ared by soft	ware.					
Bit6:	CR: PCA Co	ounter/Time	r Run Contr	ol.				
	This bit enab	oles/disable	es the PCA (	Counter/Tim	ner.			
	0: PCA Cour							
	1: PCA Cour							
Bit5:	UNUSED. R	,						
Bit4:	CCF4: PCA							
	This bit is se				•			•
	enabled, set	-				•		outine. Thi
	bit is not aut				d must be d	cleared by s	software.	
Bit3:	CCF3: PCA		•			a M/baatb	а ООГО ind	annunt in
	This bit is se							
	enabled, set bit is not aut	•						
	CCF2: PCA					leared by s	onware.	
Dit7.	001 Z. F 0A							
Bit2:	This hit is so	t hy hardw	aro whon a i	match or ca	nture occur	e Whan th	o CCE2 int	orrunt is
Bit2:	This bit is se							
Bit2:	enabled, set	ting this bit	causes the	CPU to vec	tor to the P	CA interrup	ot service re	
	enabled, set bit is not aut	ting this bit omatically	causes the cleared by h	CPU to veo ardware an	tor to the P	CA interrup	ot service re	
Bit2: Bit1:	enabled, set bit is not aut CCF1: PCA	ting this bit omatically Module 1 (	causes the cleared by h Capture/Con	CPU to veo ardware an npare Flag.	tor to the P d must be c	CA interrup cleared by s	ot service re software.	outine. Thi
	enabled, set bit is not auto CCF1: PCA This bit is se	ting this bit omatically Module 1 ( t by hardw	causes the cleared by h Capture/Con are when a	CPU to veo ardware an npare Flag. match or ca	tor to the P d must be c	CA interrup cleared by s	ot service re oftware. e CCF1 int	outine. This errupt is
	enabled, set bit is not auto CCF1: PCA This bit is se enabled, set	ting this bit omatically Module 1 ( t by hardw ting this bit	causes the cleared by h Capture/Con are when a causes the	CPU to veo ardware an npare Flag. match or ca CPU to veo	tor to the P d must be c pture occur tor to the P	CA interrup cleared by s s. When th CA interrup	ot service re coftware. e CCF1 int ot service re	outine. This errupt is
	enabled, set bit is not auto CCF1: PCA This bit is se enabled, set bit is not auto	ting this bit omatically Module 1 ( t by hardw ting this bit omatically	causes the cleared by h Capture/Con are when a causes the cleared by h	CPU to veo ardware an pare Flag. match or ca CPU to veo ardware an	tor to the P d must be c pture occur tor to the P	CA interrup cleared by s s. When th CA interrup	ot service re coftware. e CCF1 int ot service re	outine. This errupt is
Bit1:	enabled, set bit is not auto CCF1: PCA This bit is se enabled, set bit is not auto CCF0: PCA	ting this bit omatically Module 1 ( t by hardw ting this bit omatically Module 0 (	causes the cleared by h Capture/Con are when a causes the cleared by h Capture/Con	CPU to veo ardware an pare Flag. match or ca CPU to veo ardware an pare Flag.	tor to the P d must be c pture occur tor to the P d must be c	CA interrup cleared by s rs. When th CA interrup cleared by s	ot service re coftware. e CCF1 int ot service re coftware.	outine. This errupt is outine. This
Bit1:	enabled, set bit is not auto CCF1: PCA This bit is se enabled, set bit is not auto	ting this bit omatically Module 1 ( t by hardw ting this bit omatically Module 0 ( t by hardw	causes the cleared by h Capture/Con are when a causes the cleared by h Capture/Con are when a	CPU to veo ardware an pare Flag. match or ca CPU to veo ardware an pare Flag. match or ca	tor to the P d must be c pture occur tor to the P d must be c pture occur	CA interrup cleared by s rs. When th CA interrup cleared by s rs. When th	ot service re coftware. e CCF1 int ot service re coftware. e CCF0 int	errupt is outine. This outine. This outine. This errupt is



#### SFR Definition 22.2. PCA0MD: PCA Mode

R/W	R/W	R/W	R/V	V R/W	R/W	R/W	R/W	Reset Value			
CIDL	WDTE	WDLC	K -	CPS2	CPS1	CPS0	ECF	0100000			
Bit7	Bit6	Bit5	Bit	4 Bit3	Bit2	Bit1	Bit0	SFR Address 0xD9			
Bit7:	CIDL: PCA				lada						
	0: PCA con	tinues to f	function n	CPU is in Idle M ormally while th	e system co			le.			
				d while the syste	em controlle	er is in Idle N	lode.				
Bit6:	WDTE: Wa	-									
				is used as the v	watchdog tii	mer.					
	0: Watchdo	-		<del></del>							
D:46.				Vatchdog Timer							
Bit5:	WDLCK: W				r \//han \//		to (1, the	Matchdog			
				Watchdog Time		JLCK is set	to i, the	watchuog			
	•			il the next syste	m reset.						
	0: Watchdo 1: Watchdo	-		dlockod							
Bit4:		-									
Bits3–1:	UNUSED. Read = 0b, Write = don't care. CPS2-CPS0: PCA Counter/Timer Pulse Select										
DII30-1.	: CPS2–CPS0: PCA Counter/Timer Pulse Select. These bits select the timebase source for the PCA counter.										
	These bits :	select the	umebase	Source for the		÷I.					
	CPS2	CPS1	CPS0	-		mebase					
	0	0	0	System clock d	•	2					
	0	0	1	System clock d							
	0	1	0	Timer 0 overflov							
	0	1	1	High-to-low trar divided by 4)	nsitions on I	ECI (max rat	e = syste	m clock			
		^	0	System clock							
	1	0									
	1	0	1	External clock of	livided by 8	*					
		-	1 0	External clock of Reserved	livided by 8	*					
	1	0			livided by 8	*					
	1 1 1	0 1 1	0 1	Reserved Reserved	-		em clock.				
	1 1 1	0 1 1	0 1	Reserved	-		m clock.				
Bit0:	1 1 1 *Note: Exte	0 1 1 ernal oscilla	0 1 ator source	Reserved Reserved divided by 8 is sy	/nchronized		em clock.				
Bit0:	1 1 *Note: Exter ECF: PCA	0 1 1 ernal oscilla	0 1 ator source ïmer Ove	Reserved Reserved divided by 8 is sy rflow Interrupt E	vnchronized	with the syste					
Bit0:	1 1 *Note: Exte ECF: PCA ( This bit sets	0 1 2 ernal oscilla Counter/T s the mas	0 1 ator source Timer Ove king of the	Reserved Reserved divided by 8 is sy	vnchronized	with the syste					
Bit0:	1 1 *Note: Exter ECF: PCA ( This bit sets 0: Disable t	0 1 2 ernal oscilla Counter/T s the mas he CF inte	0 1 ator source ïmer Ove king of the errupt.	Reserved Reserved divided by 8 is sy rflow Interrupt E PCA Counter/	ynchronized nable. Timer Overl	with the syste flow (CF) int	errupt.	.7) is set.			
3it0:	1 1 *Note: Exter ECF: PCA ( This bit sets 0: Disable t	0 1 2 ernal oscilla Counter/T s the mas he CF inte	0 1 ator source ïmer Ove king of the errupt.	Reserved Reserved divided by 8 is sy rflow Interrupt E	ynchronized nable. Timer Overl	with the syste flow (CF) int	errupt.	.7) is set.			
3it0:	1 1 *Note: Exter ECF: PCA ( This bit sets 0: Disable t	0 1 2 ernal oscilla Counter/T s the mas he CF inte	0 1 ator source ïmer Ove king of the errupt.	Reserved Reserved divided by 8 is sy rflow Interrupt E PCA Counter/	ynchronized nable. Timer Overl	with the syste flow (CF) int	errupt.	.7) is set.			
	1 1 *Note: External ECF: PCA ( This bit sets 0: Disable t 1: Enable a	0 1 2 counter/T s the mas he CF inte PCA Cou	0 1 ator source imer Ove king of the errupt. unter/Time	Reserved Reserved divided by 8 is sy rflow Interrupt E PCA Counter/	ynchronized nable. Timer Overl	with the syste flow (CF) int st when CF (	errupt. (PCA0CN				
	1 1 *Note: External ECF: PCA ( This bit sets 0: Disable t 1: Enable a	0 1 2 counter/T s the mas he CF inte PCA Cou	0 1 ator source imer Ove king of the errupt. unter/Time set to '1',	Reserved Reserved divided by 8 is sy rflow Interrupt E e PCA Counter/ er Overflow inter	ynchronized nable. Timer Overf rrupt reques	with the syste flow (CF) int st when CF (	errupt. (PCA0CN <b>dified. To</b>	change the			



R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
PWM16	n ECOMn	CAPPn	CAPNn	MATn	TOGn	PWMn	ECCFn	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
								0xDA, 0xDB, 0xDC, 0xDD, 0xDE
PCA0CF	Mn Address:				PCA0CPM1			
					PCA0CPM3	= 0xDD (n	= 3),	
D:/-			PM4 = 0xD	```				
Bit7:	PWM16n: 16					mada ia ar	oblad (D)A	(Ma 1)
	This bit sele 0: 8-bit PWN		iode when i	uise widtr	wodulation	n mode is er	labled (PVV	$\operatorname{NVIN} = 1$ ).
	1: 16-bit PW							
Bit6:	ECOMn: Co			hle				
Bito.	This bit enal	•			ion for PCA	module n.		
	0: Disabled.							
	1: Enabled.							
Bit5:	CAPPn: Cap	oture Positiv	e Function	Enable.				
	This bit enat	oles/disable	s the positi	ve edge ca	oture for PC	A module n	I <b>.</b>	
	0: Disabled.							
D'14	1: Enabled.							
Bit4:	CAPNn: Cap	•			nture for D			
	This bit enat 0: Disabled.	Jies/disable	s the negat	ive edge ca	aplure for PC	SA module	n.	
	1: Enabled.							
Bit3:	MATn: Matcl	h Function I	Enable.					
	This bit enab			n function fo	or PCA mod	ule n. Wher	n enabled, i	matches of
	the PCA cou						-	
	register to be	e set to logi	c 1.					
	0: Disabled.							
	1: Enabled.							
Bit2:	TOGn: Togg			function fo		ula a M/har		matakan af
	This bit enab the PCA cou							
	CEXn pin to							
	Output Mode						op 01 01 00 0 11	
	0: Disabled.							
	1: Enabled.							
Bit1:	PWMn: Puls							
	This bit enab							
	modulated s mode is use							
	Frequency C			Sgic T. II the		s also set, t	ne module	operates in
	0: Disabled.							
	1: Enabled.							
Bit0:	ECCFn: Cap	oture/Comp	are Flag Int	errupt Enat	ole.			
	This bit sets		-	pture/Comp	oare Flag (C	CFn) interru	upt.	
	0: Disable C		•			00 <b>-</b> -		
	1: Enable a	Capture/Co	mpare Flag	interrupt re	equest wher	n CCFn is s	et.	

#### SFR Definition 22.3. PCA0CPMn: PCA Capture/Compare Mode



#### SFR Definition 22.4. PCA0L: PCA Counter/Timer Low Byte

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xF9
	PCA0L: PCA The PCA0L r				f the 16-bit	PCA Count	er/Timer.	

#### SFR Definition 22.5. PCA0H: PCA Counter/Timer High Byte

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xFA
	PCA0H: PCA The PCA0H		•		) of the 16-t	bit PCA Cou	unter/Time	er.

#### SFR Definition 22.6. PCA0CPLn: PCA Capture Module Low Byte

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value		
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xFB, 0xE9, 0xEB, 0xED, 0xFD		
PCA0CPLn Address:PCA0CPL0 = $0xFB (n = 0)$ , PCA0CPL1 = $0xE9 (n = 1)$ , PCA0CPL2 = $0xEB (n = 2)$ , PCA0CPL3 = $0xED (n = 3)$ , PCA0CPL4 = $0xFD (n = 4)$										
Bits7–0: PCA0CPLn: PCA Capture Module Low Byte. The PCA0CPLn register holds the low byte (LSB) of the 16-bit capture module n.										



#### SFR Definition 22.7. PCA0CPHn: PCA Capture Module High Byte

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value 00000000	
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:	
								0xFC, 0xEA, 0xEC,0xEE, 0xFE	
PCA0CPHn Address:PCA0CPH0 = $0xFC (n = 0)$ , PCA0CPH1 = $0xEA (n = 1)$ , PCA0CPH2 = $0xEC (n = 2)$ , PCA0CPH3 = $0xEE (n = 3)$ , PCA0CPH4 = $0xFE (n = 4)$									
Bits7–0: PCA0CPHn: PCA Capture Module High Byte. The PCA0CPHn register holds the high byte (MSB) of the 16-bit capture module n.									

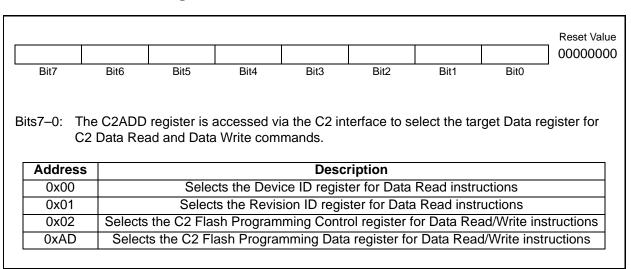


### 23. C2 Interface

C8051F34x devices include an on-chip Silicon Labs 2-Wire (C2) debug interface to allow Flash programming and in-system debugging with the production part installed in the end application. The C2 interface uses a clock signal (C2CK) and a bi-directional C2 data signal (C2D) to transfer information between the device and a host system. See the C2 Interface Specification for details on the C2 protocol.

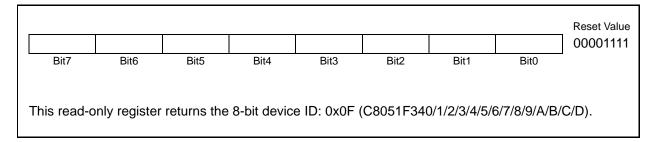
#### 23.1. C2 Interface Registers

The following describes the C2 registers necessary to perform Flash programming functions through the C2 interface. All C2 registers are accessed through the C2 interface as described in the C2 Interface Specification.



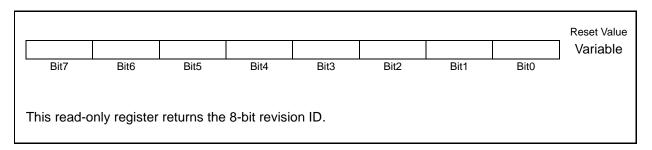
#### C2 Register Definition 23.1. C2ADD: C2 Address

#### C2 Register Definition 23.2. DEVICEID: C2 Device ID

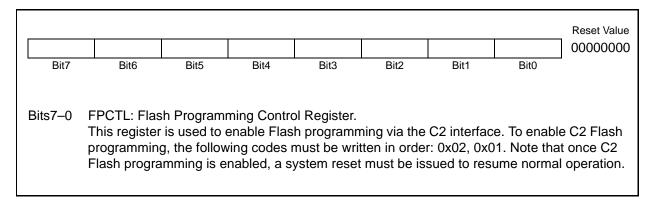




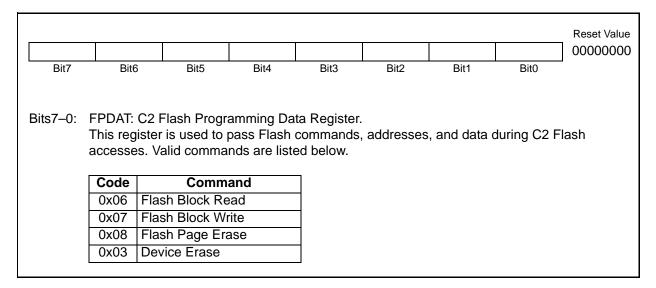
#### C2 Register Definition 23.3. REVID: C2 Revision ID



### C2 Register Definition 23.4. FPCTL: C2 Flash Programming Control



#### C2 Register Definition 23.5. FPDAT: C2 Flash Programming Data





#### 23.2. C2 Pin Sharing

The C2 protocol allows the C2 pins to be shared with user functions so that in-system debugging and Flash programming functions may be performed. This is possible because C2 communication is typically performed when the device is in the halt state, where all on-chip peripherals and user software are stalled. In this halted state, the C2 interface can safely 'borrow' the C2CK (RST) and C2D (P3.0) pins. Note that the C2D pin is shared on the 32-pin packages only (C8051F342/3/6/7/9/A/B). In most applications, external resistors are required to isolate C2 interface traffic from the user application. A typical isolation configuration is shown in Figure 23.1.

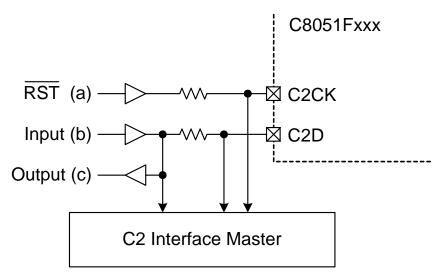


Figure 23.1. Typical C2 Pin Sharing

The configuration in Figure 23.1 assumes the following:

- 1. The <u>user input</u> (b) cannot change state while the target device is halted.
- 2. The  $\overline{RST}$  pin on the target device is used as an input only.

Additional resistors may be necessary depending on the specific application.



### DOCUMENT CHANGE LIST

#### **Revision 0.5 to Revision 1.0**

- Updated Table 3.1, "Global DC Electrical Characteristics," on page 25.
- Updated Table 5.1, "ADC0 Electrical Characteristics," on page 56.
- Various small text changes.
- Updated Table 8.1, "Voltage Regulator Electrical Specifications," on page 69.
- Updated Flash security behavior.

#### Revision 1.0 to Revision 1.1

- Added two new part numbers C8051F348/9 and made associated changes.
- Corrected the entries "24 kHz" and "48 kHz" to "24 MHz" and "48 MHz" in the "Conditions" column of Table 3.1, "Global DC Electrical Characteristics," on page 38.
- Added note to configure external interrupt pin as open-drain with a "1" in the port latch in Section 9.3.2. "External Interrupts" on page 96.
- Various small text changes.
- Updated the figures in Section 15.1. "Priority Crossbar Decoder" and added a new figure to clarify crossbar capabilities.
- Corrected the description of the UNDRUN bit in USB Register Definition 16.19. "EINCSRL: USB0 IN Endpoint Control Low Byte" on page 198 to clarify that this bit works only in Isochronous Mode.
- Corrected the maximum SMBus speed from 1/10th to 1/20th of the system clock in Section 17. "SMBus" on page 205.
- Corrected the descriptions for the following states and the corresponding typical response options in Table 17.4. "SMBus Status Decoding" on page 221:
  - Slave Transmitter (Status Vector: 0101)
  - Slave Receiver (Status Vector: 0001)
- Corrected the bit location of MSTEN from SPI0CN.6 to SPI0CFG.6 in Section 20.2. "SPI0 Master Operation" on page 243.
- Corrected the description of the WCOL bit in SFR Definition 20.2. "SPI0CN: SPI0 Control" on page 249 to match the description in Section 20.4. "SPI0 Interrupt Sources" on page 245.
- Clarified the following parameters in Table 8.1, "Voltage Regulator Electrical Specifications," on page 69:
  - VBUS Detection Input High and Low Voltages
  - Dropout Voltage
- Updated the package drawings with additional dimensions in Figure 4.2 and Table 4.2, "TQFP-48 Package Dimensions," on page 32, and Figure 4.4 and Table 4.4, "LQFP-32 Package Dimensions," on page 35.

#### **Revision 1.1 to Revision 1.2**

- Added two new part numbers C8051F34A/B and made associated changes.
- Corrected references to locations of T0M and T1M in the SFR definition of TMOD on page 240.

Rev. 1.3

• Corrected instances of "8k" to "4k" in the SFR definition of EMI0CF on page 118.

#### **Revision 1.2 to Revision 1.3**

• Added QFN-32 package.

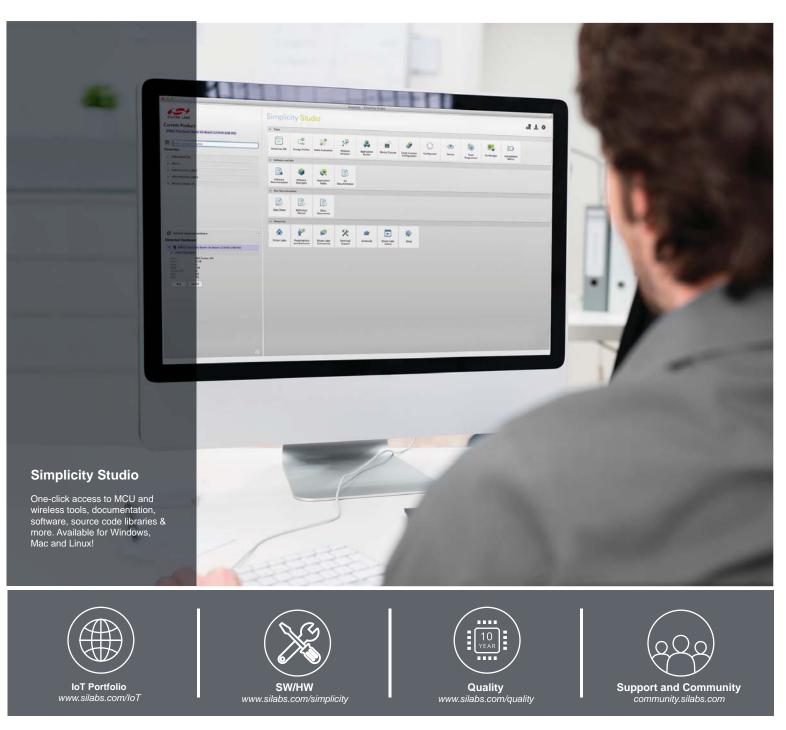
#### **Revision 1.3 to Revision 1.4**

• Added C8051F34C and C8051F34D devices.



NOTES:





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